

AMC1300B-Q1 汽车类 $\pm 250\text{mV}$ 输入、精密增强型隔离放大器

1 特性

- 符合面向汽车应用的 AEC-Q100 标准：
 - 温度等级 1: -40°C 至 125°C 、 T_A
- 提供功能安全
 - 有助于进行功能安全系统设计的文档
- $\pm 250\text{mV}$ 输入电压范围，针对使用分流电阻器测量电流进行了优化
- 固定增益： 8.2 V/V
- 低直流误差：
 - 失调电压误差： $\pm 0.2\text{mV}$ (最大值)
 - 温漂： $\pm 0.9\mu\text{V}/^{\circ}\text{C}$ (最大值)
 - 增益误差： $\pm 0.3\%$ (最大值)
 - 增益漂移： $\pm 30\text{ppm}/^{\circ}\text{C}$ (最大值)
 - 非线性度： 0.03% (最大值)
- 高侧和低侧以 3.3V 或 5V 电压运行
- 失效防护输出
- 高 CMTI： $100\text{kV}/\mu\text{s}$ (最小值)
- 低 EMI，符合 CISPR-11 和 CISPR-25 标准
- 安全相关认证：
 - 符合 DIN EN IEC 60747-17 (VDE 0884-17) 标准的 7071V_{PK} 增强型隔离
 - 符合 UL1577 标准且长达 1 分钟的 $5000\text{V}_{\text{RMS}}$ 隔离

2 应用

- 基于分流电阻器的电流感应，可用于：
 - HEV/EV 充电桩
 - HEV/EV 车载充电器 (OBC)
 - HEV/EV 直流/直流转换器
 - HEV/EV 牵引逆变器

3 说明

AMC1300B-Q1 是一款隔离式精密放大器，此放大器的输出与输入电路由抗电磁干扰性能极强的隔离栅隔开。该隔离栅经认证可提供高达 5kV_{RMS} 的增强型隔离，符合 DIN EN IEC 60747-17 (VDE 0884-17) 和 UL1577 标准，并且可支持高达 $1.5\text{kV}_{\text{RMS}}$ 的工作电压。

该隔离栅可将系统中以不同共模电压电平运行的各器件隔开，并保护电压较低的器件免受高电压冲击。

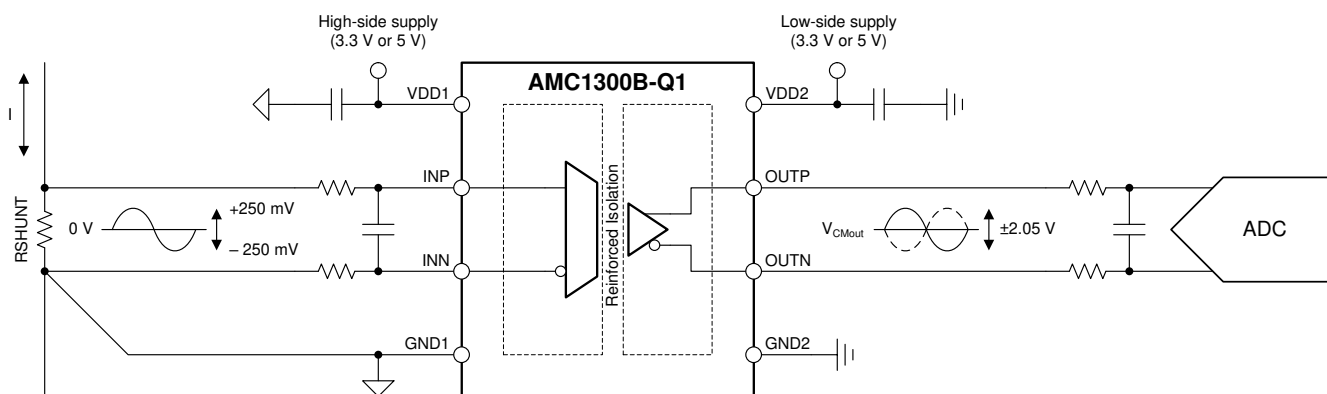
AMC1300B-Q1 的输入针对直接连接低阻抗分流电阻器或其他具有低信号电平的低阻抗电压源的情况进行了优化。出色的直流精度和低温漂支持在 -40°C 至 $+125^{\circ}\text{C}$ 的整个汽车温度范围内，在 PFC 级、直流/直流转换器、牵引逆变器和 OBC 中进行精确的电流控制。

集成的无分流器和无高侧电源检测功能可简化系统级设计和诊断。

器件信息(1)

器件型号	封装	封装尺寸 (标称值)
AMC1300B-Q1	SOIC (8)	$5.85\text{mm} \times 7.50\text{mm}$

(1) 如需了解所有可用封装，请参阅数据表末尾的可订购产品附录。



典型应用



Table of Contents

1 特性	1	7.2 Functional Block Diagram.....	18
2 应用	1	7.3 Feature Description.....	18
3 说明	1	7.4 Device Functional Modes.....	20
4 Revision History	2	8 Application and Implementation	21
5 Pin Configuration and Functions	3	8.1 Application Information.....	21
6 Specifications	4	8.2 Typical Application.....	21
6.1 Absolute Maximum Ratings.....	4	8.3 What to Do and What Not to Do.....	24
6.2 ESD Ratings.....	4	9 Power Supply Recommendations	24
6.3 Recommended Operating Conditions.....	4	10 Layout	25
6.4 Thermal Information.....	5	10.1 Layout Guidelines.....	25
6.5 Power Ratings.....	5	10.2 Layout Example.....	25
6.6 Insulation Specifications.....	6	11 Device and Documentation Support	26
6.7 Safety-Related Certifications.....	7	11.1 Documentation Support.....	26
6.8 Safety Limiting Values.....	7	11.2 接收文档更新通知.....	26
6.9 Electrical Characteristics.....	8	11.3 支持资源.....	26
6.10 Switching Characteristics.....	9	11.4 Trademarks.....	26
6.11 Timing Diagram.....	9	11.5 Electrostatic Discharge Caution.....	26
6.12 Insulation Characteristics Curves.....	10	11.6 术语表.....	26
6.13 Typical Characteristics.....	11	12 Mechanical, Packaging, and Orderable Information	26
7 Detailed Description	18		
7.1 Overview.....	18		

4 Revision History

注：以前版本的页码可能与当前版本的页码不同

Changes from Revision A (January 2021) to Revision B (June 2022)	Page
• 将隔离标准从 <i>DIN VDE V 0884-11 (VDE V 0884-11)</i> 更改为 <i>DIN EN IEC 60747-17 (VDE 0884-17)</i> ，并相应更新了 <i>绝缘规格</i> 和 <i>安全相关认证表</i>	1
• 向 <i>特性</i> 列表添加了 <i>提供功能安全型</i> 要点.....	1
• Changed external clearance (CLR) from ≥ 9 mm to ≥ 8.5 mm.....	6
• Changed C_{IO} from ~ 1 pF to ~ 1.5 pF.....	6

Changes from Revision * (April 2020) to Revision A (January 2021)	Page
• Changed external clearance (CLR) from ≥ 8.5 mm to ≥ 9 mm.....	6
• Changed TCV_{OS} from ± 3 $\mu V/^{\circ}C$ to ± 0.9 $\mu V/^{\circ}C$	8
• Changed TCE_G from ± 50 ppm/ $^{\circ}C$ to ± 30 ppm/ $^{\circ}C$	8
• Changed $VDD1_{UV}$ from 1.75 V (min) / 2.53 V (typ) / 2.7 V (max) to 2.4 V (min) / 2.6 V (typ) / 2.8 V (max).....	8

5 Pin Configuration and Functions

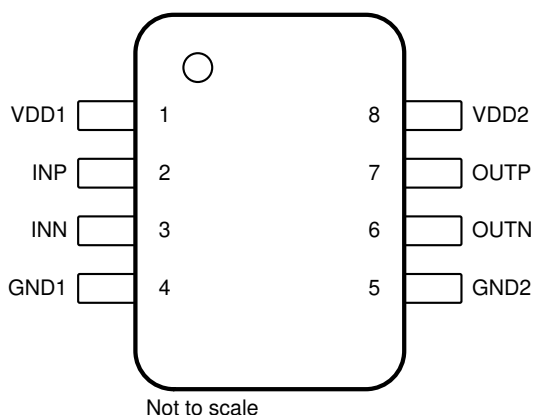


图 5-1. DWV Package, 8-Pin SOIC (Top View)

表 5-1. Pin Functions

PIN		TYPE	DESCRIPTION
NO.	NAME		
1	VDD1	High-side power	High-side power supply. ⁽¹⁾
2	INP	Analog input	Noninverting analog input. Either INP or INN must have a DC current path to GND1 to define the common-mode input voltage. ⁽²⁾
3	INN	Analog input	Inverting analog input. Either INP or INN must have a DC current path to GND1 to define the common-mode input voltage. ⁽²⁾
4	GND1	High-side ground	High-side analog ground.
5	GND2	Low-side ground	Low-side analog ground.
6	OUTN	Analog output	Inverting analog output.
7	OUTP	Analog output	Noninverting analog output.
8	VDD2	Low-side power	Low-side power supply. ⁽¹⁾

(1) See the [Power Supply Recommendations](#) section for power-supply decoupling recommendations.

(2) See the [Layout](#) section for details.

6 Specifications

6.1 Absolute Maximum Ratings

see⁽¹⁾

		MIN	MAX	UNIT
Power-supply voltage	High-side VDD1 to GND1	- 0.3	6.5	V
	Low-side VDD2 to GND2	- 0.3	6.5	
Analog input voltage	INP, INN	GND1 - 6	VDD1 + 0.5	V
Output voltage	OUTP, OUTN	GND2 - 0.5	VDD2 + 0.5	V
Input current	Continuous, any pin except power-supply pins	- 10	10	mA
Temperature	Junction, T _J		150	°C
	Storage, T _{stg}	- 65	150	

- (1) Operation outside the *Absolute Maximum Ratings* may cause permanent device damage. *Absolute Maximum Ratings* do not imply functional operation of the device at these or any other conditions beyond those listed under *Recommended Operating Conditions*. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime

6.2 ESD Ratings

			VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 ⁽¹⁾ , HBM ESD classification Level 2	±2000	V
		Charged-device model (CDM), per AEC Q100-011, CDM ESD classification Level C6	±1000	

- (1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

6.3 Recommended Operating Conditions

over operating ambient temperature range (unless otherwise noted)

			MIN	NOM	MAX	UNIT
POWER SUPPLY						
	High-side power supply	VDD1 to GND1	3	5	5.5	V
	Low-side power supply	VDD2 to GND2	3	3.3	5.5	V
ANALOG INPUT						
V _{Clipping}	Differential input voltage before clipping output	V _{IN} = V _{INP} - V _{INN}		±320		mV
V _{FSR}	Specified linear differential full-scale voltage	V _{IN} = V _{INP} - V _{INN}	- 250		250	mV
V _{CM}	Operating common-mode input voltage	(V _{INP} + V _{INN}) / 2 to GND1	- 0.16	VDD1 - 2.1		V
TEMPERATURE RANGE						
T _A	Specified ambient temperature		- 40		125	°C

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		AMC1300B-Q1	UNIT
		DWV (SOIC)	
		8 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	85.4	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	26.8	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	43.5	°C/W
ψ_{JT}	Junction-to-top characterization parameter	4.8	°C/W
ψ_{JB}	Junction-to-board characterization parameter	41.2	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	n/a	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

6.5 Power Ratings

PARAMETER		TEST CONDITIONS	VALUE	UNIT
P_D	Maximum power dissipation (both sides)	VDD1 = VDD2 = 5.5 V	99	mW
P_{D1}	Maximum power dissipation (high-side)	VDD1 = 3.6 V	31	mW
		VDD1 = 5.5 V	54	
P_{D2}	Maximum power dissipation (low-side)	VDD2 = 3.6 V	26	mW
		VDD2 = 5.5 V	45	

6.6 Insulation Specifications

over operating ambient temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	VALUE	UNIT
GENERAL				
CLR	External clearance ⁽¹⁾	Shortest pin-to-pin distance through air	≥ 8.5	mm
CPG	External creepage ⁽¹⁾	Shortest pin-to-pin distance across the package surface	≥ 8.5	mm
DTI	Distance through insulation	Minimum internal gap (internal clearance) of the double insulation	≥ 0.021	mm
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	≥ 600	V
	Material group	According to IEC 60664-1	I	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 600 V _{RMS}	I-IV	
		Rated mains voltage ≤ 1000 V _{RMS}	I-III	
DIN EN IEC 60747-17 (VDE 0884-17) ⁽²⁾				
V _{IORM}	Maximum repetitive peak isolation voltage	At AC voltage	2120	V _{PK}
V _{IOWM}	Maximum-rated isolation working voltage	At AC voltage (sine wave)	1500	V _{RMS}
		At DC voltage	2120	V _{DC}
V _{IOTM}	Maximum transient isolation voltage	V _{TEST} = V _{IOTM} , t = 60 s (qualification test)	7000	V _{PK}
		V _{TEST} = 1.2 × V _{IOTM} , t = 1 s (100% production test)	8400	
V _{IMP}	Maximum impulse voltage ⁽³⁾	Tested in air, 1.2/50-μs waveform per IEC 62368-1	9800	V _{PK}
V _{IOSM}	Maximum surge isolation voltage ⁽⁴⁾	Tested in oil (qualification test), 1.2/50-μs waveform per IEC 62368-1	12800	V _{PK}
q _{pd}	Apparent charge ⁽⁵⁾	Method a, after input/output safety test subgroups 2 and 3, V _{ini} = V _{IOTM} , t _{ini} = 60 s, V _{pd(m)} = 1.2 × V _{IORM} , t _m = 10 s	≤ 5	pC
		Method a, after environmental tests subgroup 1, V _{ini} = V _{IOTM} , t _{ini} = 60 s, V _{pd(m)} = 1.6 × V _{IORM} , t _m = 10 s	≤ 5	
		Method b1, at routine test (100% production) and preconditioning (type test), V _{ini} = V _{IOTM} , t _{ini} = 1 s, V _{pd(m)} = 1.875 × V _{IORM} , t _m = 1 s	≤ 5	
C _{IO}	Barrier capacitance, input to output ⁽⁶⁾	V _{IO} = 0.5 V _{PP} at 1 MHz	~1.5	pF
R _{IO}	Insulation resistance, input to output ⁽⁶⁾	V _{IO} = 500 V at T _A = 25°C	> 10 ¹²	Ω
		V _{IO} = 500 V at 100°C ≤ T _A ≤ 125°C	> 10 ¹¹	
		V _{IO} = 500 V at T _S = 150°C	> 10 ⁹	
	Pollution degree		2	
	Climatic category		55/125/21	
UL1577				
V _{ISO}	Withstand isolation voltage	V _{TEST} = V _{ISO} = 5000 V _{RMS} , t = 60 s (qualification), V _{TEST} = 1.2 × V _{ISO} = 6000 V _{RMS} , t = 1 s (100% production test)	5000	V _{RMS}

- (1) Apply creepage and clearance requirements according to the specific equipment isolation standards of an application. Care must be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed circuit board (PCB) do not reduce this distance. Creepage and clearance on a PCB become equal in certain cases. Techniques such as inserting grooves, ribs, or both on a PCB are used to help increase these specifications.
- (2) This coupler is suitable for *safe electrical insulation* only within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.
- (3) Testing is carried out in air to determine the surge immunity of the package.
- (4) Testing is carried in oil to determine the intrinsic surge immunity of the isolation barrier.
- (5) Apparent charge is electrical discharge caused by a partial discharge (pd).
- (6) All pins on each side of the barrier are tied together, creating a two-pin device.

6.7 Safety-Related Certifications

VDE	UL
DIN EN IEC 60747-17 (VDE 0884-17), EN IEC 60747-17, DIN EN IEC 62368-1 (VDE 0868-1), EN IEC 62368-1, IEC 62368-1 Clause : 5.4.3 ; 5.4.4.4 ; 5.4.9	Recognized under 1577 component recognition
Reinforced insulation	Single protection
Certificate number: 40040142	File number: E181974

6.8 Safety Limiting Values

Safety limiting⁽¹⁾ intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry. A failure of the I/O can allow low resistance to ground or the supply and, without current limiting, dissipate sufficient power to over-heat the die and damage the isolation barrier potentially leading to secondary system failures.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I _S	Safety input, output, or supply current	R _{θJA} = 85.4°C/W, VDDx = 5.5 V, T _J = 150°C, T _A = 25°C			266	mA
I _S	Safety input, output, or supply current	R _{θJA} = 85.4°C/W, VDDx = 3.6 V, T _J = 150°C, T _A = 25°C			407	mA
P _S	Safety input, output, or total power	R _{θJA} = 85.4°C/W, T _J = 150°C, T _A = 25°C			1464	mW
T _S	Maximum safety temperature				150	°C

- (1) The maximum safety temperature, T_S, has the same value as the maximum junction temperature, T_J, specified for the device. The I_S and P_S parameters represent the safety current and safety power, respectively. Do not exceed the maximum limits of I_S and P_S. These limits vary with the ambient temperature, T_A.

The junction-to-air thermal resistance, R_{θJA}, in the Thermal Information table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:

T_J = T_A + R_{θJA} × P, where P is the power dissipated in the device.

T_{J(max)} = T_S = T_A + R_{θJA} × P_S, where T_{J(max)} is the maximum junction temperature.

P_S = I_S × VDD_{max}, where VDD_{max} is the maximum supply voltage for high-side and low-side.

6.9 Electrical Characteristics

minimum and maximum specifications apply from $T_A = -40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$, $V_{DD1} = 3.0\text{ V}$ to 5.5 V , $V_{DD2} = 3.0\text{ V}$ to 5.5 V , $\text{INP} = -250\text{ mV}$ to $+250\text{ mV}$, and $\text{INN} = \text{GND1}$; typical specifications are at $T_A = 25^{\circ}\text{C}$, $V_{DD1} = 5\text{ V}$, and $V_{DD2} = 3.3\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ANALOG INPUT						
V _{CMov}	Common-mode overvoltage detection level	(V _{INP} + V _{INN}) / 2 to GND1	VDD1 - 2			V
	Hysteresis of common-mode overvoltage detection level		60			mV
V _{OS}	Input offset voltage ⁽¹⁾	Initial, at T _A = 25°C, INP = INN = GND1	- 0.2	±0.01	0.2	mV
TCV _{OS}	Input offset drift ^{(1) (4)}		- 0.9	±0.1	0.9	μV/°C
CMRR	Common-mode rejection ratio	f _{IN} = 0 Hz, V _{CM min} ≤ V _{CM} ≤ V _{CM max}	- 100			dB
		f _{IN} = 10 kHz, V _{CM min} ≤ V _{CM} ≤ V _{CM max}	- 98			
R _{IN}	Single-ended input resistance	INN = GND1	19			k Ω
R _{IND}	Differential input resistance		22			k Ω
I _{IB}	Input bias current	INP = INN = GND1; I _{IB} = (I _{IBP} + I _{IBN}) / 2	- 41	- 30	- 24	μA
I _{IO}	Input offset current	I _{IO} = I _{IBP} - I _{IBN} ; INP = INN = GND1	±5			nA
C _{IN}	Single-ended input capacitance	INN = GND1, f _{IN} = 275 kHz	2			pF
C _{IND}	Differential input capacitance	f _{IN} = 275 kHz	1			pF
ANALOG OUTPUT						
	Nominal gain		8.2			V/V
E _G	Gain error ⁽¹⁾	at T _A = 25°C	- 0.3%	±0.04%	0.3%	
TCE _G	Gain drift ^{(1) (5)}		- 30	±5	30	ppm/°C
	Nonlinearity ⁽¹⁾		- 0.03%	±0.01%	0.03%	
THD	Total harmonic distortion ⁽³⁾	f _{IN} = 10 kHz	- 85			dB
	Output noise	INP = INN = GND1, f _{IN} = 0 Hz, BW = 100 kHz brickwall filter	230			μV _{RMS}
SNR	Signal-to-noise ratio	f _{IN} = 1 kHz, BW = 10 kHz	81.5	85		dB
		f _{IN} = 10 kHz, BW = 100 kHz	72			
PSRR	Power-supply rejection ratio ⁽²⁾	PSRR vs VDD1, at DC	- 103			dB
		PSRR vs VDD1, 100-mV and 10-kHz ripple	- 96			
		PSRR vs VDD2, at DC	- 106			
		PSRR vs VDD2, 100-mV and 10-kHz ripple	- 86			
V _{CMout}	Common-mode output voltage		1.39	1.44	1.49	V
V _{CLIPout}	Clipping differential output voltage	V _{OUT} = (V _{OUTP} - V _{OUTN}); V _{IN} = V _{INP} - V _{INN} > V _{Clipping}	- 2.52	±2.49	2.52	V
V _{Failsafe}	Failsafe differential output voltage	V _{CM} ≥ V _{CMov} , or VDD1 missing	- 2.63	- 2.57	- 2.53	V
BW	Output bandwidth		250	310		kHz
R _{OUT}	Output resistance	On OUTP or OUTN	< 0.2			Ω
	Output short-circuit current	On OUTP or OUTN, sourcing or sinking, INN = INP = GND1, outputs shorted to either GND2 or VDD2	14			mA
CMTI	Common-mode transient immunity	GND1 - GND2 = 1 kV	100	150		kV/μs

6.9 Electrical Characteristics (continued)

minimum and maximum specifications apply from $T_A = -40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$, $V_{DD1} = 3.0\text{ V}$ to 5.5 V , $V_{DD2} = 3.0\text{ V}$ to 5.5 V , $\text{INP} = -250\text{ mV}$ to $+250\text{ mV}$, and $\text{INN} = \text{GND1}$; typical specifications are at $T_A = 25^{\circ}\text{C}$, $V_{DD1} = 5\text{ V}$, and $V_{DD2} = 3.3\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
POWER SUPPLY						
VDD1 _{UV}	VDD1 undervoltage detection threshold voltage	VDD1 falling	2.4	2.6	2.8	V
IDD1	High-side supply current	3.0 V ≤ VDD1 ≤ 3.6 V	6.3		8.5	mA
		4.5 V ≤ VDD1 ≤ 5.5 V	7.2		9.8	
IDD2	Low-side supply current	3.0 V ≤ VDD2 ≤ 3.6 V	5.3		7.2	mA
		4.5 V ≤ VDD2 ≤ 5.5 V	5.9		8.1	

- (1) The typical value includes one standard deviation (σ) at nominal operating conditions.
- (2) This parameter is input referred.
- (3) THD is the ratio of the rms sum of the amplitudes of first five higher harmonics to the amplitude of the fundamental.
- (4) Offset error temperature drift is calculated using the box method, as described by the following equation:

$$TCV_{OS} = (V_{OS,MAX} - V_{OS,MIN}) / \text{TempRange}$$
where $V_{OS,MAX}$ and $V_{OS,MIN}$ refer to the maximum and minimum V_{OS} values measured within the temperature range (-40 to 125°C).
- (5) Gain error temperature drift is calculated using the box method, as described by the following equation:

$$TCE_G (\text{ppm}) = ((E_{G,MAX} - E_{G,MIN}) / \text{TempRange}) \times 10^4$$
where $E_{G,MAX}$ and $E_{G,MIN}$ refer to the maximum and minimum E_G values (in %) measured within the temperature range (-40 to 125°C).

6.10 Switching Characteristics

over operating ambient temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_r	Output signal rise time		1.3		μs
t_f	Output signal fall time		1.3		μs
	V_{INx} to V_{OUTx} signal delay (50% - 10%)	Unfiltered output	1	1.5	μs
	V_{INx} to V_{OUTx} signal delay (50% - 50%)	Unfiltered output	1.6	2.1	μs
	V_{INx} to V_{OUTx} signal delay (50% - 90%)	Unfiltered output	2.5	3	μs
t_{AS}	Analog settling time	VDD1 step to 3.0 V with $V_{DD2} \geq 3.0\text{ V}$, to V_{OUTP} , V_{OUTN} valid, 0.1% settling	500		μs

6.11 Timing Diagram

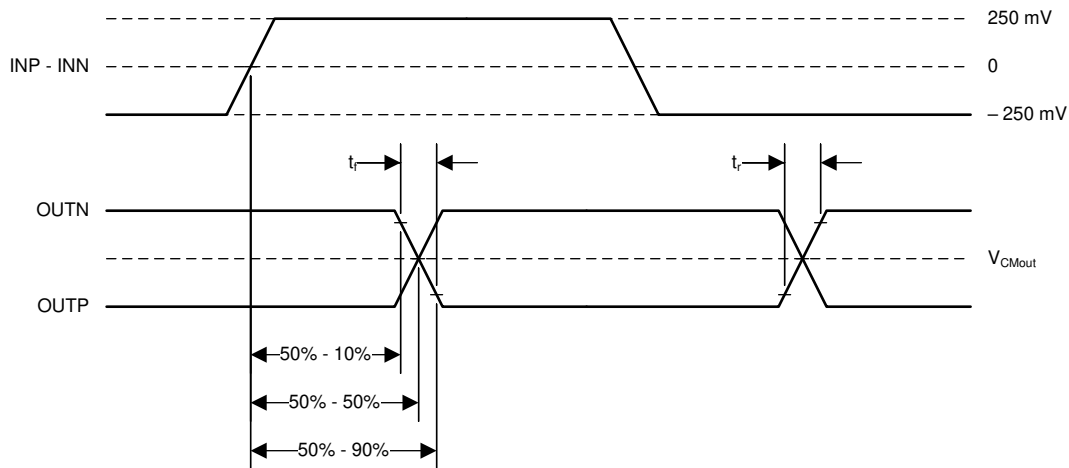


图 6-1. Rise, Fall, and Delay Time Definition

6.12 Insulation Characteristics Curves

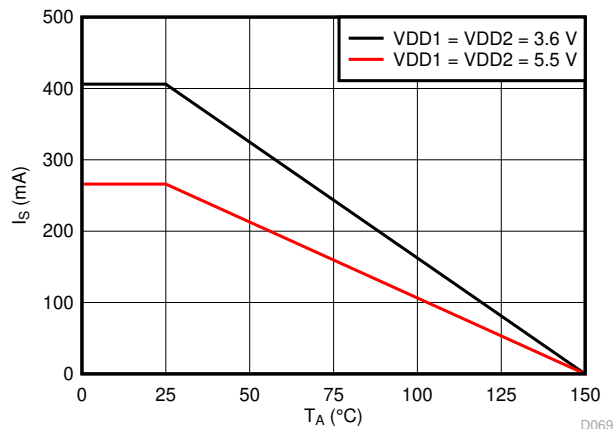


图 6-2. Thermal Derating Curve for Safety-Limiting Current per VDE

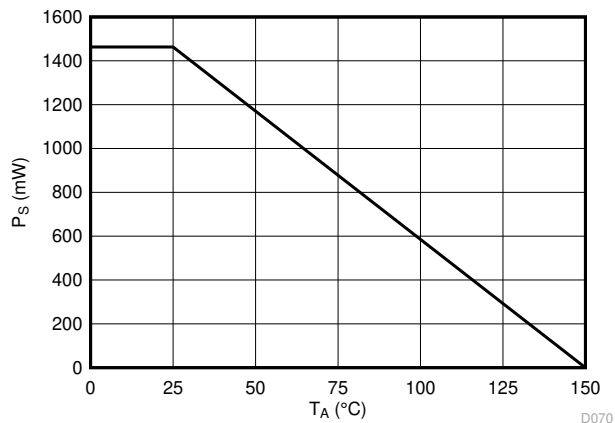
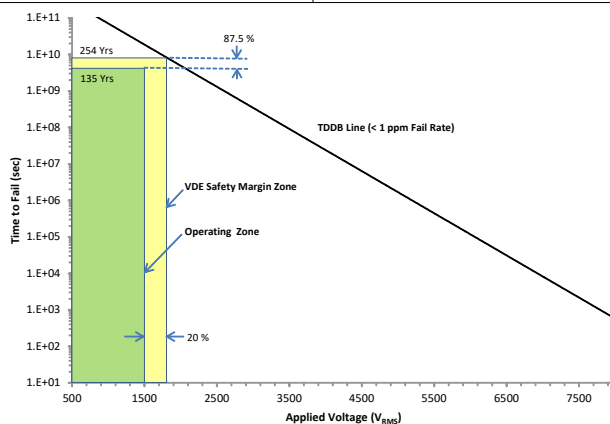


图 6-3. Thermal Derating Curve for Safety-Limiting Power per VDE



T_A up to 150°C, stress-voltage frequency = 60 Hz, isolation working voltage = 1500 V_{RMS} , operating lifetime = 135 year

图 6-4. Reinforced Isolation Capacitor Lifetime Projection

6.13 Typical Characteristics

at VDD1 = 5 V, VDD2 = 3.3 V, INP = -250 mV to 250 mV, INN = 0 V, and f_{IN} = 10 kHz (unless otherwise noted)

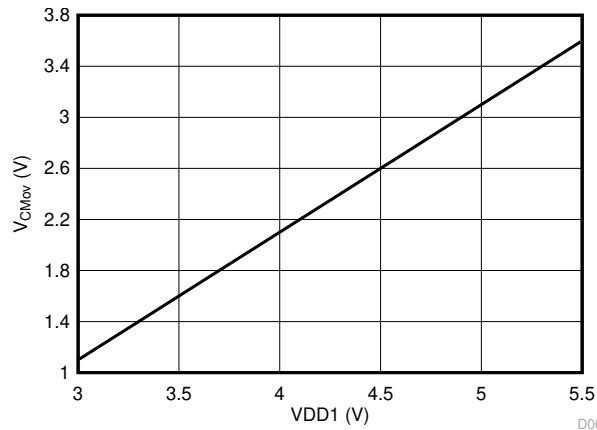


图 6-5. Common-Mode Overvoltage Detection Level vs High-Side Supply Voltage

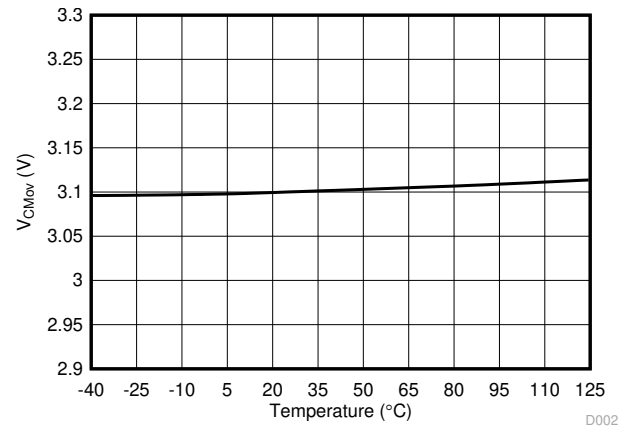


图 6-6. Common-Mode Overvoltage Detection Level vs Temperature

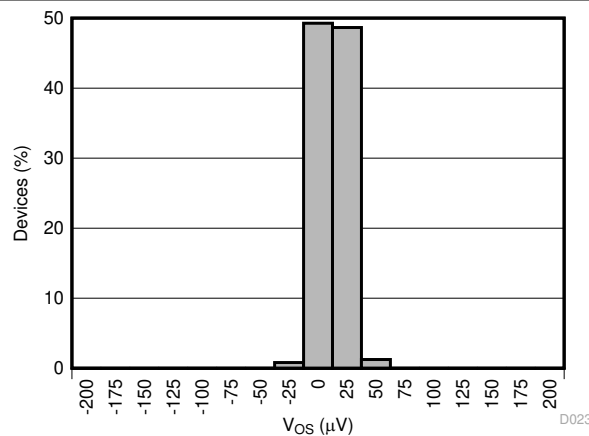


图 6-7. Input Offset Voltage Histogram

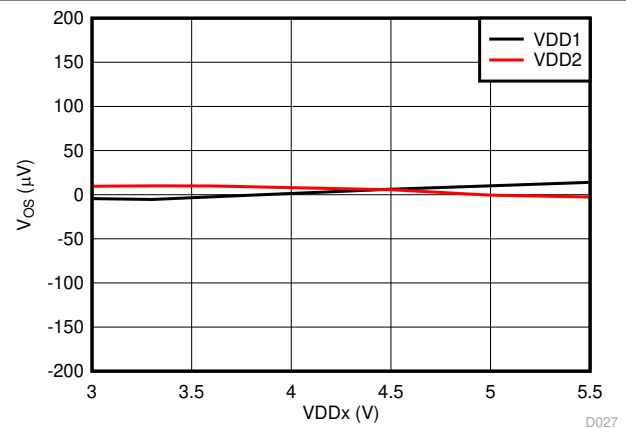


图 6-8. Input Offset Voltage vs Supply Voltage

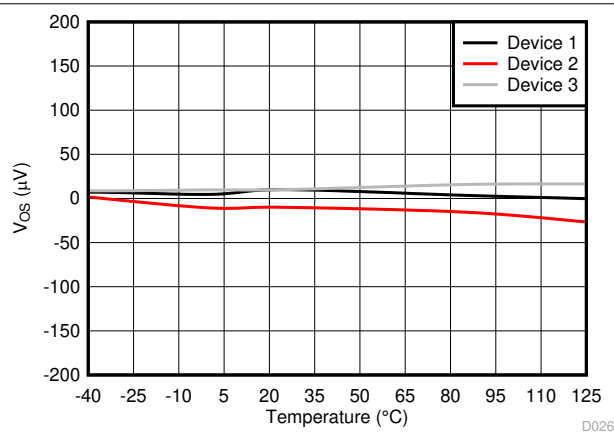


图 6-9. Input Offset Voltage vs Temperature

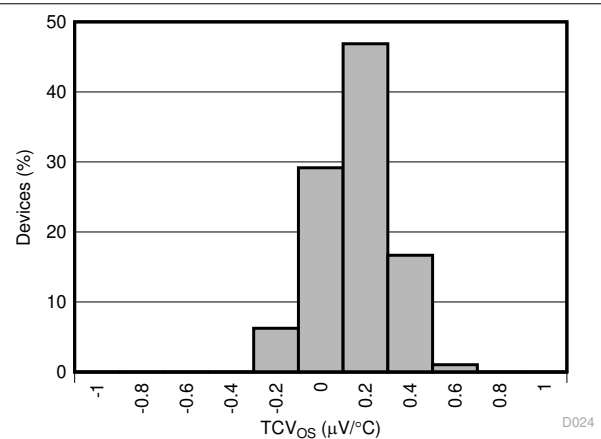


图 6-10. Input Offset Drift Histogram

6.13 Typical Characteristics (continued)

at VDD1 = 5 V, VDD2 = 3.3 V, INP = -250 mV to 250 mV, INN = 0 V, and f_{IN} = 10 kHz (unless otherwise noted)

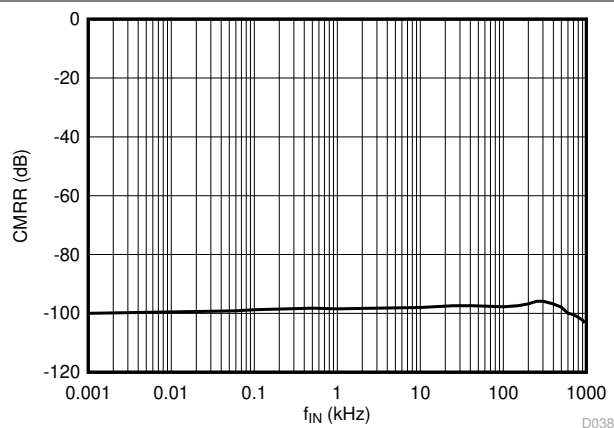


图 6-11. Common-Mode Rejection Ratio vs Input Frequency

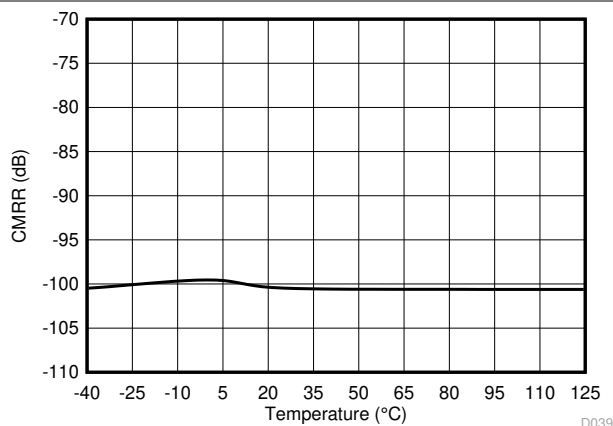


图 6-12. Common-Mode Rejection Ratio vs Temperature

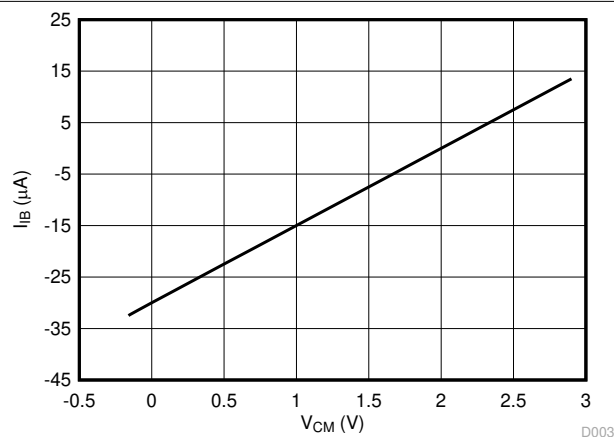


图 6-13. Input Bias Current vs Common-Mode Input Voltage

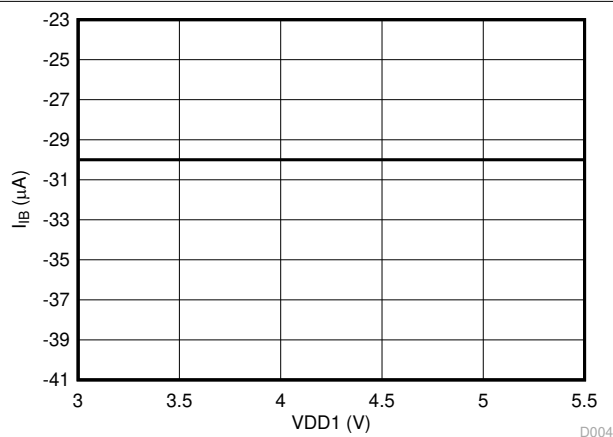


图 6-14. Input Bias Current vs High-Side Supply Voltage

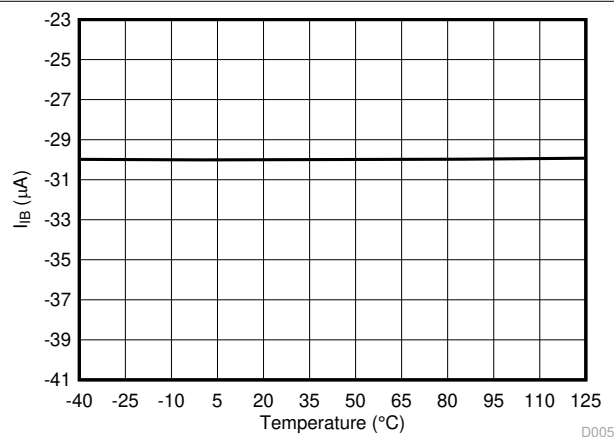


图 6-15. Input Bias Current vs Temperature

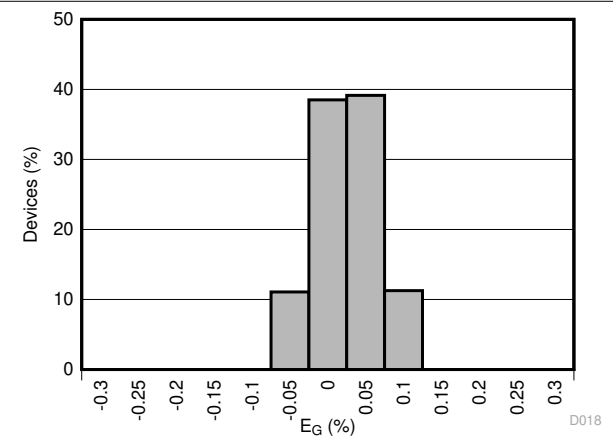


图 6-16. Gain Error Histogram

6.13 Typical Characteristics (continued)

at VDD1 = 5 V, VDD2 = 3.3 V, INP = -250 mV to 250 mV, INN = 0 V, and $f_{IN} = 10$ kHz (unless otherwise noted)

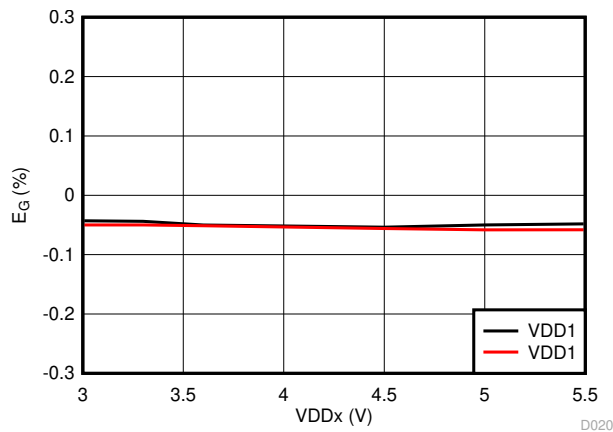


图 6-17. Gain Error vs Supply Voltage

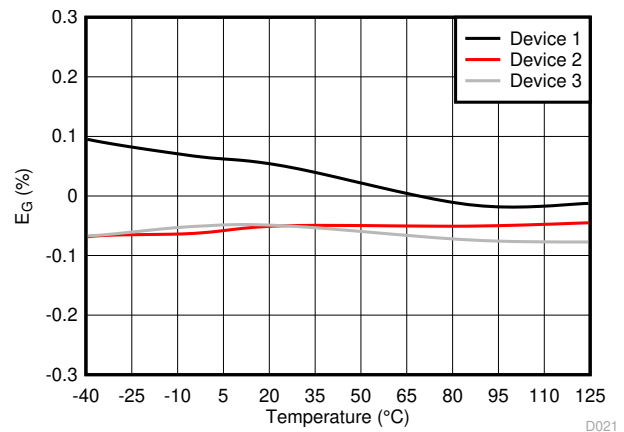


图 6-18. Gain Error vs Temperature

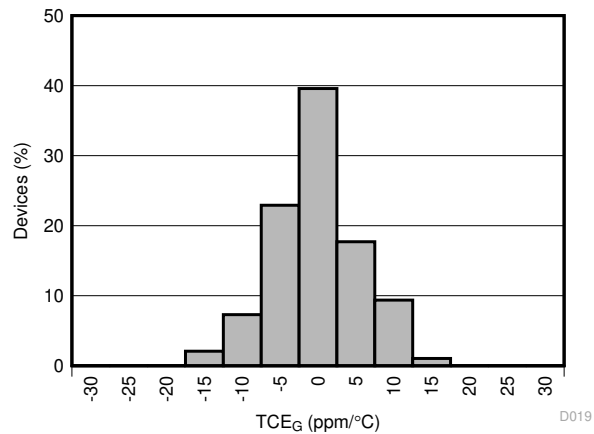


图 6-19. Gain Error Drift Histogram

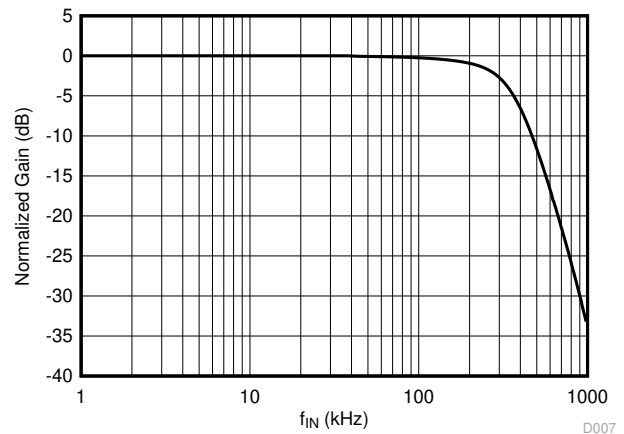


图 6-20. Normalized Gain vs Input Frequency

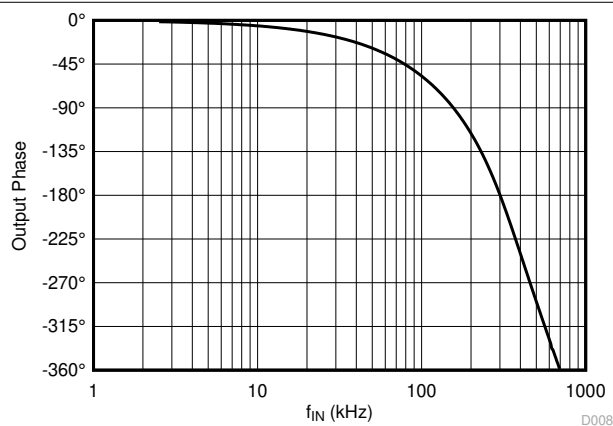


图 6-21. Output Phase vs Input Frequency

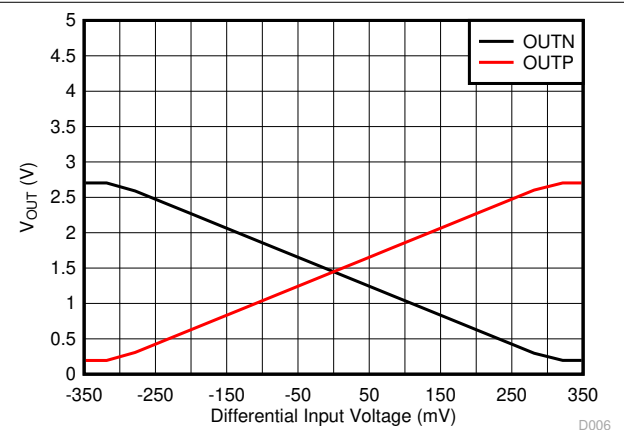


图 6-22. Output Voltage vs Input Voltage

6.13 Typical Characteristics (continued)

at VDD1 = 5 V, VDD2 = 3.3 V, INP = -250 mV to 250 mV, INN = 0 V, and f_{IN} = 10 kHz (unless otherwise noted)

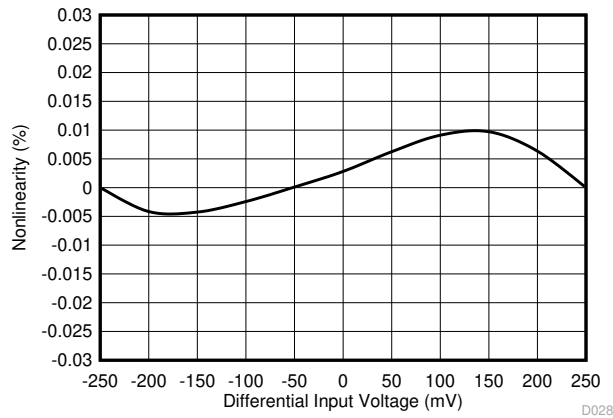


图 6-23. Nonlinearity vs Input Voltage

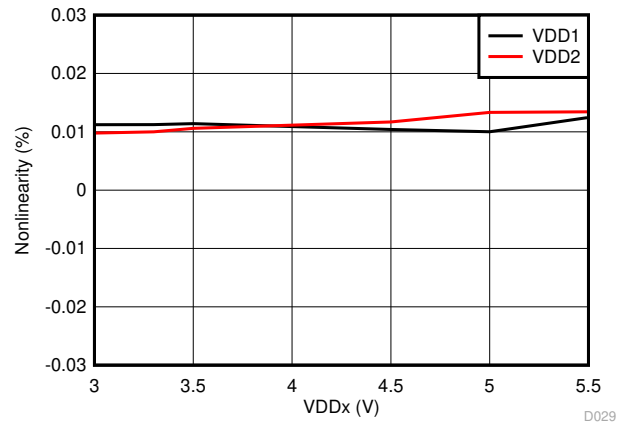


图 6-24. Nonlinearity vs Supply Voltage

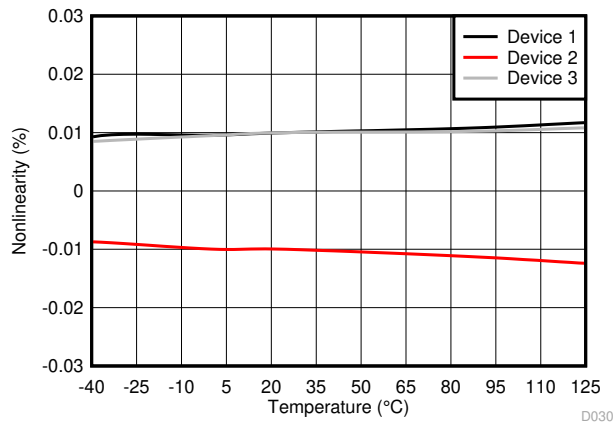


图 6-25. Nonlinearity vs Temperature

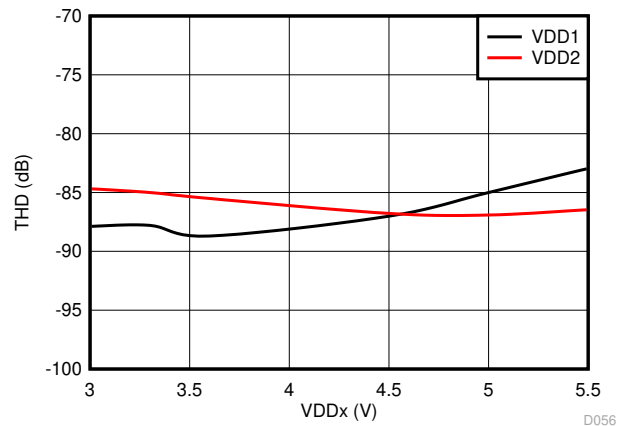


图 6-26. Total Harmonic Distortion vs Supply Voltage

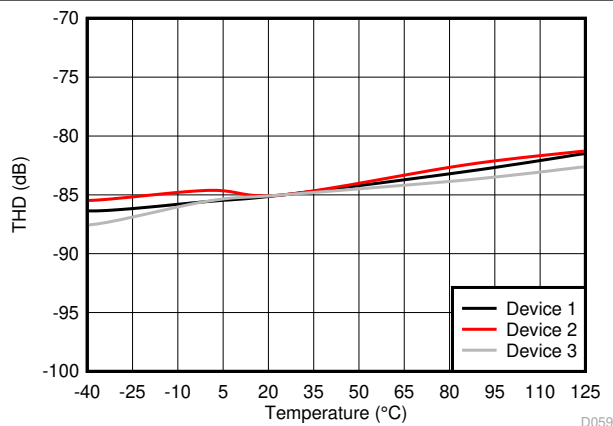


图 6-27. Total Harmonic Distortion vs Temperature

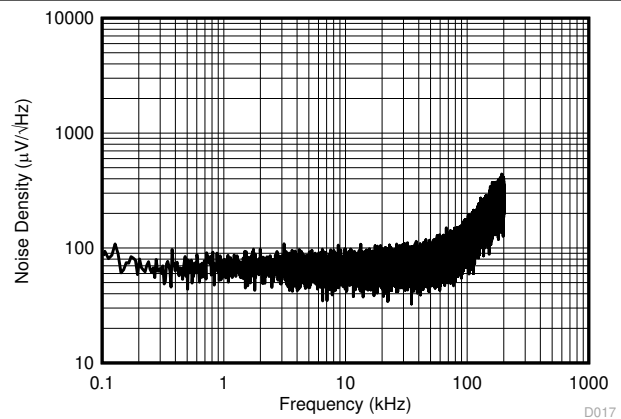


图 6-28. Input-Referred Noise Density vs Frequency

6.13 Typical Characteristics (continued)

at VDD1 = 5 V, VDD2 = 3.3 V, INP = -250 mV to 250 mV, INN = 0 V, and f_{IN} = 10 kHz (unless otherwise noted)

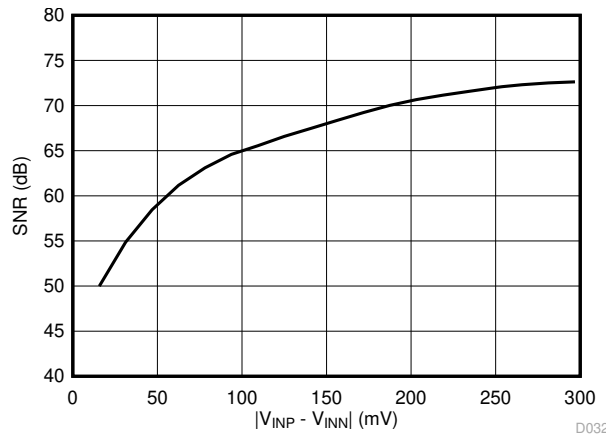


图 6-29. Signal-to-Noise Ratio vs Input Voltage

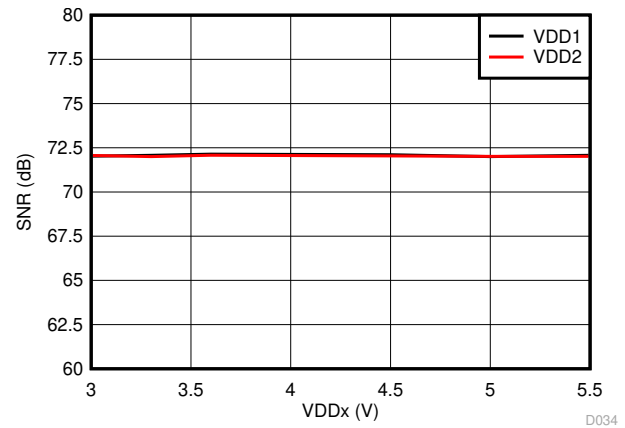


图 6-30. Signal-to-Noise Ratio vs Supply Voltage

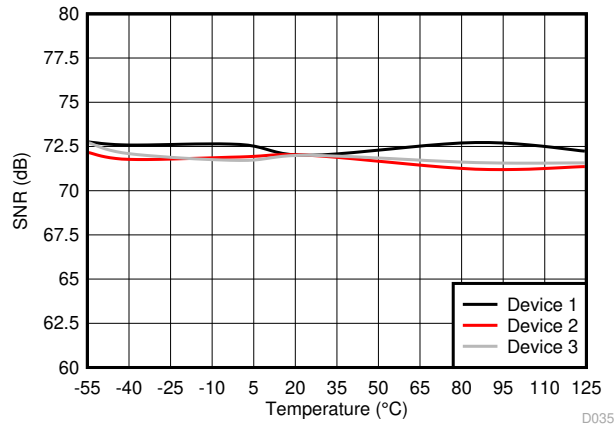


图 6-31. Signal-to-Noise Ratio vs Temperature

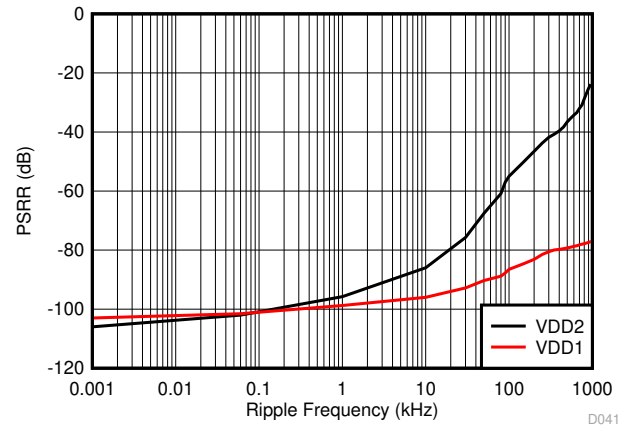


图 6-32. Power-Supply Rejection Ratio vs Ripple Frequency

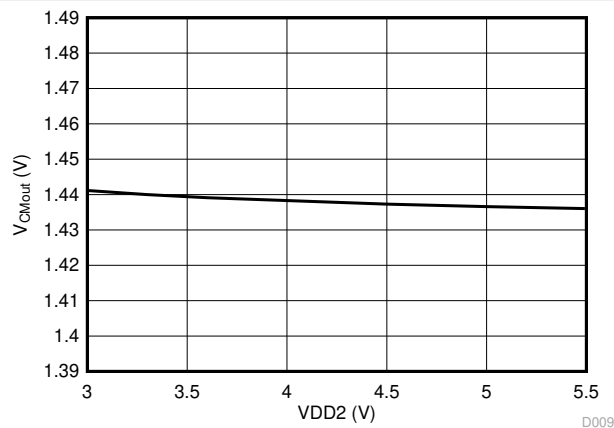


图 6-33. Output Common-Mode Voltage vs Low-Side Supply Voltage

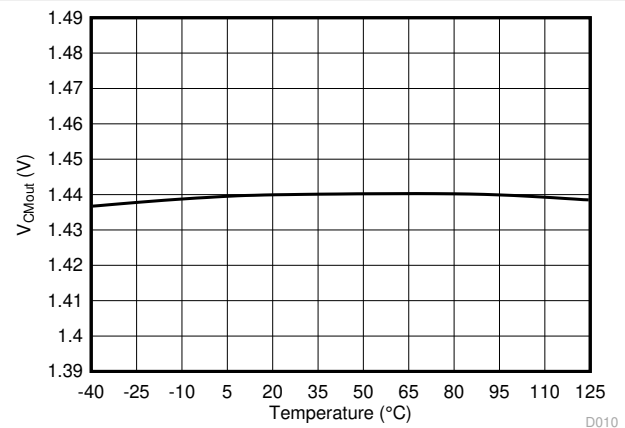


图 6-34. Output Common-Mode Voltage vs Temperature

6.13 Typical Characteristics (continued)

at VDD1 = 5 V, VDD2 = 3.3 V, INP = -250 mV to 250 mV, INN = 0 V, and $f_{IN} = 10$ kHz (unless otherwise noted)

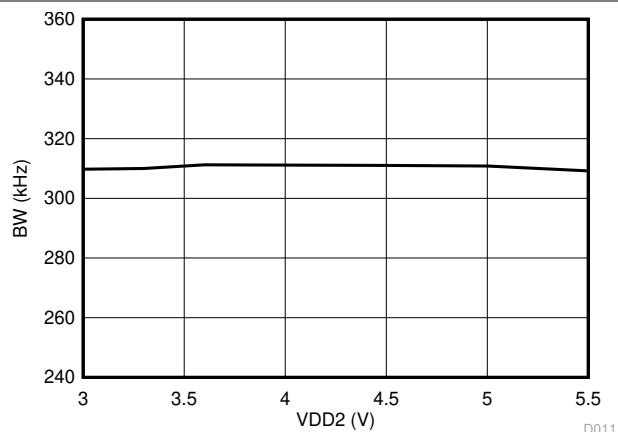


图 6-35. Output Bandwidth vs Low-Side Supply Voltage

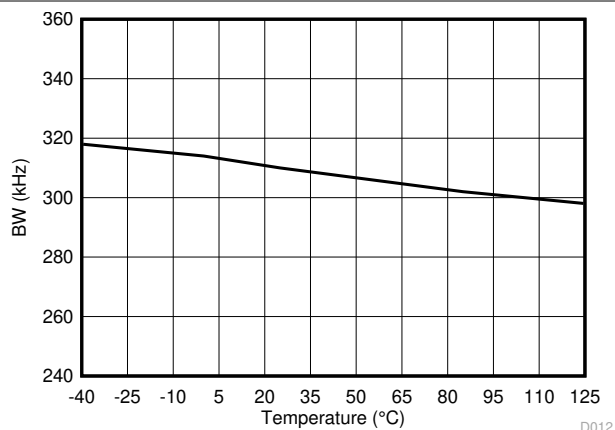


图 6-36. Output Bandwidth vs Temperature

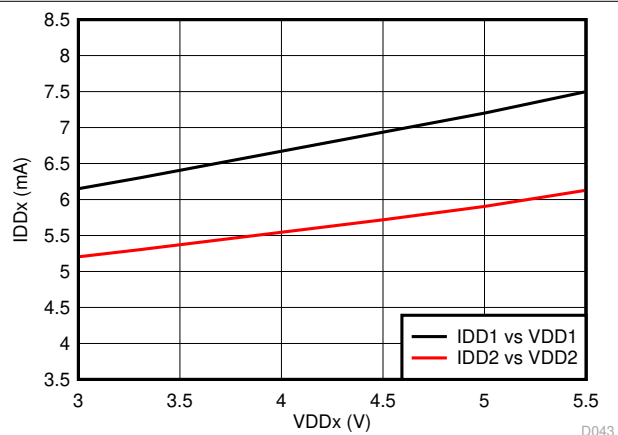


图 6-37. Supply Current vs Supply Voltage

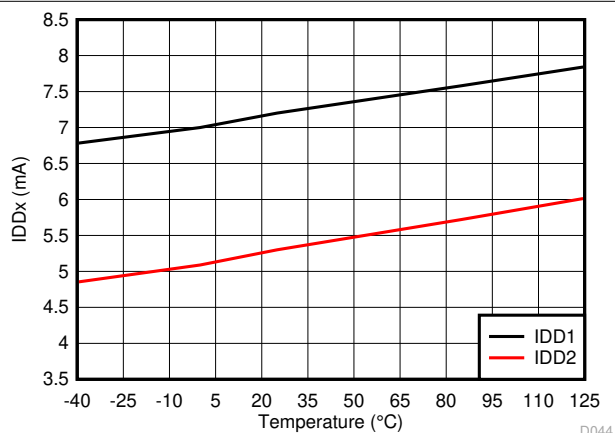


图 6-38. Supply Current vs Temperature

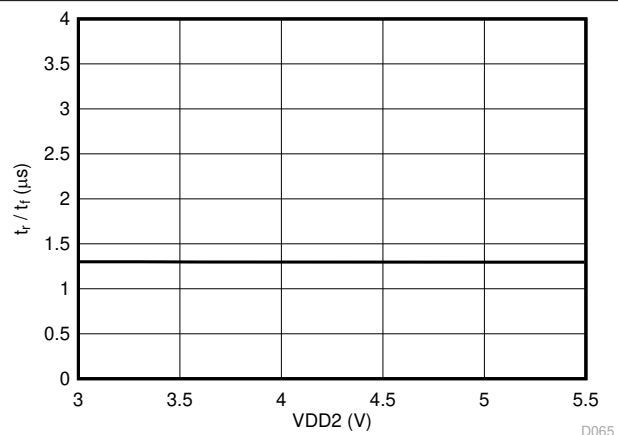


图 6-39. Output Rise and Fall Time vs Low-Side Supply

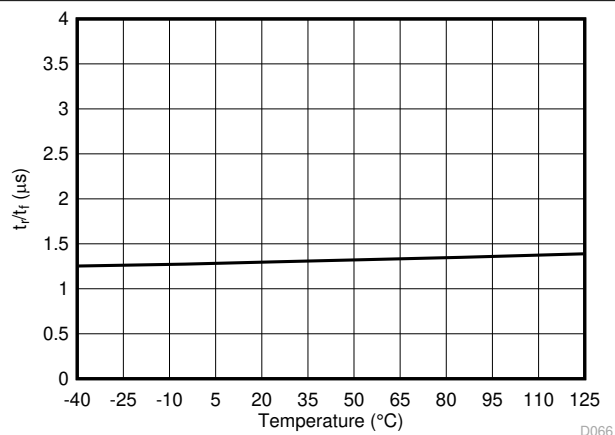


图 6-40. Output Rise and Fall Time vs Temperature

6.13 Typical Characteristics (continued)

at VDD1 = 5 V, VDD2 = 3.3 V, INP = -250 mV to 250 mV, INN = 0 V, and $f_{IN} = 10$ kHz (unless otherwise noted)

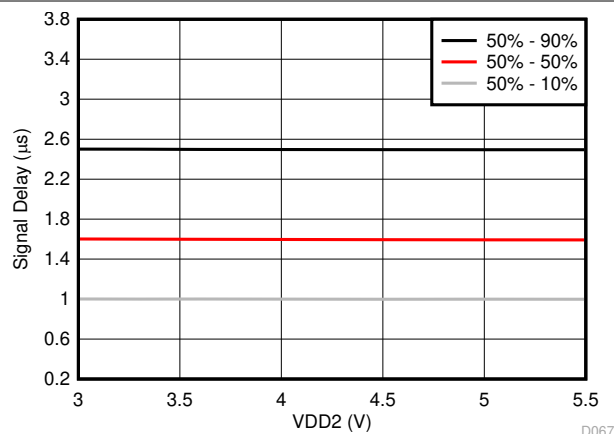


图 6-41. V_{IN} to V_{OUT} Signal Delay vs Low-Side Supply Voltage

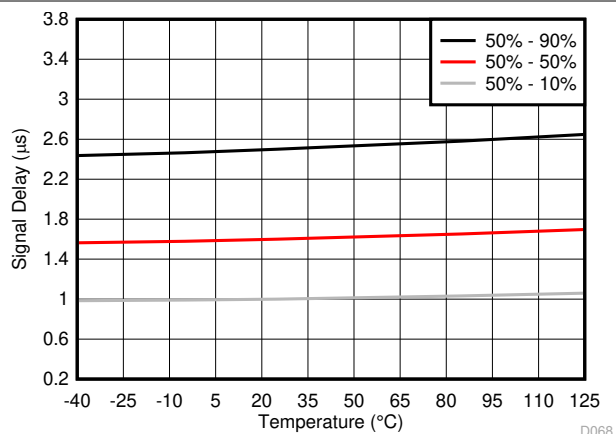


图 6-42. V_{IN} to V_{OUT} Signal Delay vs Temperature

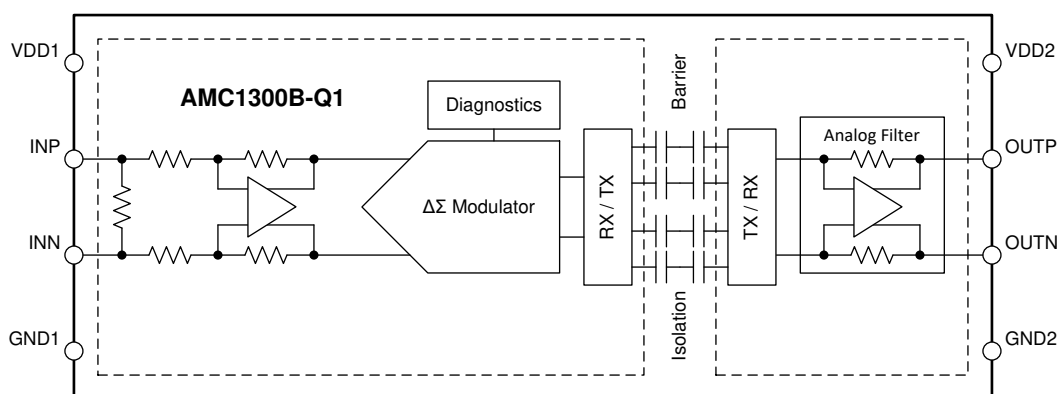
7 Detailed Description

7.1 Overview

The AMC1300B-Q1 is a fully differential, precision, isolated amplifier. The input stage of the device consists of a fully differential amplifier that drives a second-order, delta-sigma ($\Delta\Sigma$) modulator. The modulator converts the analog input signal into a digital bitstream that is transferred across the isolation barrier that separates the high-side from the low-side. On the low-side, the received bitstream is processed by a fourth-order analog filter that outputs a differential signal at the OUTP and OUTN pins that is proportional to the input signal.

The SiO₂-based, capacitive isolation barrier supports a high level of magnetic field immunity, as described in the [ISO72x Digital Isolator Magnetic-Field Immunity application report](#). The digital modulation used in the AMC1300B-Q1 to transmit data across the isolation barrier, and the isolation barrier characteristics itself, result in high reliability and common-mode transient immunity.

7.2 Functional Block Diagram



7.3 Feature Description

7.3.1 Analog Input

The differential amplifier input stage of the AMC1300B-Q1 feeds a second-order, switched-capacitor, feed-forward $\Delta\Sigma$ modulator. The gain of the differential amplifier is set by internal precision resistors with a differential input impedance of R_{IND} . The modulator converts the analog input signal into a bitstream that is transferred across the isolation barrier, as described in the [Isolation Channel Signal Transmission](#) section.

There are two restrictions on the analog input signals INP and INN. First, if the input voltages V_{INP} or V_{INN} exceed the range specified in the [Absolute Maximum Ratings](#) table, the input currents must be limited to the absolute maximum value, because the electrostatic discharge (ESD) protection turns on. In addition, the linearity and parametric performance of the device are ensured only when the analog input voltage remains within the linear full-scale range (V_{FSR}) and within the common-mode input voltage range (V_{CM}) as specified in the [Recommended Operating Conditions](#) table.

7.3.2 Isolation Channel Signal Transmission

The AMC1300B-Q1 uses an on-off keying (OOK) modulation scheme, as shown in 图 7-1, to transmit the modulator output bitstream across the SiO₂-based isolation barrier. The transmit driver (TX) shown in the [Functional Block Diagram](#) transmits an internally-generated, high-frequency carrier across the isolation barrier to represent a digital *one* and does not send a signal to represent a digital *zero*. The nominal frequency of the carrier used inside the AMC1300B-Q1 is 480 MHz.

The receiver (RX) on the other side of the isolation barrier recovers and demodulates the signal and provides the input to the 4th-order analog filter. The AMC1300B-Q1 transmission channel is optimized to achieve the highest level of common-mode transient immunity (CMTI) and lowest level of radiated emissions caused by the high-frequency carrier and RX/TX buffer switching.

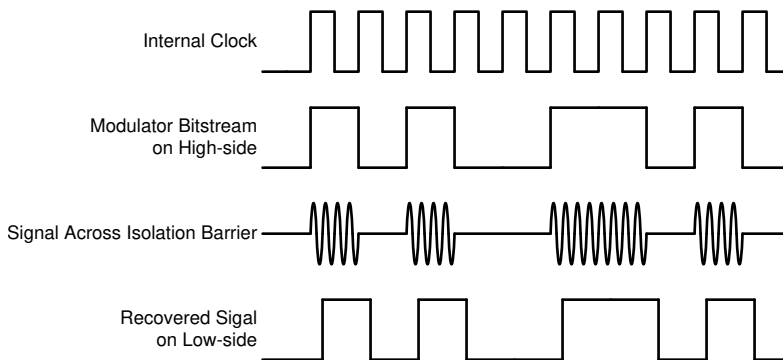


图 7-1. OOK-Based Modulation Scheme

7.3.3 Analog Output

The AMC1300B-Q1 offers a differential analog output comprised of the OUTP and OUTN pins. For differential input voltages ($V_{INP} - V_{INN}$) in the range from -250 mV to 250 mV , the device provides a linear response with a nominal gain of 8.2. For example, for a differential input voltage of 250 mV , the differential output voltage ($V_{OUTP} - V_{OUTN}$) is 2.05 V . At zero input (INP shorted to INN), both pins output the same common-mode output voltage V_{CMout} , as specified in the [Electrical Characteristics](#) table. For absolute differential input voltages greater than 250 mV but less than 320 mV , the differential output voltage continues to increase in magnitude but with reduced linearity performance. The outputs saturate at a differential output voltage of $V_{CLIPout}$, as shown in [Figure 7-2](#), if the differential input voltage exceeds the $V_{Clipping}$ value.

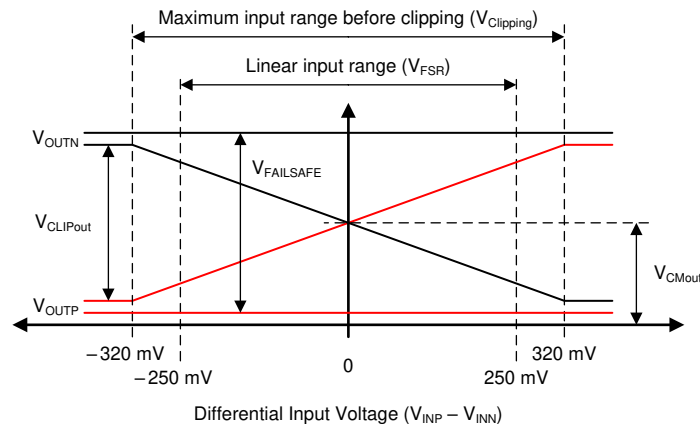


图 7-2. Output Behavior of the AMC1300B-Q1

The AMC1300B-Q1 offers a failsafe feature that simplifies diagnostics on system level. [Figure 7-2](#) shows the failsafe mode, in which the AMC1300B-Q1 outputs a negative differential output voltage that does not occur under normal operating conditions. The failsafe output is active in two cases:

- When the high-side supply is missing or below the $V_{DD1_{UV}}$ threshold
- When the common-mode input voltage, that is $V_{CM} = (V_{INP} + V_{INN}) / 2$, exceeds the common-mode overvoltage detection level V_{CMov}

Use the maximum $V_{FAILSAFE}$ voltage specified in the [Electrical Characteristics](#) table as a reference value for failsafe detection on system level.

7.4 Device Functional Modes

The AMC1300B-Q1 is operational when the power supplies V_{DD1} and V_{DD2} are applied, as specified in the [Recommended Operating Conditions](#) table.

8 Application and Implementation

备注

以下应用部分中的信息不属于 TI 器件规格的范围，TI 不担保其准确性和完整性。TI 的客户应负责确定器件是否适用于其应用。客户应验证并测试其设计，以确保系统功能。

8.1 Application Information

The low analog input voltage range, excellent accuracy, and low temperature drift make the AMC1300B-Q1 a high-performance solution for automotive applications where shunt-based current sensing in the presence of high common-mode voltage levels is required.

8.2 Typical Application

The AMC1300B-Q1 is ideally suited for shunt-based current sensing applications where accurate current monitoring is required in the presence of high common-mode voltages.

图 8-1 shows the AMC1300B-Q1 in a typical application. The load current flowing through an external shunt resistor RSHUNT produces a voltage drop that is sensed by the AMC1300B-Q1. The AMC1300B-Q1 digitizes the analog input signal on the high-side, transfers the data across the isolation barrier to the low-side, reconstructs the analog signal, and presents that signal as a differential voltage on the output pins.

The differential input, differential output, and the high common-mode transient immunity (CMTI) of the AMC1300B-Q1 ensure reliable and accurate operation even in high-noise environments.

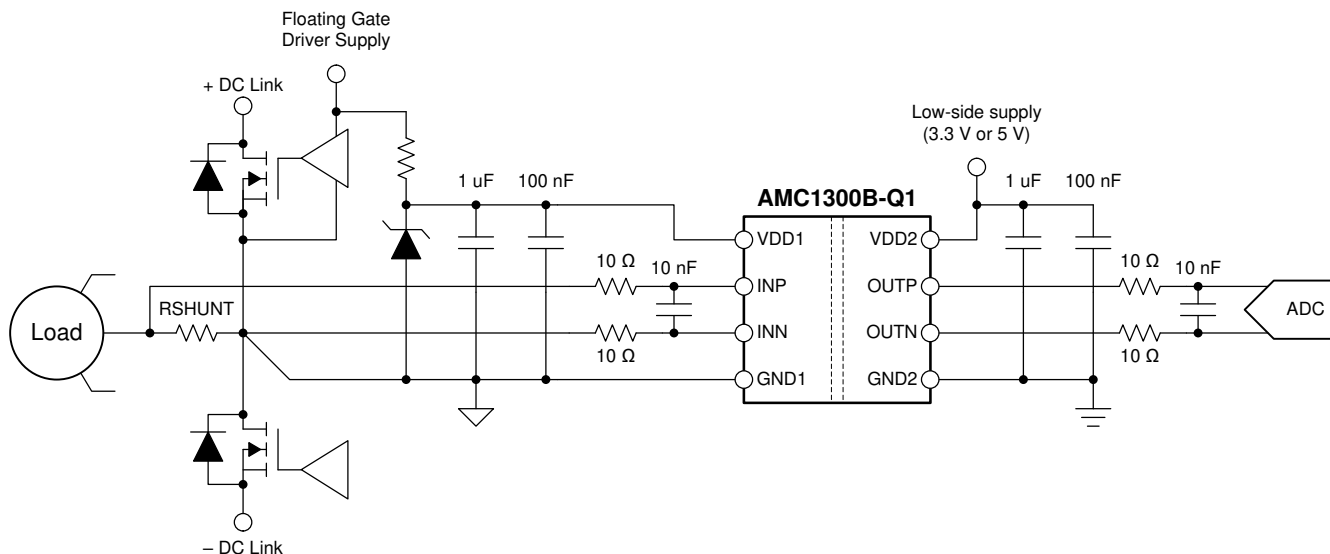


图 8-1. Using the AMC1300B-Q1 for Current Sensing in a Typical Application

8.2.1 Design Requirements

表 8-1 lists the parameters for this typical application.

表 8-1. Design Requirements

PARAMETER	VALUE
High-side supply voltage	3.3 V or 5 V
Low-side supply voltage	3.3 V or 5 V
Voltage drop across RSHUNT for a linear response	±250 mV (maximum)
Signal delay (50% V_{IN} to 90% OUTP, OUTN)	3 μ s (maximum)

8.2.2 Detailed Design Procedure

In 图 8-1, the high-side power supply (VDD1) for the AMC1300B-Q1 is derived from the floating power supply of the upper gate driver.

The floating ground reference (GND1) is derived from the end of the shunt resistor that is connected to the negative input of the AMC1300B-Q1 (INN). If a four-pin shunt is used, the inputs of the AMC1300B-Q1 are connected to the inner leads and GND1 is connected to the outer lead on the INN-side of the shunt. To minimize offset and improve accuracy, route the ground connection as a separate trace that connects directly to the shunt resistor rather than shorting GND1 to INN directly at the input to the device. See the [Layout](#) section for more details.

8.2.2.1 Shunt Resistor Sizing

Use Ohm's Law to calculate the voltage drop across the shunt resistor (V_{SHUNT}) for the desired measured current: $V_{SHUNT} = I \times R_{SHUNT}$.

Consider the following two restrictions when selecting the value of the shunt resistor, R_{SHUNT} :

- The voltage drop caused by the nominal current range must not exceed the recommended differential input voltage range for a linear response: $|V_{SHUNT}| \leq |V_{FSR}|$
- The voltage drop caused by the maximum allowed overcurrent must not exceed the input voltage that causes a clipping output: $|V_{SHUNT}| \leq |V_{Clipping}|$

8.2.2.2 Input Filter Design

TI recommends placing an RC-filter in front of the isolated amplifier to improve signal-to-noise performance of the signal path. Design the input filter such that:

- The cutoff frequency of the filter is at least one order of magnitude lower than the sampling frequency (20 MHz) of the $\Delta \Sigma$ modulator
- The input bias current does not generate significant voltage drop across the DC impedance of the input filter
- The impedances measured from the analog inputs are equal

For most applications, the structure shown in 图 8-2 achieves excellent performance.

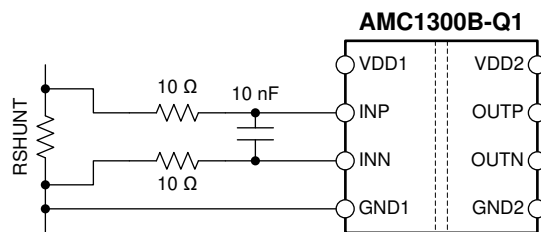


图 8-2. Differential Input Filter

8.2.2.3 Differential to Single-Ended Output Conversion

图 8-3 shows an example of a TLVx313-Q1-based signal conversion and filter circuit for systems using single-ended-input ADCs to convert the analog output voltage into digital. With $R1 = R2 = R3 = R4$, the output voltage equals $(V_{OUTP} - V_{OUTN}) + V_{REF}$. Tailor the bandwidth of this filter stage to the bandwidth requirement of the system. For most applications, $R1 = R2 = R3 = R4 = 3.3\text{ k}\Omega$ and $C1 = C2 = 330\text{ pF}$ yields good performance.

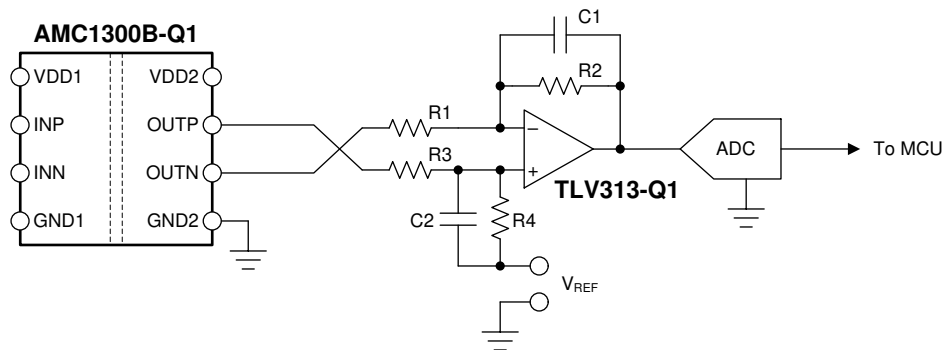


图 8-3. Connecting the AMC1300B-Q1 Output to a Single-Ended Input ADC

For more information on the general procedure to design the filtering and driving stages of SAR ADCs, see the [18-Bit, 1MSPS Data Acquisition Block \(DAQ\) Optimized for Lowest Distortion and Noise](#) and [18-Bit Data Acquisition Block \(DAQ\) Optimized for Lowest Power](#) reference guides, available for download at www.ti.com.

8.2.3 Application Curve

One important aspect of power-stage design is the effective detection of an overcurrent condition to protect the switching devices and passive components from damage. To power off the system quickly in the event of an overcurrent condition, a low delay caused by the isolated amplifier is required. 图 8-4 shows the typical full-scale step response of the AMC1300B-Q1.

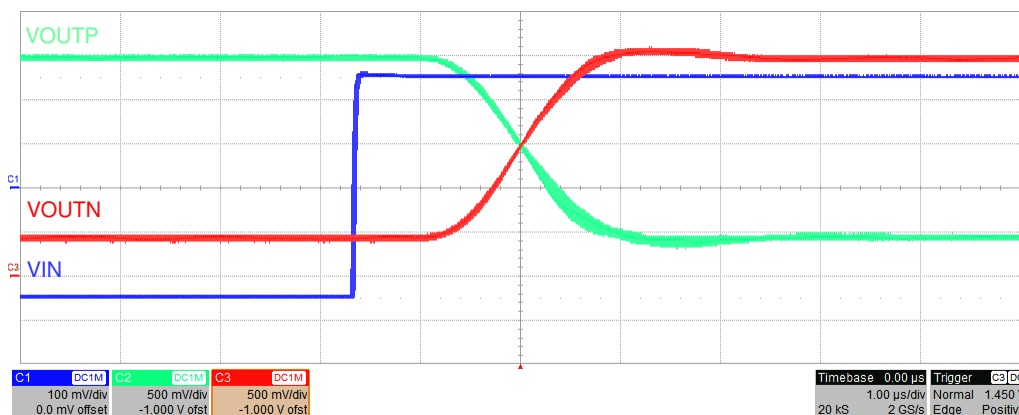


图 8-4. Step Response of the AMC1300B-Q1

8.3 What to Do and What Not to Do

Do not leave the inputs of the AMC1300B-Q1 unconnected (floating) when the device is powered up. If the device inputs are left floating, the input bias current may drive the inputs to a positive value that exceeds the operating common-mode input voltage and the device outputs the fail-safe voltage as described in the [Analog Output](#) section.

Connect the high-side ground (GND1) to INN, either by a hard short or through a resistive path. A DC current path between INN and GND1 is required to define the input common-mode voltage. Take care not to exceed the input common-mode range as specified in the [Recommended Operating Conditions](#) table. For best accuracy, route the ground connection as a separate trace that connects directly to the shunt resistor rather than shorting GND1 to INN directly at the input to the device. See the [Layout](#) section for more details.

9 Power Supply Recommendations

The AMC1300B-Q1 does not require any specific power up sequencing. The high-side power-supply (VDD1) is decoupled with a low-ESR 100-nF capacitor (C1) parallel to a low-ESR 1-μF capacitor (C2). The low-side power supply (VDD2) is equally decoupled with a low-ESR 100-nF capacitor (C3) parallel to a low-ESR 1-μF capacitor (C4). Place all four capacitors (C1, C2, C3, and C4) as close to the device as possible.

The ground reference for the high-side (GND1) is derived from the end of the shunt resistor, which is connected to the negative input (INN) of the device. For best DC accuracy, use a separate trace (as shown in [Figure 9-1](#)) to make this connection instead of shorting GND1 to INN directly at the device input. If a four-terminal shunt is used, the device inputs are connected to the inner leads and GND1 is connected to the outer lead on the INN-side of the shunt.

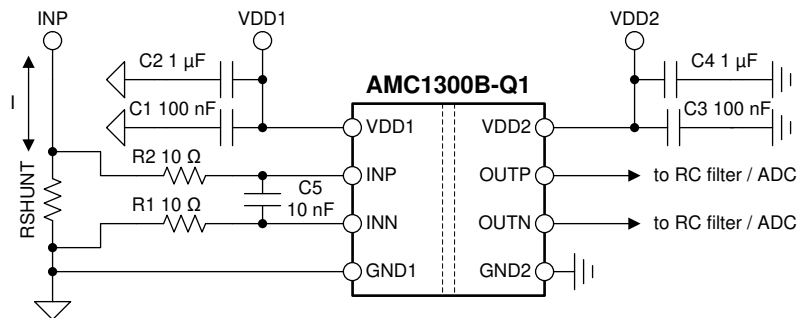


图 9-1. Decoupling of the AMC1300B-Q1

Capacitors must provide adequate effective capacitance under the applicable DC bias conditions they experience in the application. Multilayer ceramic capacitors (MLCCs) typically exhibit only a fraction of their nominal capacitance under real-world conditions and this factor must be taken into consideration when selecting these capacitors. This problem is especially acute in low-profile capacitors, in which the dielectric field strength is higher than in taller components. Reputable capacitor manufacturers provide capacitance versus DC bias curves that greatly simplify component selection.

10 Layout

10.1 Layout Guidelines

Figure 10-1 shows a layout recommendation with the critical placement of the decoupling capacitors (as close as possible to the AMC1300B-Q1 supply pins) and placement of the other components required by the device. For best performance, place the shunt resistor close to the INP and INN inputs of the AMC1300B-Q1 and keep the layout of both connections symmetrical.

10.2 Layout Example

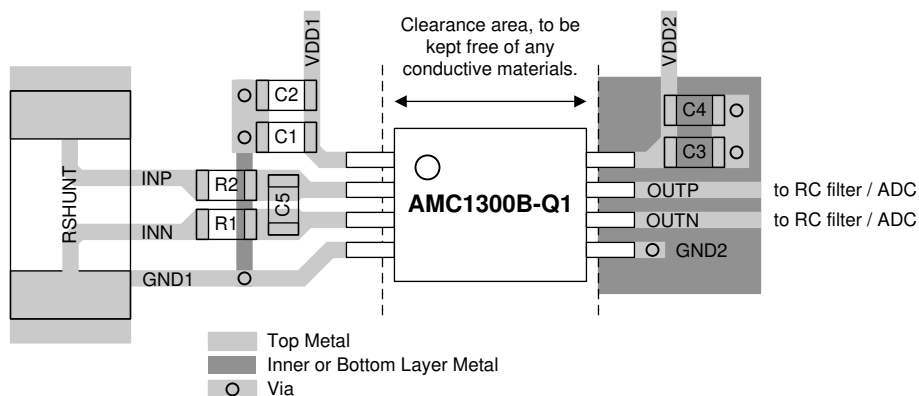


Figure 10-1. Recommended Layout of the AMC1300B-Q1

11 Device and Documentation Support

11.1 Documentation Support

11.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, [Isolation Glossary application report](#)
- Texas Instruments, [Semiconductor and IC Package Thermal Metrics application report](#)
- Texas Instruments, [ISO72x Digital Isolator Magnetic-Field Immunity application report](#)
- Texas Instruments, [TLVx313-Q1 Low-Power, Rail-to-Rail In/Out, 750- \$\mu\$ V Typical Offset, 1-MHz Operational Amplifier for Cost-Sensitive Systems data sheet](#)
- Texas Instruments, [18-Bit, 1-MSPS Data Acquisition Block \(DAQ\) Optimized for Lowest Distortion and Noise reference guide](#)
- Texas Instruments, [18-Bit, 1-MSPS Data Acquisition Block \(DAQ\) Optimized for Lowest Power reference guide](#)
- Texas Instruments, [Isolated Amplifier Voltage Sensing Excel Calculator design tool](#)

11.2 接收文档更新通知

要接收文档更新通知，请导航至 [ti.com](https://www.ti.com) 上的器件产品文件夹。点击 [订阅更新](#) 进行注册，即可每周接收产品信息更改摘要。有关更改的详细信息，请查看任何已修订文档中包含的修订历史记录。

11.3 支持资源

TI E2E™ 支持论坛 是工程师的重要参考资料，可直接从专家获得快速、经过验证的解答和设计帮助。搜索现有解答或提出自己的问题可获得所需的快速设计帮助。

链接的内容由各个贡献者“按原样”提供。这些内容并不构成 TI 技术规范，并且不一定反映 TI 的观点；请参阅 TI 的《[使用条款](#)》。

11.4 Trademarks

TI E2E™ is a trademark of Texas Instruments.

所有商标均为其各自所有者的财产。

11.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

11.6 术语表

TI 术语表 本术语表列出并解释了术语、首字母缩略词和定义。

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
AMC1300BQDWVRQ1	Active	Production	SOIC (DWV) 8	1000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	MC1300BQ
AMC1300BQDWVRQ1.A	Active	Production	SOIC (DWV) 8	1000 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 125	MC1300BQ

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

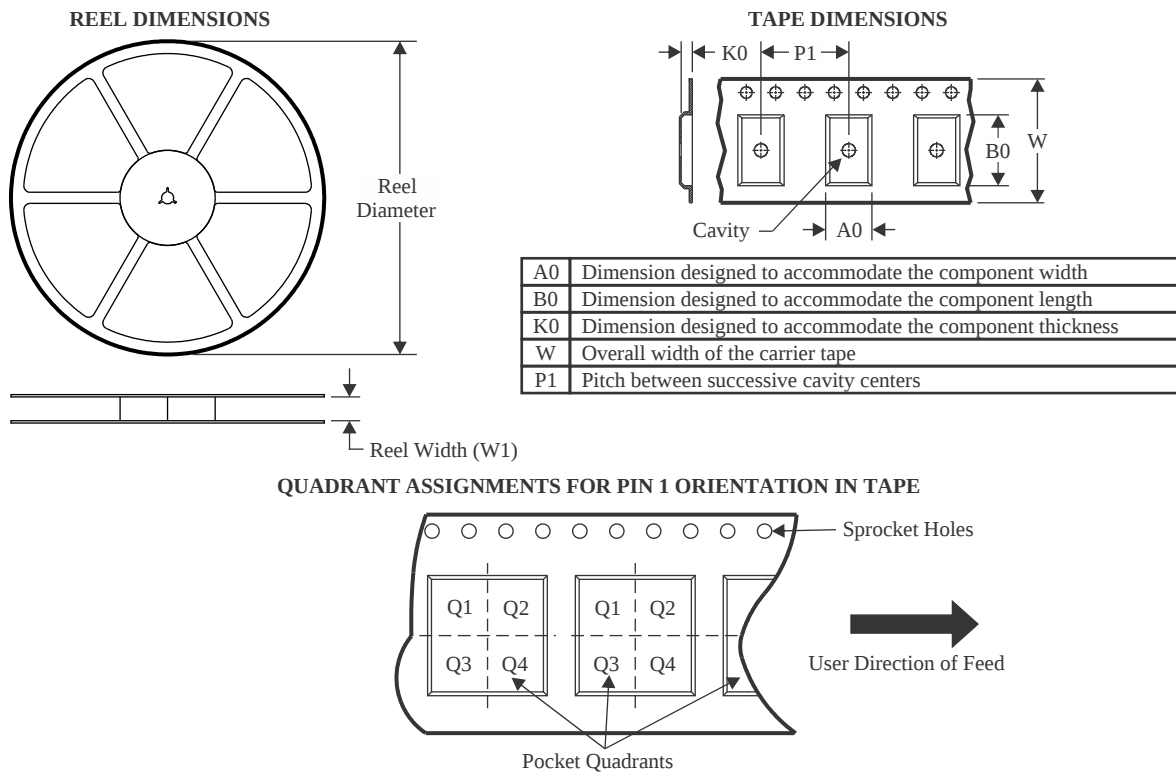
⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

Important Information and Disclaimer:The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

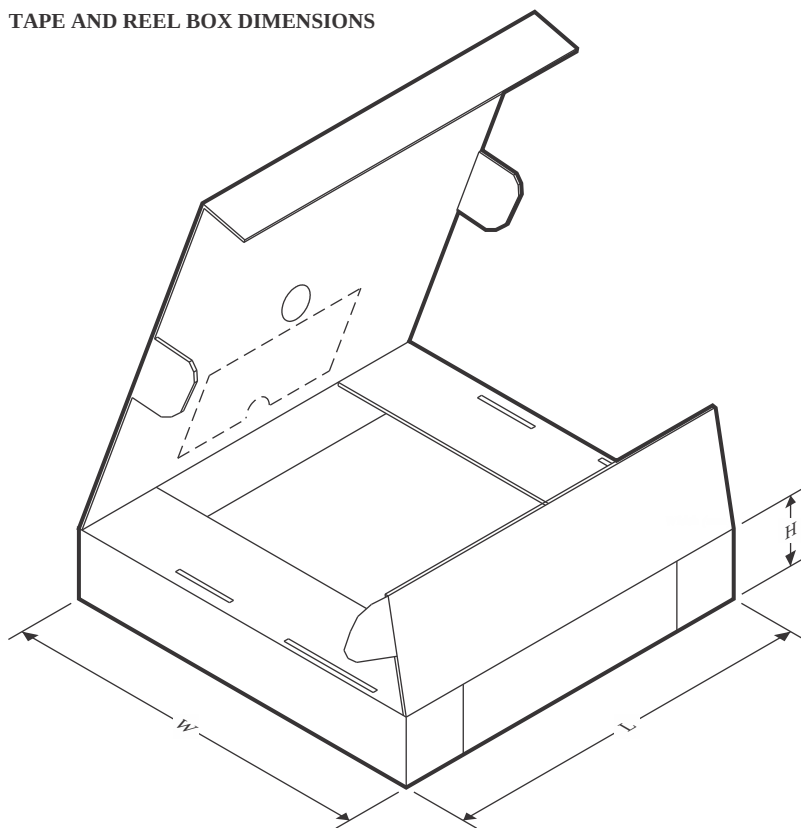
TAPE AND REEL INFORMATION



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
AMC1300BQDWVRQ1	SOIC	DWV	8	1000	330.0	16.4	12.15	6.2	3.05	16.0	16.0	Q1
AMC1300BQDWVRQ1	SOIC	DWV	8	1000	330.0	16.4	12.05	6.15	3.3	16.0	16.0	Q1

TAPE AND REEL BOX DIMENSIONS

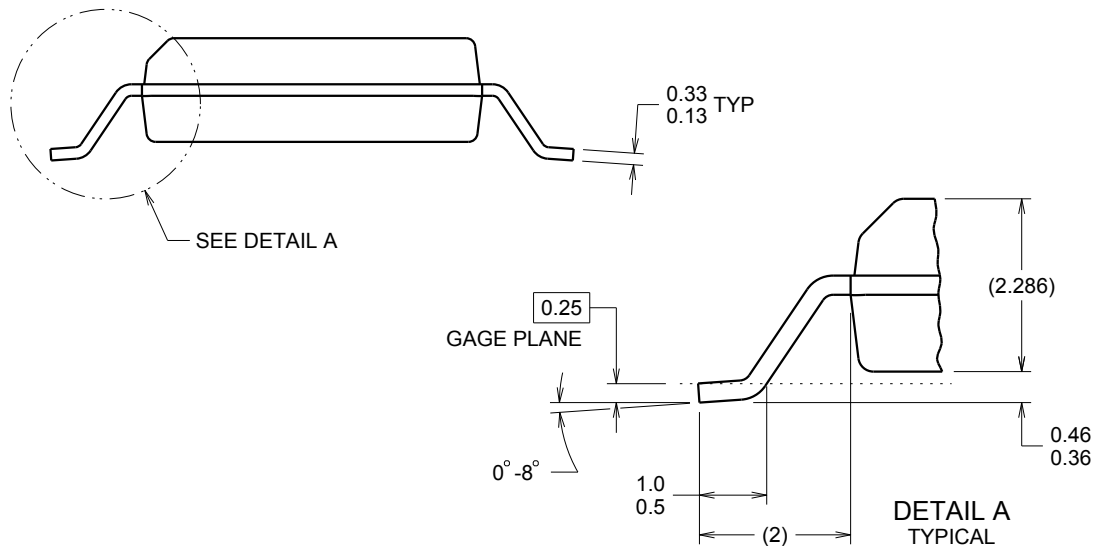


*All dimensions are nominal

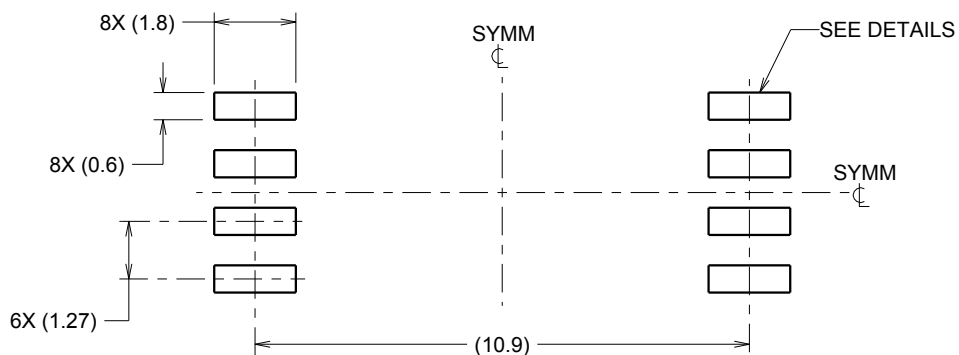
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
AMC1300BQDWVRQ1	SOIC	DWV	8	1000	353.0	353.0	32.0
AMC1300BQDWVRQ1	SOIC	DWV	8	1000	350.0	350.0	43.0

SOIC - 2.8 mm max height

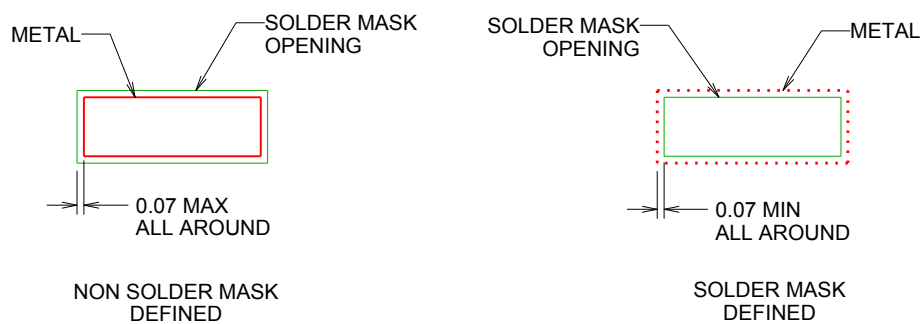
PIN 1 ID AREA
 11.5 ± 0.25 TYP
 5.95
 5.75
 NOTE 3
 1
 4
 8
 5
 6X 1.27
 2X 3.81
 8X 0.51 0.31
 7.6
 7.4
 NOTE 4
 A
 B
 SEATING PLANE
 0.1 C
 2.8 MAX
 C
 A
 B



1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.



LAND PATTERN EXAMPLE
9.1 mm NOMINAL CLEARANCE/CREEPAGE
SCALE:6X

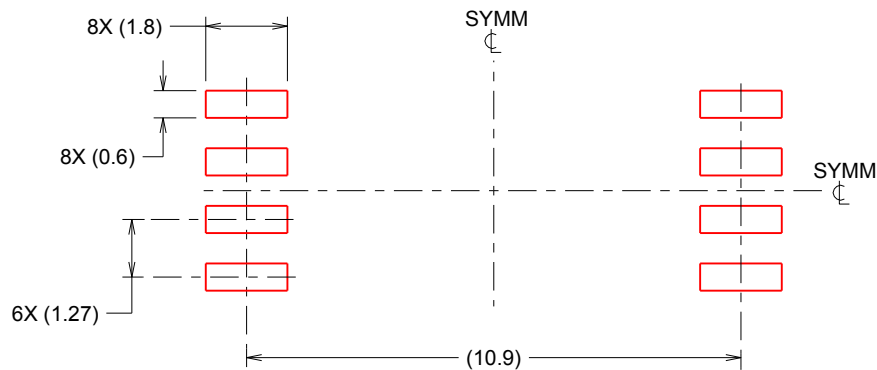


SOLDER MASK DETAILS

4218796/A 09/2013

NOTES: (continued)

5. Publication IPC-7351 may have alternate designs.
6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE:6X

4218796/A 09/2013

NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
8. Board assembly site may have different recommendations for stencil design.

重要通知和免责声明

TI“按原样”提供技术和可靠性数据（包括数据表）、设计资源（包括参考设计）、应用或其他设计建议、网络工具、安全信息和其他资源，不保证没有瑕疵且不做任何明示或暗示的担保，包括但不限于对适销性、与某特定用途的适用性或不侵犯任何第三方知识产权的暗示担保。

这些资源可供使用 TI 产品进行设计的熟练开发人员使用。您将自行承担以下全部责任：(1) 针对您的应用选择合适的 TI 产品，(2) 设计、验证并测试您的应用，(3) 确保您的应用满足相应标准以及任何其他安全、安保法规或其他要求。

这些资源如有变更，恕不另行通知。TI 授权您仅可将这些资源用于研发本资源所述的 TI 产品的相关应用。严禁以其他方式对这些资源进行复制或展示。您无权使用任何其他 TI 知识产权或任何第三方知识产权。对于因您对这些资源的使用而对 TI 及其代表造成的任何索赔、损害、成本、损失和债务，您将全额赔偿，TI 对此概不负责。

TI 提供的产品受 [TI 销售条款](#)、[TI 通用质量指南](#) 或 [ti.com](#) 上其他适用条款或 TI 产品随附的其他适用条款的约束。TI 提供这些资源并不会扩展或以其他方式更改 TI 针对 TI 产品发布的适用的担保或担保免责声明。除非德州仪器 (TI) 明确将某产品指定为定制产品或客户特定产品，否则其产品均为按确定价格收入目录的标准通用器件。

TI 反对并拒绝您可能提出的任何其他或不同的条款。

版权所有 © 2026，德州仪器 (TI) 公司

最后更新日期：2025 年 10 月