

Support & training

[LM3645](https://www.ti.com/product/LM3645) [SNVSCV4](https://www.ti.com/lit/pdf/SNVSCV4) – SEPTEMBER 2024

LM3645 Synchronous Boost Quad LED Flash Driver with 2-A High-Side Current Sources

1 Features

- Quad independent led current source programmability
- Accurate and programmable LED current range from 0.525 mA to 2 A
- Optimized flash LED current during low battery conditions (I_{VFM})
- > 85% efficiency in torch mode (at 100 mA) and flash mode (at 1 A to 2 A)
- Grounded cathode LED operation for improved thermal management
- Dual hardware strobe enable (STR1, STR2)
- Synchronization input for RF power amplifier pulse events (TORCH/TX)
- Hardware torch enable (TORCH/Tx)
- Dual remote NTC monitoring (TEMP1, TEMP2)
- Fixed output voltage mode
- 1-MHz I²C-compatible interface
	- $-$ (1^2 C Address = 0x65)

2 Applications

- Camera phone white LED flash
- Digital still cameras
- Barcode scanners
- Handheld data terminals

3 Description

The LM3645 is a quad LED flash driver that provides a high level of adjustability within a small solution size.

The LM3645 utilizes a synchronous boost converter to provide power to the quad 2A constant current LED sources. The quad 256 level current sources provide the flexibility to adjust the current ratios between LED1, LED2, LED3 and LED4 with each driver capable of delivering a maximum of 2A. An adaptive regulation method ensures the current sources remain in regulation and maximizes efficiency.

Features of the LM3645 are controlled via an 1^2C compatible interface. These features include: dual hardware flash pins and a hardware torch or Tx interrupt pin (STR1, STR2 and TORCH/TX), and dual NTC thermistor monitors. The device offers independently programmable currents in each output leg to drive the LEDs in a flash or movie mode (Torch) condition as well as a fast turn-on and turn-off IR Mode.

Two switching frequency options, overvoltage protection (OVP), and adjustable current limit allow for the use of tiny, low-profile inductors and 10µF ceramic capacitors. The device operates over a –40°C to +85°C ambient temperature range.

Device Information (1)

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Simplified Application

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4 Pin Configuration and Functions

Figure 4-1. YCG Package, 25-Pin DSBGA

Table 4-1. Pin Functions

5 Specifications

5.1 Absolute Maximum Ratings

overoperating free-air temperature range (unless otherwise noted) (1) (2)

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. Theseare stress ratings only, which do not imply functional operation of the device at these or anyother conditions beyond those indicated under *Recommended OperatingConditions*. Exposure to absolute-maximum-rated conditions for extended periods mayaffect device reliability.

- (2) All voltages are with respect to the potential at the GND pin.
- (3) Internal thermal shutdown circuitry protects the device frompermanent damage. Thermal shutdown engages at T $_{\rm J}$ = 150°C (typical). Readback of the thermal shutdown bit resets the detection mechanism. Thermal shutdown is ensured bydesign.

5.2 ESD Ratings

(1) JEDEC document JEP155 states that 500-V HBM allows safemanufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safemanufacturing with a standard ESD control process.

5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted) (1) (2)

(1) Stresses beyond those listed under *absolute maximum ratings* may cause permanent damage to the device. Theseare stress ratings only, and functional operation of the device at these or any other conditionsbeyond those indicated under *recommended operating conditions* isnot implied. Exposure to absolute-maximum-rated conditions for extended periods may affect devicereliability. All voltages are with respect to the potential at the GNDterminal.

(3) In applications where high power dissipation and/or poor packagethermal resistance is present, the maximum ambient temperature may have to be derated. Maximumambient temperature $(T_{A\text{-MAX}})$ is dependent on the maximum operating junctiontemperature $(T_{J-MAX-OP} = 125^{\circ}C)$, the maximum power dissipation of the devicein the application (P_{D-MAX}) , and the junction-to-ambient thermal resistanceof the part/package in the application $(R_{\theta JA})$, as given by the followingequation: $T_{A\text{-MAX}} = T_{J\text{-MAX-D}} - (R_{\theta JA} \times P_{D\text{-MAX}})$.

5.4 Thermal Information

(1) For more information about traditional and new thermalmetrics, see the *[Semiconductor and ICPackage Thermal Metrics](https://www.ti.com/lit/pdf/SPRA953)* application report.

5.5 Electrical Characteristics

Typical limits tested at $T_A = 25^{\circ}$ C. Minimum and maximumlimits apply over the full operating ambient temperature range $(-40^{\circ}$ C ≤ T_A ≤+85°C). Unless otherwise specified, V_{IN} = 3.6 V, EN = 1.8 V,V_{IO} = 1.8 V.^{(1) (2)}

(1) Minimum (MIN) and Maximum (MAX) limits are specified by design,test, or statistical analysis. Typical (TYP) numbers are not verified, but do represent the mostlikely norm. Unless otherwise specified, conditions for typical specifications are: V_{IN} = 3.6 V and T_A = 25°C. (2) All voltages are with respect to the potential at the GND pin.

(3) Matching is defined as the greater of the (ILEDMAX - ILEDAVERAGE)/ILEDAVERAGE OF |(ILEDMIN-ILEDAVERAGE)|/ILEDAVERAGE

5.6 Timing Requirements

5.7 Switching Characteristics

over operating free-air temperature range (unless otherwise noted)

Figure 5-1. I ²C-Compatible Interface Specifications

5.8 Typical Characteristics

 I_{CL} = 5 A

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6 Detailed Description

6.1 Overview

The LM3645 is a high-power white LED flash driver capable of delivering up to 2 A in any of the four parallel LEDs. The device incorporates a 2-MHz or 4-MHz constant frequency-synchronous current-mode PWM boost converter and quad high-side current sources to regulate the LED current over the 2.3-V to 5.5-V input voltage range.

The LM3645 PWM DC/DC boost converter switches and boosts the output to maintain at least V_{HR} across each of the current sources (LED1/2/3/4). This minimum headroom voltage ensures that the current sources remain in regulation. If the input voltage is above the LED voltage + current source headroom voltage the device does not switch, but turns the P-channel MOSFET on continuously (Pass mode). In Pass mode the difference between $(V_{\text{IN}} - I_{\text{IFD}} \times R_{\text{PMOS}})$ and the voltage across the LED is dropped across the current source.

The LM3645 has logic inputs including two hardware Flash Enables (STROBE1 and STROBE2), and a hardware Torch Enable muxed with a Flash Interrupt input (TX) designed to interrupt the flash pulse during high battery-current conditions (TORCH/Tx, Tx = default). These logic inputs have internal 300-kΩ (typical) pulldown resistors to GND.

Additional features of the LM3645 include internal comparators for LED thermal sensing via external NTC thermistors (TEMP1, TEMP2) and an input voltage monitor that can reduce the Flash current during low V_{IN} conditions. It also has a Hardware Enable (EN) pin that can be used to reset the state of the device and the registers by pulling the EN pin to ground.

Control is done via an $1²C$ -compatible interface. This includes adjustment of the Flash and Torch current levels, changing the Flash Timeout Duration, and changing the switch current limit. Additionally, there are flag and status bits that indicate flash current time-out, LED over-temperature condition, LED failure (open/short), device thermal shutdown, TX interrupt, and V_{IN} undervoltage conditions.

6.2 Functional Block Diagram

6.3 Feature Description

6.3.1 Power Amplifier Synchronization (TORCH/TX)

The TORCH/TX pin is either a Power Amplifier Synchronization input, or external torch enable. In Tx mode, this function is designed to reduce the flash LED current and thus limit the battery current during high battery current conditions such as PA transmit events. When the LM3645 is engaged in a Flash event, and the TX pin is pulled high, the LED current is forced into Torch Mode at the programmed Torch current setting. If the TX pin is then pulled low before the Flash pulse terminates, the LED current returns to the previous Flash current level. At the end of the Flash time-out, whether the TX pin is high or low, the LED current turns off. If a Tx event occurs during a flash, a flag will be set in the FLAG_RPT_REG register (bit[5] in reg 0x14).

6.3.2 Input Voltage Flash Monitor (IVFM)

The LM3645 has the ability to adjust the flash current based upon the voltage level present at the IN pin utilizing the Input Voltage Flash Monitor (IVFM). The adjustable threshold IVFM-D (set in register IVFM_THRES) ranges from 2.5 V to 3.25 V in 50-mV steps, with three different usage modes set in the IVFM_THRES register (Stop

and Hold, Adjust Down Only, Adjust Up/Down). The Voltage Fault Register has the IVFM flag (bit[4] reg 0x15) bit set when the input voltage crosses the IVFM-D value. Additionally, the IVFM-D threshold sets the input voltage boundary that forces the LM3645 to either stop ramping the flash current during start-up (Stop and Hold Mode) or to start decreasing the LED current during the flash (Down Adjust Only and Up/Down Adjust). In Adjust Up/Down mode, the IVFM-D value plus the hysteresis voltage threshold set the input voltage boundary that forces the LM3645 to start ramping the flash current back up towards the target.

Figure 6-1. IVFM Modes

6.3.3 Fault/Protections

6.3.3.1 Fault Operation

If the LM3645 enters a fault condition, the device sets the appropriate flag in the Volt Fault Reg and Therm Fault Reg Registers (0x15 and 0x16), and places the device into standby by clearing the Dx EN Bits ([3:0]) in CTRL_REG2 (0x01). The LM3645 remains in standby until an I²C read of the fault registers is completed. Upon clearing the flags/faults, the device can be restarted (Flash, Torch, IR, etc.). If the fault is still present, the LM3645 re-enters the fault state and enters standby again. When a fault flag bit is present in either of the fault registers, the D_MODE register (0x02) becomes read only.

Note

It is recommended that the fault registers be read either before a device enable or after a device disable. Failure to do so can cause inadvertent enabling of the device upon a read of the fault registers if an enable command has been issued prior to clearing any fault flags.

6.3.3.2 Flash Time-Out

The Flash Time-Out period sets the amount of time that the Flash Current is being sourced to the current sources (LED1/2/3/4). The LM3645 has 16 timeout levels (bits [3:0] in reg 0x0A) ranging from 10 ms to 400 ms or from 40 ms to 1.6 s depending upon the time-out multiplier bit ([4] in reg 0x0A). Additionally, there is an option to disable the time-out all together however care must be taken to ensure that the LEDs are not held in a high current condition longer than they are rated for, and care must be taken to ensure LM3645 does not hit the thermal scaleback or thermal shutdown faults. Additionally, changing the time out duration while in a flash event is not recommended as doing so will force the flash duration to be extended.

6.3.3.3 Overvoltage Protection (OVP)

The output voltage is limited to typically 5.7 V. In situations such as an open LED, the LM3645 raises the output voltage in order to try and keep the LED current at its target value. When V_{OUT} reaches 5.7 V (typical) the overvoltage comparator trips and turns off the internal N-channel MOSFET. When V_{OUT} falls below the "V_{OVP} **Off Threshold**", the LM3645 begins switching again. The Dx_EN bits are cleared, and the OVP flag is set (bit[6] in reg 0x15), when an OVP condition is present for three rising OVP edges. This prevents momentary OVP events from forcing the device to shut down.

When operating in voltage mode, the OVP protection mechanism is enabled, however the fault reporting and shut down functions are not enabled.

6.3.3.4 Current Limit

The LM3645 features four selectable inductor current limits that are programmable through the I²C-compatible interface. When the inductor current limit is reached, the LM3645 terminates the charging phase of the switching cycle. Switching resumes at the start of the next switching period. If the overcurrent condition persists, the device operates continuously in current limit.

Since the current limit is sensed in the NMOS switch, there is no mechanism to limit the current when the device operates in Pass Mode (current does not flow through the NMOS in pass mode). In Boost mode or Pass mode if V_{OUT} falls below 2.3 V, the device stops switching, and the P-channel MOSFET operates as a current source limiting the current to 200 mA. This prevents damage to the LM3645 and excessive current draw from the battery during output short-circuit conditions. The output enable bits are not cleared upon a Current Limit event, but a flag is set (bit [4] in reg 0x14).

The over current protection (OCP) circuit compares the inductor current to the input current limit (I_{CL}) setting after a fixed blanking time of t_{DLY} . The blanking time ensures that the OCP signal does not false fire as the low-side NMOS of the boost converter turns on. Use [Equation 1](#page-13-0) to estimate the maximum recommended total LED current to ensure proper current limit level operation accounting for the delay time.

$$
I_{\text{LOAD}} \le \left(I_{\text{CL}} + \frac{V_{\text{IN}}}{2 \times L} \times \left(\frac{D}{f_{\text{SW}}} - 2 \times t_{\text{DLY}} \right) \right) \times (1 - D) \times \eta
$$

(1)

where

- D is the duty cycle
- η is the boost efficiency
- I_{CL} is the current limit setting
- L is the inductance value
- f_{SW} is the boost switching frequency

Figure 6-2. Current Limit Timing

$$
Slew = \frac{\Delta I_{P-P}}{D \times T}
$$

(2)

6.3.3.5 NTC Thermistor Input/Outputs (TEMP1, TEMP2)

The TEMP1 and TEMP2 pins serve as a threshold detector and bias source for negative temperature coefficient (NTC) thermistors. When the voltage at TEMPx goes below the programmed threshold, the LM3645 device transitions into one of these three modes:

- Force standby mode
- Force torch mode
- Report only mode (bits [5:2] in reg 0x0D)

The device allows adjustment of the NTC threshold voltage from 250 mV to 1 V in 50 mV steps (bits [7:0] in reg 0x0F). Additionally, the NTC1 and NTC2 voltage levels provide an indicator of the remotely sensed temperature. The NTC bias current is adjustable in 25 µA steps from 25 µA to 100 µA (bits [7:6] in reg 0x0D). The NTC1

and NTC2 detection circuitry can be enabled or disabled via the NTC_MODE Register (bits [1:0] in reg 0x0D). If enabled, the NTC block turns on and off during the start and stop of a Flash/Torch/IR event.

Additionally, the NTC input looks for an open NTC connection and a shorted NTC connection. If the NTC input falls below 100 mV, the NTC short flag is set, and the device is disabled (bits [7:6] in reg 0x16). If the NTC input rises above 2.0 V, the NTC Open flag is set, and the device is disabled (bits [5:4] in reg 0x16). These fault detections can be individually disabled or enabled via the NTC Open Fault Enable bit (bits [2] in reg 0x13) and the NTC Short Fault Enable bit (bit [0] in reg 0x13).

Figure 6-3. Temp Detection Diagram

To further extend the functionality of the NTC inputs, the voltage on both TEMP1 and TEMP2 can be read back to provide real time feedback regarding the temperature at the point of detection (bits [7:0] in reg 0x10). When either TEMP1 or TEMP2 pin is enabled, a 4-bit ADC is enabled and continually updates the NTC Voltage Register. The voltage on both TEMP1 and TEMP2 are measured if either NTC detection block is enabled. In this case, the data read into the NTC voltage register for the disabled TEMP detection block can be random and can be ignored. Only enabled TEMP blocks yield accurate readings.

Each NTC block can be associated with any of the four outputs. **Each Dx Output should only be assigned to one of the NTC detection blocks.**

Figure 6-4. NTC Detection Diagram

6.3.3.6 Thermal Scale Back

In the event that the LM3645 junction temperature reaches 125°C, the device will adjust the LED current in one of two different ways; turn off the current completely or force the current to a torch level (bits [5:4] in reg 0x13). The thermal scale back feature can be enabled or disabled as well (bit [3] in reg 0x13). The torch scale back mode is valid for flash mode and IR mode only. If the torch scaleback mode is used in torch mode, no action will occur. The force standby mode option is valid in all modes.

Figure 6-5. TSB Adjust Modes

6.3.3.7 Thermal Shutdown (TSD)

When the LM3645 die temperature reaches 150°C, the thermal shutdown detection circuit trips, forcing the LM3645 into standby and writing a '1' to the corresponding bit of the Thermal Fault Report Register (bit [0] in reg 0x16). The LM3645 is only allowed to restart after the Thermal Fault Report Register (0x17) is read, clearing the fault flag. Upon restart, if the die temperature is still above 150°C, the LM3645 resets the Fault flag and re-enters standby.

6.3.3.8 Undervoltage Lockout (UVLO)

The LM3645 has an internal comparator that monitors the voltage at IN and forces the LM3645 into standby if the input voltage drops to 2.2 V. If the UVLO monitor threshold is tripped, the UVLO flag bit is set in the VOLT_FAULT Register (bit [7] in reg 0x15). If the input voltage rises above 2.2 V, LM3645 is not available for operation until there is an I²C read of the VOLT_FAULT_REG Register (0x15). Upon a read, the Flags1 register is cleared, and normal operation can resume if the input voltage is greater than 2.3 V.

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6.3.3.9 LED and/or VOUT Short Fault

The LED Fault flags read back a '1' if the device is active in Flash, IR or Torch mode and either active LED output experiences a short condition. The Output Short Fault flags (bits [3:0] in reg 0x15) reads back a '1' if the device is active in Flash, Torch or IR mode and the boost output experiences a short condition. An LED short condition is determined if the voltage at LED1, LED2, LED3 or LED4 goes below 500 mV (typ.) while the device is in Torch or Flash mode. Only the output detecting a short will be shutdown. There is a deglitch time of 256 μs before the LED Short flag is valid and a deglitch time of 2.048 ms before the VOUT Short flag (bit[6] in reg 0x15) is valid. The LED Short Faults can be reset to '0' by removing power to the input pin , setting EN to '0', setting the SW RESET bit to a '1', or by reading back the Voltage Fault Register (0x15). The Dx_EN bits are cleared upon an LED and/or V_{OUT} short fault.

6.3.3.10 Fault Behavior Table

Each fault has a specific interaction behavior with the output enable control (Dx_EN bits). Some faults are global in nature while others are specific to a certain output or function.

Table 6-1. Fault Behavior Table

6.4 Device Functioning Modes

6.4.1 Flash Mode

In Flash Mode, the LED current sources (LED1/2/3/4) provide 256 target current levels from 7.325 mA to 2 A. Once the Flash sequence is activated the current source (LED) ramps up to the programmed Flash current by stepping through all current steps until the programmed current is reached. The headroom in the four current sources can be regulated to provide 7.325 mA to 2 A on each of the four output legs.

When the device is enabled in Flash Mode through the DMODE Register (reg 0x02) and Dx_EN bits in Control Register 2 (bits [3:0] in reg 0x01), all Dx_EN bits cleared after a flash time-out event. If a flash event is initiated

by a strobe pin, the Dx_EN bits do not clear at the end of the pulse unless a fault or timeout occurs allowing multiple flash pulses to occur without further I2C interaction.

6.4.2 Torch Mode

In Torch mode, the LED current sources (LED1/2/3/4) provide 256 target current levels from 0.525 mA to 360 mA. The Torch currents are adjusted via the Dx Torch Brightness Registers. Torch mode is activated by the Enable Register (setting Dx_MODE, bits to '10'), or by pulling the TORCH/TX pin HIGH when the pin is enabled (Control Register2) and set to Torch Mode. Once the TORCH sequence is activated the active current sources (LED1/2/3/4) ramps up to the programmed Torch current by stepping through all current steps until the programmed current is reached. The rate at which the current ramps is determined by the value chosen in the Torch_Ramp register.

Torch Mode is not affected by Flash Timeout or by a TX Interrupt event.

6.4.3 IR Mode

In IR Mode, the target LED current is equal to the value stored in the LED1/2/3/4 Flash Brightness Registers. When IR mode is enabled on any of the outputs, the boost converter turns on and set the output equal to the input (pass-mode) or to the selected voltage mode depending upon the state of the voltage mode settings. At this point, toggling the STROBE pins enables and disables the LED1/2/3/4 current sources (if enabled via the Dx Mode bits). The strobe pins can only be set to be Level sensitive, meaning all timing of the IR pulse is externally controlled. If the strobe pin is set to edge mode, the LM3645 device times out after the end of the set duration. In IR Mode, the current sources either ramp quickly or do not ramp the LED outputs to the target based upon the ramp time setting chosen.

Figure 6-6. IR Mode with Boost

Figure 6-8. IR Mode Voltage Mode Enabled

6.4.4 Voltage Mode

LM3645 has the ability to turn the DC/DC boost converter typically used to power the four LED outputs into a general purpose fixed output power rail (bit [0] in reg 0x00) for use in other system applications that need voltage conversion. Four user selectable regulation voltage levels can be generated; 4.0 V, 4.5 V, 5.0V and 5.2 V (bits [7:6] in register 0x00). Voltage Mode can be used in conjunction with driving LEDs and will behave based upon the Output Mode Priority Bit (bit [1] in reg 0x00). If the bit is set to '0', maintaining voltage regulation is of highest importance. In Voltage Mode priority mode, care must be taken to manage the on chip thermal dissipation. If the

bit is set to '1', the boost is allowed to regulate to a level that is optimized for driving the enabled LEDs. In LED drive priority mode, the output voltage returns to the target voltage mode level at the termination of LED driving.

In voltage mode, there are current limitations for each mode due to the peak switch node voltage. Using a 1-Ω resistor and 4-nF capacitance for the snubber, the maximum allowed currents are as follows:

- \cdot 4.0 V mode = 3 A peak current
- 4.5 V mode = 1.5 A peak current
- 5.0 V mode = 750 mA peak current
- 5.2 V mode = Not recommended

6.4.5 Mode Transitions

The LM3645 can immediately transition between modes, it is best practice to transition from mode to mode by first entering standby. Implementing a standby period minimizes input current and output voltage transients, and provides an overall resetting of all timers and state machines. Additionally, all external control pins should be at '0' during the standby state to ensure proper timing and synchronization.

6.4.6 Boost Operation

6.4.6.1 Start-Up (Enabling The Device)

Turn on of the LM3645 modes can be done through a combination of Control Register1, Control Register2, D_Mode Register and Strobe Control Register depending upon which mode is enabled. On start-up, when V_{OUT} is less than V_{IN} the internal synchronous P-channel MOSFET turns on as a current source and delivers 200 mA (typical) to the output capacitor. During this time the current source (LED) is off. When the voltage across the output capacitor reaches 2.2 V (typical) the current source turns on. At turnon the current source steps through each FLASH or TORCH level until the target LED current is reached unless voltage mode is enabled with the LED outputs disabled. This gives the device a controlled turnon and limits inrush current from the V_{IN} supply.

6.4.6.2 Pass Mode

The LM3645 starts up in Pass Mode and stays there until Boost Mode is needed to maintain regulation. If the voltage difference between V_{OUT} and V_{LED} falls below V_{HR} , the device switches to Boost Mode. In Pass Mode the boost converter does not switch, and the synchronous P-channel MOSFET turns fully on bringing V_{OUT} up to V_{IN} – I_{LED} × R_{PMOS}. In Pass Mode the inductor current is not limited by the peak current limit.

6.4.6.3 Output Voltage Regulation

In LED drive mode, the LM3645 boost will either operate in pass mode if the output voltage minus the LED voltage is greater than the target current source headroom voltage. If the output can no longer operate in pass mode, the DC/DC boost will turn on and constantly monitor all enabled outputs and ensure that the output with the highest forward voltage LED always has the correct headroom voltage.

In Voltage drive mode, the output is always regulated to the target value so long as the output voltage is less than the input voltage. If the input voltage is higher than the output voltage, the device will not be regulated and will instead operate in pass mode.

Table 6-2. Output Voltage Assignment Table

Figure 6-9. Output Voltage and Mode Diagram

6.5 Programming and Control

6.5.1 Dx_EN Bits

In order for current to flow in the LEDs, the appropriate Dx_EN bits in Configuration Register 2 must be set to logic '1' (bits [3:0) in reg 0x01). When enabled, the Dx outputs will behave based upon the mode that pin is set to (DMODE register). If the Dx_EN bit is set to a '0', the output will remain off regardless of the mode the output is placed in. At the end of an I2C controlled flash event when a time-out occurs, the Dx_EN bits will auto-clear. If the mode is changed during an I2C controlled flash, the Dx_EN bit will not autoclear. When used along with the strobe pins, the Dx_EN pins do not automatically reset unless a fault event occurs.

6.5.2 STR1 and STR2 Usage

Two external synchronization Strobe pins are available to enable the four LED outputs in either Flash or IR mode. Each of the four outputs can be assigned to either Strobe pin via the Strobe Control Register (bits [7:0] in reg 0x03). Each output assigned to a strobe pin must be set to the same LED mode, either all Flash or all IR. **Additionally, no output should be assigned to both strobe pins.** If D1 is assigned to STR1, it should not be assigned to STR2 as well.

If the LM3645 is already enabled, the minimum strobe duration in either level or edge mode must be 10 µs. This restriction applies to both flash and IR mode operation. If the LM3645 is in standby and either strobe pin is asserted, it is recommended that the strobe duration be no less than 1ms to ensure that the boost has enough time to reach the proper level to ensure current into the LED can regulate.

Note

It is recommended that the STR1 and STR2 pins be held low before enabling the outputs assigned to each strobe pin.

Table 6-3. Strobe Control Truth Table

6.5.3 TOR/TX Usage

When the TOR/TX pin is configured for use as a hardware controlled torch enable (bits [7:6] in reg 0x01), all outputs assigned to Torch Mode in the Diode Mode Register (bits [7:0] in reg 0x02) will be enabled or disabled via the external pin.

Note

If configured to the Torch configuration, it is recommended that the TOR/TX pin be held low before enabling the outputs assigned to Torch Mode.

Table 6-4. Torch Control Truth Table

6.5.4 Control State Diagram

Figure 6-10. Mode Control Diagram

6.5.5 I ²C-Compatible Interface

6.5.5.1 Data Validity

The data on SDA must be stable during the HIGH period of the clock signal (SCL). In other words, the state of the data line can only be changed when SCL is LOW.

Figure 6-11. Data Validity Data

A pullup resistor between the controller's VIO line and SDA must be greater than [**(VIO - VOL) / 3 mA**] to meet the V_{OL} requirement on SDA. Using a larger pullup resistor results in lower switching current with slower edges, while using a smaller pullup results in higher switching currents with faster edges.

6.5.5.2 Start and Stop Conditions

START and STOP conditions classify the beginning and the end of the I²C session. A START condition is defined as the SDA signal transitioning from HIGH to LOW while SCL line is HIGH. A STOP condition is defined as the SDA transitioning from LOW to HIGH while SCL is HIGH. The I²C master always generates START and STOP conditions. The I²C bus is considered busy after a START condition and free after a STOP condition. During data transmission, the I²C master can generate repeated START conditions. First START and repeated START conditions are equivalent, function-wise.

Figure 6-12. Start and Stop Conditions

6.5.5.3 Transferring Data

Every byte put on the SDA line must be eight bits long, with the most significant bit (MSB) transferred first. Each byte of data has to be followed by an acknowledge bit. The acknowledge related clock pulse is generated by the master. The master releases the SDA line (HIGH) during the acknowledge clock pulse. The LM3645 pulls down the SDA line during the 9th clock pulse, signifying an acknowledge. The LM3645 generates an acknowledge after each byte is received. There is no acknowledge created after data is read from the device.

After the START condition, the I²C master sends a chip address. This address is seven bits long followed by an eighth bit which is a data direction bit (R/W). The LM3645 7-bit address is 0x65. For the eighth bit, a '0' indicates a WRITE and a '1' indicates a READ. The second byte selects the register to which the data is written. The third byte contains data to write to the selected register.

Figure 6-13. Write Cycle W = Write (SDA = "0") R = Read (SDA = "1") Ack = Acknowledge (SDA Pulled Down by Either Master or Slave) ID = Chip Address, 65h for LM3645

6.5.5.4 I ²C-Compatible Chip Address

The device address for the LM3645 is 1100101 (0x65). After the START condition, the I²C-compatible master sends the 7-bit address followed by an eighth read or write bit (R/W). R/W = 0 indicates a WRITE and R/W = 1 indicates a READ. The second byte following the device address selects the register address to which the data is written. The third byte contains the data for the selected register.

Figure 6-14. I ²C-Compatible Chip Address

6.6 Register Descriptions

6.6.1 MainReg Registers

Table 6-5 lists the memory-mapped registers for the MainReg registers. All register offset addresses not listed in Table 6-5 should be considered as reserved locations and the register contents should not be modified.

Complex bit access types are encoded to fit into small table cells. Table 6-6 shows the codes that are used for access types in this section.

Table 6-6. MainReg Access Type Codes

6.6.1.1 CTRL_REG1 Register (Address = 0x0) [reset = 0x80]

CTRL_REG1 is shown in Figure 6-15 and described in Table 6-7.

Return to [Summary Table](#page-26-0).

The CTRL_REG1 register handles control when operating in voltage output mode and configures the parameters associated with the DC/DC boost converter

Figure 6-15. CTRL_REG1 Register

Table 6-7. CTRL_REG1 Register Field Descriptions

6.6.1.2 CTRL_REG2 Register (Address = 0x1) [reset = 0x0]

CTRL_REG2 is shown in Figure 6-16 and described in Table 6-8.

Return to [Summary Table](#page-26-0).

CTRL_REG2 configures the input and output pins. The Dx_EN bits determine whether or not to enable the Dx current sources, while the upper four bits enable and disable the Strobe1, Strobe2 and Tor/Tx pins. When the Dx EN bits are set to a '1' and the enabled outputs are assigned to either strobe pin or the tor/tx pin, the LEDs will remain off until the externally control pins are set to a '1'. If an Dx_EN bit is set to a '1' and that output is not assigned to an external control pin, the device will begin the turn on sequence assigned to that output. When used in conjunciton with the external control pins, the Dx_EN pins will remain set to a '1' (if enabled) until a fault or a time-out occurs. If the external control pins are not used and a fault or time-out occurs, the device will automatically clear the Dx_EN bit of the offending output.

1 | D2_EN | R/W | 1b0 | D2 Output Pin Enable

1b0 = Disabled 1b1 = Enabled

1b0 = Disabled 1b1 = Enabled

Table 6-8. CTRL_REG2 Register Field Descriptions (continued)

6.6.1.3 D_MODE_REG Register (Address = 0x2) [reset = 0x0]

D_MODE_REG is shown in Figure 6-17 and described in Table 6-9.

Return to [Summary Table](#page-26-0).

The D_MODE_REG sets the mode performance of each current source output. The DC/DC boost will adjust based upon the active modes and LEDs enabled. The mode requiring the highest output voltage will take priority.

Figure 6-17. D_MODE_REG Register

Table 6-9. D_MODE_REG Register Field Descriptions

6.6.1.4 STR_CTRL_REG Register (Address = 0x3) [reset = 0x0]

STR_CTRL_REG is shown in Figure 6-18 and described in Table 6-10.

Return to [Summary Table](#page-26-0).

The STR_CTRL_REG assigns which current sources become enabled when either of the two Strobe pins become active. Each Dx output should only be assigned to one strobe pin at a time.

Figure 6-18. STR_CTRL_REG Register

Table 6-10. STR_CTRL_REG Register Field Descriptions

6.6.1.5 STR_TIME_REG Register (Address = 0x4) [reset = 0xA]

STR_TIME_REG is shown in Figure 6-19 and described in Table 6-11.

Return to [Summary Table](#page-26-0).

STR_TIME_REG sets the flash time-out durations, as well as Strobe type

Table 6-11. STR_TIME_REG Register Field Descriptions

Table 6-11. STR_TIME_REG Register Field Descriptions (continued)

6.6.1.6 D1_FLASH_REG Register (Address = 0x5) [reset = 0x0]

D1_FLASH_REG is shown in Figure 6-20 and described in Table 6-12.

Return to [Summary Table](#page-26-0).

D1_FLASH_REG sets the desired Flash and IR current levels

Figure 6-20. D1_FLASH_REG Register

Table 6-12. D1_FLASH_REG Register Field Descriptions

6.6.1.7 D2_FLASH_REG Register (Address = 0x6) [reset = 0x0]

D2_FLASH_REG is shown in Figure 6-21 and described in Table 6-13.

Return to [Summary Table](#page-26-0).

D2_FLASH_REG sets the desired Flash and IR current levels

Figure 6-21. D2_FLASH_REG Register

Table 6-13. D2_FLASH_REG Register Field Descriptions

6.6.1.8 D3_FLASH_REG Register (Address = 0x7) [reset = 0x0]

D3_FLASH_REG is shown in Figure 6-22 and described in Table 6-14.

Return to [Summary Table](#page-26-0).

D3_FLASH_REG sets the desired Flash and IR current levels

Figure 6-22. D3_FLASH_REG Register

Table 6-14. D3_FLASH_REG Register Field Descriptions

6.6.1.9 D4_FLASH_REG Register (Address = 0x8) [reset = 0x0]

D4_FLASH_REG is shown in Figure 6-23 and described in Table 6-15.

Return to [Summary Table](#page-26-0).

D4_FLASH_REG sets the desired Flash and IR current levels

Figure 6-23. D4_FLASH_REG Register

Table 6-15. D4_FLASH_REG Register Field Descriptions

6.6.1.10 D1_TORCH_REG Register (Address = 0x9) [reset = 0x0]

D1_TORCH_REG is shown in Figure 6-24 and described in Table 6-16.

Return to [Summary Table](#page-26-0).

D1_TORCH_REG sets the desired D1 Torch current level

Figure 6-24. D1_TORCH_REG Register

Table 6-16. D1_TORCH_REG Register Field Descriptions

6.6.1.11 D2_TORCH_REG Register (Address = 0xA) [reset = 0x0]

D2_TORCH_REG is shown in Figure 6-25 and described in Table 6-17.

Return to [Summary Table](#page-26-0).

D2_TORCH_REG sets the desired D2 Torch current level

Figure 6-25. D2_TORCH_REG Register

Table 6-17. D2_TORCH_REG Register Field Descriptions

6.6.1.12 D3_TORCH_REG Register (Address = 0xB) [reset = 0x0]

D3_TORCH_REG is shown in Figure 6-26 and described in Table 6-18.

Return to [Summary Table](#page-26-0).

D3_TORCH_REG sets the desired D3 Torch current level

Figure 6-26. D3_TORCH_REG Register

Table 6-18. D3_TORCH_REG Register Field Descriptions

6.6.1.13 D4_TORCH_REG Register (Address = 0xC) [reset = 0x0]

D4_TORCH_REG is shown in Figure 6-27 and described in Table 6-19.

Return to [Summary Table](#page-26-0).

D4_TORCH_REG sets the desired D4 Torch current level

Figure 6-27. D4_TORCH_REG Register

Table 6-19. D4_TORCH_REG Register Field Descriptions

6.6.1.14 NTC_MODE_REG Register (Address = 0xD) [reset = 0x40]

NTC_MODE_REG is shown in Figure 6-28 and described in Table 6-20.

Return to [Summary Table](#page-26-0).

NTC_MODE_REG Enables and disables the two NTC blocks and sets the drive current and current reduction mode

Figure 6-28. NTC_MODE_REG Register

Table 6-20. NTC_MODE_REG Register Field Descriptions

6.6.1.15 NTC_ASSIGN_REG Register (Address = 0xE) [reset = 0x0]

NTC_ASSIGN_REG is shown in Figure 6-29 and described in Table 6-21.

Return to [Summary Table](#page-26-0).

NTC_ASSIGN_REG determines how each of the Dx output pins responds to a NTC1 or NTC2 event. No output should be assigned to both NTC detection blocks.

Figure 6-29. NTC_ASSIGN_REG Register

Table 6-21. NTC_ASSIGN_REG Register Field Descriptions

6.6.1.16 NTC_VOLT_REG Register (Address = 0xF) [reset = 0x0]

NTC_VOLT_REG is shown in Figure 6-30 and described in Table 6-22.

Return to [Summary Table](#page-26-0).

NTC_VOLT_REG sets the NTC trip voltage

Figure 6-30. NTC_VOLT_REG Register

Bit Field Type Reset Description 7-4 | NTC2_Set | R/W | 4b0000 | NTC2 Trip Voltage. '0000' = 250 mV, '1111' = 1 V, 50 mV steps 3-0 NTC1_Set R/W 4b0000 NTC1 Trip Voltage. '0000' = 250 mV, '1111' = 1 V, 50 mV steps

Table 6-22. NTC_VOLT_REG Register Field Descriptions

6.6.1.17 NTC_READ_REG Register (Address = 0x10) [reset = 0x0]

NTC_READ_REG is shown in Figure 6-31 and described in Table 6-23.

Return to [Summary Table](#page-26-0).

An I2C read of the NTC_READ_REG fetches the voltage on the NTC1 and NTC2 pin for temperature readback

Figure 6-31. NTC_READ_REG Register

Table 6-23. NTC_READ_REG Register Field Descriptions

6.6.1.18 IVFM_SET_REG Register (Address = 0x11) [reset = 0x80]

IVFM_SET_REG is shown in Figure 6-32 and described in Table 6-24.

Return to [Summary Table](#page-26-0).

The IVFM_SET_REG configures the input voltage monitor trip point, mode, hystersis and UVLO functionality. The Ramp Disable option for the Flash_Ramp should not be used if IVFM is enabled.

Figure 6-32. IVFM_SET_REG Register

Table 6-24. IVFM_SET_REG Register Field Descriptions

6.6.1.19 CUR_RAMP_REG Register (Address = 0x12) [reset = 0x39]

CUR_RAMP_REG is shown in Figure 6-33 and described in Table 6-25.

Return to [Summary Table](#page-26-0).

The CUR_RAMP_REG sets the ramp time of the LEDs currents in Torch, Flash and IR modes. The Ramp Disable option for the Flash_Ramp should not be used if IVFM is enabled.

Figure 6-33. CUR_RAMP_REG Register

Table 6-25. CUR_RAMP_REG Register Field Descriptions

6.6.1.20 FAULT_CTRL_REG Register (Address = 0x13) [reset = 0x19]

FAULT_CTRL_REG is shown in Figure 6-34 and described in Table 6-26.

Return to [Summary Table](#page-26-0).

Enables and disables the NTC fault detection and Thermal scale back blocks, along with boost related operation

Figure 6-34. FAULT_CTRL_REG Register

Table 6-26. FAULT_CTRL_REG Register Field Descriptions

6.6.1.21 FLAG_RPT_REG Register (Address = 0x14) [reset = 0x0]

FLAG_RPT_REG is shown in Figure 6-35 and described in Table 6-27.

Return to [Summary Table](#page-26-0).

The FLAG_RPT_REG reports the flag events that can occur. These event flags do not need to be cleared to allow a restart and are for status only.

Figure 6-35. FLAG_RPT_REG Register

Table 6-27. FLAG_RPT_REG Register Field Descriptions

6.6.1.22 VOLT_FAULT_REG Register (Address = 0x15) [reset = 0x0]

VOLT_FAULT_REG is shown in Figure 6-36 and described in Table 6-28.

Return to [Summary Table](#page-26-0).

The VOLT_FAULT_REG register signals voltage faults that occur during any of the operating modes.

Table 6-28. VOLT_FAULT_REG Register Field Descriptions

6.6.1.23 THERM_FAULT_REG Register (Address = 0x16) [reset = 0x0]

THERM_FAULT_REG is shown in Figure 6-37 and described in Table 6-29.

Return to [Summary Table](#page-26-0).

The THERM_FAULT_REG is used to report termperature related faults that occur during any of the operating modes

Figure 6-37. THERM_FAULT_REG Register

Table 6-29. THERM_FAULT_REG Register Field Descriptions

6.6.1.24 LAST_CUR_REG1 Register (Address = 0x17) [reset = 0x0]

LAST_CUR_REG1 is shown in Figure 6-38 and described in Table 6-30.

Return to [Summary Table](#page-26-0).

Stores the last current value reported as a fraction of full scale in the event of a current scale back event (IVFM, thermal scale back, etc)

Figure 6-38. LAST_CUR_REG1 Register

Table 6-30. LAST_CUR_REG1 Register Field Descriptions

6.6.1.25 LAST_CUR_REG2 Register (Address = 0x18) [reset = 0x0]

LAST_CUR_REG2 is shown in Figure 6-39 and described in Table 6-31.

Return to [Summary Table](#page-26-0).

Stores the last current value reported as a fraction of full scale in the event of a current scale back event (IVFM, thermal scale back, etc)

Figure 6-39. LAST_CUR_REG2 Register

Table 6-31. LAST_CUR_REG2 Register Field Descriptions

6.6.1.26 LAST_CUR_REG3 Register (Address = 0x19) [reset = 0x0]

LAST_CUR_REG3 is shown in Figure 6-40 and described in Table 6-32.

Return to [Summary Table](#page-26-0).

Stores the last current value reported as a fraction of full scale in the event of a current scale back event (IVFM, thermal scale back, etc)

Figure 6-40. LAST_CUR_REG3 Register

Table 6-32. LAST_CUR_REG3 Register Field Descriptions

6.6.1.27 LAST_CUR_REG4 Register (Address = 0x1A) [reset = 0x0]

LAST_CUR_REG4 is shown in Figure 6-41 and described in Table 6-33.

Return to [Summary Table](#page-26-0).

Stores the last current value reported as a fraction of full scale in the event of a current scale back event (IVFM, thermal scale back, etc)

Figure 6-41. LAST_CUR_REG4 Register

Table 6-33. LAST_CUR_REG4 Register Field Descriptions

6.6.1.28 DEV_INFO_REG Register (Address = 0x1B) [reset = 0x41]

DEV_INFO_REG is shown in Figure 6-42 and described in Table 6-34.

Return to [Summary Table](#page-26-0).

Stores Device Information

Figure 6-42. DEV_INFO_REG Register

Table 6-34. DEV_INFO_REG Register Field Descriptions

7 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

7.1 Application Information

The LM3645 can drive four flash LEDs at currents up to 2 A per LED. The total LED current the LM3645 boost can deliver is 5 A (I_{LED1} + I_{LED2} + I_{LED3} + I_{LED4}). The 2-MHz or 4-MHz DC-DC boost regulator allows for the use of small value discrete external components.

7.2 Typical Application

Figure 7-1. LM3645 Typical Application

7.2.1 Design Requirements

Example requirements based on default register values:

7.2.2 Output Control Examples

The LM3645 is highly configurable enabling multiple different LED drive use cases. To highlight the flexibility, sample sequences for different use cases are provided. Additional features such as Tx control and NTC

interactions can be enabled and disabled in each mode, however they are not included in the following examples for simplicity.

7.2.2.1 Four Channel Flash with Strobe1 Trigger Starting in Standby

The following sequence highlights how to control all four current sources in Flash mode starting with the LM3645 in standby.

7.2.2.2 Four Channel Flash with Strobe1 Trigger Starting in I2C Torch

The following sequence highlights how to control all four current sources in Flash mode after I2C Torch mode has already been enabled.

7.2.2.3 Mixed Mode Functionality

[Figure 7-2](#page-59-0) highlights how to control LED1 and LED2 in flash mode, with LED3 set to IR Mode and LED 4 set to external Torch Mode.

Figure 7-2. Mixed Mode Use Case

Because an output is assigned to IR Mode (LED3), the OUT pin voltage increases to approximately VIN until current is delivered to the outputs. If Strobe1 goes high, LED1 and LED2 enable in Flash. If Strobe2 goes high, LED3 operates in IR Mode. If TOR/TX goes high LED4 enable in Torch Mode.

7.2.2.4 Voltage Mode Only

The following sequence highlights how to place the LM3645 into fixed output voltage mode with the current sources disabled.

7.2.2.5 Voltage Mode With Advanced IR

The following sequence highlights how to place the LM3645 into fixed output voltage mode with two of the LED outputs enabled and both strobe pins (D1 assigned to STR1, D4 assigned to STR2)

When the strobe pins go high, the current to the LED outputs will ramp to their target. When the strobe pins go low, the LED outputs stop delivering current and VOUT will stay at 4V ready for the next strobe pulses.

7.2.3 Detailed Design Procedure

7.2.3.1 Snubber Requirement

The LM3645 requires a snubber circuit attached to the SW (switch) node to suppress voltage spikes during switching. It is recommended that a 1 ohm resistor in series with a 4 nF capacitance be used. The ceramic snubber capacitor must have a minimum voltage rating of 6.3V and the effective capacitance must take into account the effects of DC bias.

7.2.3.2 Output Capacitor Selection

The LM3645 is designed to operate with a single 10-µF ceramic output capacitor, however two 10 µF capacitors in parallel is recommended to minimize LED current ripple. When the boost converter is running, the output capacitor supplies the load current during the boost converter on-time. When the NMOS switch turns off, the inductor energy is discharged through the internal PMOS switch, supplying power to the load and restoring charge to the output capacitor. This causes a sag in the output voltage during the on-time and a rise in the output voltage during the off-time. The output capacitor is therefore chosen to limit the output ripple to an acceptable level depending on load current and input/output voltage differentials and also to ensure the converter remains stable. To maintain stable boost operation, it is recommended that a minimum effective output capacitance (accounting for DC bias, temperature and part to part variation) be greater than **3 µF**.

Larger capacitors such as a 22-µF or capacitors in parallel can be used if lower output voltage ripple is desired. To estimate the output voltage ripple considering the ripple due to capacitor discharge (ΔV_{Ω}) and the ripple due to the capacitors ESR (ΔV_{FSR}) use the following equations:

For continuous conduction mode, the output voltage ripple due to the capacitor discharge is:

$$
\Delta V_{\rm Q} = \frac{I_{\rm LED} \times (V_{\rm OUT} - V_{\rm IN})}{f_{\rm SW} \times V_{\rm OUT} \times C_{\rm OUT}} \tag{3}
$$

The output voltage ripple due to the output capacitors ESR is found by:

$$
\Delta V_{ESR} = R_{ESR} \times \left(\frac{I_{LED} \times V_{OUT}}{V_{IN}} + \Delta I_{L} \right)
$$

where

$$
\Delta I_{L} = \frac{V_{IN} \times (V_{OUT} - V_{IN})}{2 \times f_{SW} \times L \times V_{OUT}}
$$
 (4)

In ceramic capacitors the ESR is very low so the assumption is that 80% of the output voltage ripple is due to capacitor discharge and 20% from ESR.

7.2.3.3 Input Capacitor Selection

Choosing the correct size and type of input capacitor helps minimize the voltage ripple caused by the switching of the LM3645 boost converter and reduce noise on the boost converter's input pin that can feed through and disrupt internal analog signals. In the typical application circuit a 10-µF ceramic input capacitor works well. It is

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important to place the input capacitor as close as possible to the LM3645 input (IN) pin. This reduces the series resistance and inductance that can inject noise into the device due to the input switching currents.

7.2.3.4 Inductor Selection

The LM3645 is designed to use a 0.47-µH or 1-µH inductor. When the device is boosting ($V_{\text{OUT}} > V_{\text{IN}}$) the inductor is typically the largest area of efficiency loss in the circuit. Therefore, choosing an inductor with the lowest possible series resistance is important. Additionally, the saturation rating of the inductor should be greater than the maximum operating peak current of the LM3645. This prevents excess efficiency loss that can occur with inductors that operate in saturation. For proper inductor operation and circuit performance, ensure that the inductor saturation and the peak current limit setting of the LM3645 are greater than I_{PEAK} in the following calculation:

$$
I_{PEAK} = \frac{I_{LOAD}}{\eta} \times \frac{V_{OUT}}{V_{IN}} + \Delta I_L \quad \text{where} \quad \Delta I_L = \frac{V_{IN} \times (V_{OUT} - V_{IN})}{2 \times f_{SW} \times L \times V_{OUT}}
$$
(5)

where

• f_{SW} = 2 MHz or 4 MHz

7.2.4 Application Curves

Ambient temperature is 25°C, input voltage is 3.6 V, EN = V_{IN}, C_{IN} = 2 × 10 µF, C_{OUT} = 2 × 10 µF and L = 1 µH, unless otherwise noted.

[LM3645](https://www.ti.com/product/LM3645) [SNVSCV4](https://www.ti.com/lit/pdf/SNVSCV4) – SEPTEMBER 2024

8 Power Supply Recommendations

The LM3645 is designed to operate from an input voltage supply range between 2.3 V and 5.5 V. This input supply must be well regulated and capable to supply the required input current. If the input supply is located far from the LM3645 additional bulk capacitance may be required in addition to the ceramic bypass capacitors.

9 Layout

9.1 Layout Guidelines

The high switching frequency and large switching currents of the LM3645 make the choice of layout important. Follow these guidelines to ensure the device is stable and maintains proper LED current regulation across its intended operating voltage and current range.

- Place C_{IN} on the top layer (same layer as the LM3645) and as close to the device as possible. The input capacitor conducts the driver currents during the low-side MOSFET turn-on and turn-off and can detect current spikes over 1 A in amplitude. Connecting the input capacitor through short, wide traces to both the IN and GND pins reduces the inductive voltage spikes that occur during switching which can corrupt the V_{IN} line.
- Place C_{OUT} on the top layer (same layer as the LM3645) and as close as possible to the OUT and GND pin. The returns for both C_{IN} and C_{OUT} should come together at one point, as close to the GND pin as possible. Connecting C_{OUT} through short, wide traces reduce the series inductance on the OUT and GND pins that can corrupt the V_{OUT} and GND lines and cause excessive noise in the device and surrounding circuitry.
- Connect the inductor on the top layer close to the SW pin. There should be a low-impedance connection from the inductor to SW due to the large DC inductor current, and at the same time the area occupied by the SW node should be small so as to reduce the capacitive coupling of the high dV/dT present at SW that can couple into nearby traces.
- Avoid routing logic traces near the SW node so as to avoid any capacitively coupled voltages from SW onto any high-impedance logic lines such as TOR/Tx, STR1, STR2, EN, SDA, and SCL. A good approach is to insert an inner layer GND plane underneath the SW node and between any nearby routed traces. This creates a shield from the electric field generated at SW.
- Terminate the Flash LED cathodes directly to the GND pin of the LM3645. If possible, route the LED returns with a dedicated path so as to keep the high amplitude LED currents out of the GND plane. For Flash LEDs that are routed relatively far away from the LM3645, a good approach is to sandwich the forward and return current paths over the top of each other on two layers. This helps reduce the inductance of the LED current paths.

9.2 Layout Example

Figure 9-1. Layout Example

10 Device and Documentation Support

10.1 Third-Party Products Disclaimer

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10.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on [ti.com.](https://www.ti.com) Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

10.3 Support Resources

TI E2E™ [support forums](https://e2e.ti.com) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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10.4 Trademarks

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10.5 Electrostatic Discharge Caution

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

10.6 Glossary

[TI Glossary](https://www.ti.com/lit/pdf/SLYZ022) This glossary lists and explains terms, acronyms, and definitions.

11 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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TEXAS

TAPE AND REEL INFORMATION

ISTRUMENTS

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

www.ti.com 7-Nov-2024

PACKAGE MATERIALS INFORMATION

*All dimensions are nominal

PACKAGE OUTLINE

YCG0025 DSBGA - 0.5 mm max height

DIE SIZE BALL GRID ARRAY

NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.

EXAMPLE BOARD LAYOUT

YCG0025 **DSBGA - 0.5 mm max height**

DIE SIZE BALL GRID ARRAY

NOTES: (continued)

3. Final dimensions may vary due to manufacturing tolerance considerations and also routing constraints. See Texas Instruments Literature No. SNVA009 (www.ti.com/lit/snva009).

EXAMPLE STENCIL DESIGN

YCG0025 **DSBGA - 0.5 mm max height**

DIE SIZE BALL GRID ARRAY

NOTES: (continued)

4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release.

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