

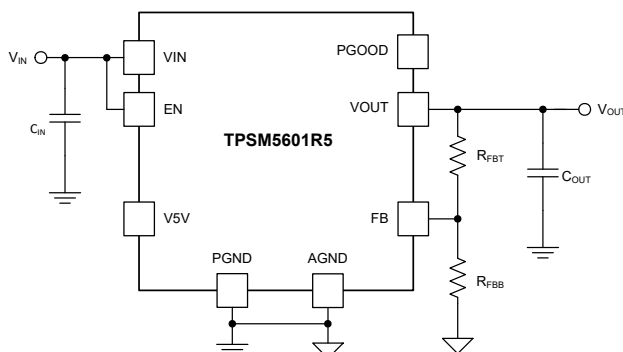
# TPSM5601R5 采用增强型 Hotrod™ QFN 封装的 60V 输入、1V 至 6V 输出、1.5A 电源模块

## 1 特性

- 提供功能安全
  - 可提供用于功能安全系统设计的文档
- 5.0mm × 5.5mm × 4.0mm 增强型 HotRod™ QFN
  - 出色的热性能：在 85°C 且无散热的情况下高达 18W 的输出功率
  - 标准封装尺寸：单个大型散热焊盘和所有引脚均分布在封装外围
- 专为可靠耐用的应用而设计
  - 宽输入电压范围：4.2V 至 60V
  - 高达 66V 的输入电压瞬态保护
  - 工作结温范围：-40°C 至 +125°C
  - 带 EXT 后缀器件的结温范围：-55°C 至 +125°C
- 400kHz 固定开关频率
- FPWM 运行模式
- 针对超低 EMI 要求进行了优化
  - 集成屏蔽式电感器和 高频旁路电容器
  - 符合 EN55011 EMI 标准
  - 扩频选项降低干扰 (EMI)
- 26μA 非开关静态电流
- 单调启动至预偏置输出
- 无环路补偿或自举组件
- 具有迟滞功能的精密使能和输入 UVLO
- 具有迟滞功能的热关断保护
- 使用 TPSM5601R5 并借助 WEBENCH® Power Designer 创建定制设计

## 2 应用

- 现场发送器和传感器、PLC 模块
- 恒温器、视频监控、HVAC 系统
- 交流和伺服驱动器、旋转编码器
- 工业运输、资产跟踪
- 负输出应用



典型电路原理图

## 3 说明

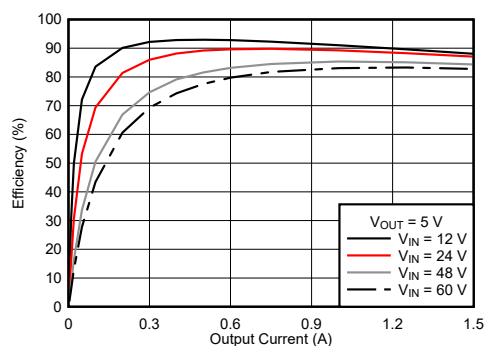
TPSM5601R5 电源模块是一款高度集成的 1.5A 电源解决方案，在热增强型 QFN 封装内整合了一个带有功率 MOSFET 的 60V 输入降压直流/直流转换器、一个屏蔽式电感器和多个无源器件。该 5mm × 5.5mm × 4mm、15 引脚 QFN 封装采用增强型 HotRod QFN 技术来实现增强的热性能、小尺寸和低 EMI。封装引脚外露，具有单个大型散热焊盘，方便布局布线和组装。

TPSM5601R5 是一款紧凑、易用的电源模块，具有 1.0V 至 6V 的可调节宽输出电压范围。该总体解决方案仅需四个外部元件，并且省去了设计流程中的环路补偿和磁性元件选型过程。TPSM5601R5 具有全套功能，包括电源正常状态指示、可编程 UVLO、预偏置启动、过流和温度保护，因此是为各种应用供电的出色器件。空间受限型应用可从 5.0mm × 5.5mm 封装中受益。此外，TPSM5601R5S 还提供展频操作。

### 器件信息

器件型号	封装 <sup>(1)</sup>	封装尺寸 (标称值)
TPSM5601R5	QFN (15)	5.0mm × 5.5mm
TPSM5601R5S		

- (1) 如需了解所有可用封装，请参阅数据表末尾的可订购产品附录。



典型效率 (V<sub>OUT</sub> = 5V)



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## 4 Revision History

DATE	REVISION	NOTES
September 2021	*	Initial release

## 5 Device Comparison Table

DEVICE	DESCRIPTION
TPSM5601R5	60-V input voltage, 1-V to 6-V output voltage, 1.5-A power module, fixed 400-kHz switching, -40°C to +125°C operating junction temperature range
TPSM5601R5S	Equivalent to TPSM5601R5, but with spread spectrum operation

## 6 Pin Configuration and Functions

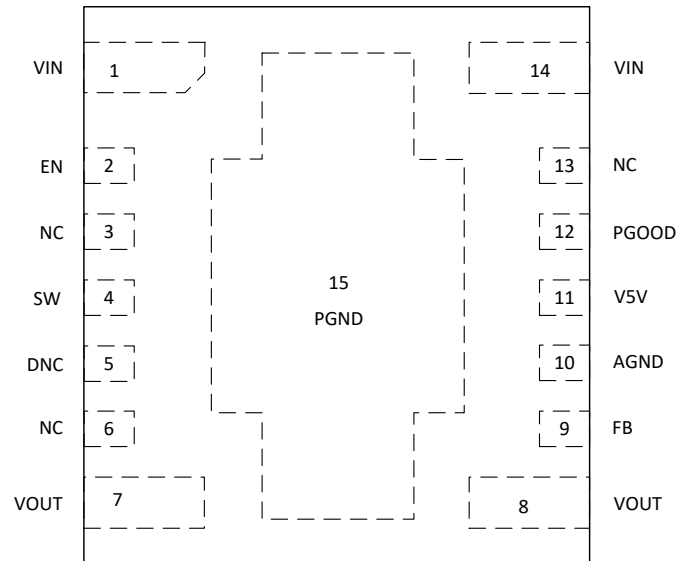


图 6-1. 15-Pin QFN RDA Package (Top View)

表 6-1. Pin Functions

PIN		TYPE <sup>(1)</sup>	DESCRIPTION
NO.	NAME		
10	AGND	G	Analog ground. Zero voltage reference for internal references and logic. All electrical parameters are measured with respect to this pin. <i>This pin must be connected to PGND at a single point.</i> See 节 11.2 for a recommended layout.
5	DNC	—	Do not connect. Do not connect this pin to ground, to another pin, or to any other voltage. This pin is connected to the internal bootstrap capacitor. This pin must be soldered to an isolated pad.
2	EN	I	Enable pin. This pin turns the converter on when pulled high and turns off the converter when pulled low. This pin can be connected directly to VIN. <i>Do not float.</i> This pin can be used to set the input undervoltage lockout with two resistors. See 节 8.3.4.
9	FB	I	Feedback input. Connect the mid-point of the feedback resistor divider to this pin. Connect the upper resistor ( $R_{FBT}$ ) of the feedback divider to $V_{OUT}$ at the desired point of regulation. Connect the lower resistor ( $R_{FBB}$ ) of the feedback divider to AGND.
3, 6, 13	NC	—	Not connected. These pins are not connected to any circuitry within the module. Leaving these pins unconnected to any other signal increases spacing near the high voltage pins (VIN, SW, EN, and DNC). However, if the high voltage spacing is not needed in the application, connecting these pins to the PGND plane can help enhance shielding and thermal performance.
15	PGND	G	Power ground. This is the return current path for the power stage of the device. Connect this pad to the input supply return, load return, and capacitors associated with the VIN and VOUT pins. See 节 11.2 for a recommended layout.
12	PGOOD	O	Power-good pin. Open-drain output that asserts low if the feedback voltage is not within the specified window thresholds. A 10-k $\Omega$ to 100-k $\Omega$ pullup resistor is required and can be tied to the V5V pin or other DC voltage less than 18 V. If not used, this pin can be left open or connected to PGND.
4	SW	O	Switch node. Do not place any external component on this pin or connect to any signal.
1, 14	VIN	I	Input supply voltage. Connect the input supply to these pins. Connect input capacitors between these pins and PGND in close proximity to the device.

表 6-1. Pin Functions (continued)

PIN		TYPE <sup>(1)</sup>	DESCRIPTION
NO.	NAME		
7, 8	VOUT	O	Output voltage. These pins are connected to the internal output inductor. Connect these pins to the output load and connect external output capacitors between these pins and PGND.
11	V5V	O	Internal 5-V LDO output. Supplies internal control circuits. Do not connect to external loads. This pin can be used as logic supply for the PGOOD pin.

(1) G = Ground, I = Input, O = Output

## 7 Specifications

### 7.1 Absolute Maximum Ratings

Over the operating ambient temperature range<sup>(1)</sup>

PARAMETER		MIN	MAX	UNIT
Input voltage	VIN to PGND	- 0.3	66	V
	EN to AGND <sup>(2)</sup>	- 0.3	V <sub>IN</sub> + 0.3	
	PGOOD to AGND <sup>(2)</sup>	- 0.3	22	
	FB to AGND	- 0.3	5.5	
	AGND to PGND	- 0.3	0.3	
Output voltage	VOUT to PGND <sup>(2)</sup>	- 0.3	30	
	VCC to AGND	0	5.5	
Operating IC junction temperature, T <sub>J</sub> <sup>(3)</sup>		- 40	125	°C
Storage temperature, T <sub>stg</sub>		- 55	150	°C
Peak reflow case temperature			245	°C
Maximum number of reflows allowed			3	
Mechanical vibration	Mil-STD-883H, Method 2007.3, 1 msec, 1/2 sine, mounted		20	G
Mechanical shock	Mil-STD-883H, Method 2002.5, 20 to 2000Hz		500	G

- (1) Operation outside the *Absolute Maximum Ratings* may cause permanent device damage. *Absolute Maximum Ratings* do not imply functional operation of the device at these or any other conditions beyond those listed under *Recommended Operating Conditions*. If used outside the *Recommended Operating Conditions* but within the *Absolute Maximum Ratings*, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.
- (2) The voltage on this pin must not exceed the voltage on the VIN pin by more than 0.3 V.
- (3) The ambient temperature is the air temperature of the surrounding environment. The junction temperature is the temperature of the internal power IC when the device is powered. Operating below the maximum ambient temperature, as shown in the safe operating area (SOA) curves in the typical characteristics sections, ensures that the maximum junction temperature of any component inside the module is never exceeded.

### 7.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic discharge	Human-body model (HBM) <sup>(1)</sup>	±1500	V
		Charged-device model (CDM) <sup>(2)</sup>	±1500	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

## 7.3 Recommended Operating Conditions

Over operating ambient temperature range (unless otherwise noted) <sup>(1)</sup>

	MIN	MAX	UNIT
Input voltage, $V_{IN}$	4.2	60	V
Output voltage, $V_{OUT}$	1	6 <sup>(3)</sup>	V
Output current, $I_{OUT}$	0	1.5	A
EN voltage, $V_{EN}$ <sup>(2)</sup>	0	$V_{IN}$	V
PGOOD pullup voltage, $V_{PGOOD}$ <sup>(2)</sup>	0	18	V
Operating ambient temperature, $T_A$	-40	105	°C

- (1) Recommended operating conditions indicate conditions for which the device is intended to be functional, but do not ensure specific performance limits. For ensured specifications, see the [Electrical Characteristics](#).
- (2) The voltage on this pin must not exceed the voltage on the VIN pin by more than 0.3 V.
- (3) The recommended maximum output voltage varies depending input voltage.

## 7.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		TPSM5601R5		UNIT
		RDA (QFN)		
		15 PINS		
$R_{\theta JA}$	Junction-to-ambient thermal resistance <sup>(2)</sup>	Nat Conv	20.4	°C/W
		100 LFM	18.9	°C/W
		200 LFM	17.6	°C/W
$\psi_{JT}$	Junction-to-top characterization parameter <sup>(3)</sup>		3.6	°C/W
$\psi_{JB}$	Junction-to-board characterization parameter <sup>(4)</sup>		15.3	°C/W
$T_{SHDN}$	Thermal shutdown temperature		170	°C
	Recovery temperature		158	°C

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.
- (2) The junction-to-ambient thermal resistance,  $R_{\theta JA}$ , applies to devices soldered directly to a 6.35-cm × 8.25-cm, four-layer PCB with 2-oz. copper. Additional airflow and PCB copper area reduces  $R_{\theta JA}$ . See [Section 11.2.1](#) for more information.
- (3) The junction-to-top board characterization parameter,  $\psi_{JT}$ , estimates the junction temperature,  $T_J$ , of a device in a real system, using a procedure described in JESD51-2A (section 6 and 7).  $T_J = \psi_{JT} \times P_{dis} + T_T$ ; where  $P_{dis}$  is the power dissipated in the device and  $T_T$  is the temperature of the top of the device.
- (4) The junction-to-board characterization parameter,  $\psi_{JB}$ , estimates the junction temperature,  $T_J$ , of a device in a real system, using a procedure described in JESD51-2A (sections 6 and 7).  $T_J = \psi_{JB} \times P_{dis} + T_B$ ; where  $P_{dis}$  is the power dissipated in the device and  $T_B$  is the temperature of the board 1mm from the device.

## 7.5 Electrical Characteristics

Limits apply over  $T_A = -40^\circ\text{C}$  to  $+105^\circ\text{C}$ ,  $V_{IN} = 24\text{ V}$ ,  $V_{OUT} = 3.3\text{ V}$ ,  $I_{OUT} = 1.5\text{ A}$ , 600 mA, (unless otherwise noted); minimum and maximum limits are specified through production test or by design. Typical values represent the most likely parametric norm and are provided for reference only.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>INPUT VOLTAGE (<math>V_{IN}</math>)</b>						
$V_{IN}$	Input voltage range	Over $I_{OUT}$ range	4.2 <sup>(1)</sup>		60	V
	$V_{IN}$ turn-on	$V_{IN}$ increasing, $I_{OUT} = 0\text{ A}$ , $V_{EN} = V_{IN}$		3.8		V
	$V_{IN}$ turn-off	$V_{IN}$ decreasing, $I_{OUT} = 0\text{ A}$ , $V_{EN} = V_{IN}$		3.3		V
$I_{SHDN}$	Shutdown supply current	$V_{EN} = 0\text{ V}$ , $I_{OUT} = 0\text{ A}$		5		$\mu\text{A}$
<b>INTERNAL LDO (<math>V5V</math>)</b>						
$V5V$	Internal LDO output voltage appearing at the $V5V$ pin	$6\text{ V} \leq V_{IN} \leq 60\text{ V}$	4.75	5	5.25	V
<b>FEEDBACK</b>						
$V_{FB}$	Load regulation	$T_A = +25^\circ\text{C}$ , $0\text{ A} \leq I_{OUT} \leq 1.5\text{ A}$		0.057%		
$V_{FB}$	Line regulation	$T_A = +25^\circ\text{C}$ , $I_{OUT} = 0\text{ A}$ , $6\text{ V} \leq V_{IN} \leq 60\text{ V}$		0.024%		
$I_{FB}$	Current into FB pin	$FB = 1\text{ V}$		0.2		nA
<b>CURRENT</b>						
$I_{OUT}$	Output current	$T_A = 25^\circ\text{C}$	0		1.5	A
$I_{OUT}$	Overcurrent threshold	$V_{OUT} = 3.3\text{ V}$ , $T_A = 25^\circ\text{C}$		1.9		A
$V_{HC}$	FB pin voltage required to trip short-circuit Hiccup mode			0.4		V
$t_{HC}$	Time between current-limit hiccup burst			94		ms
<b>ENABLE (EN PIN)</b>						
$V_{EN-VCC-H}$	EN input level required to turn on internal LDO	Rising threshold			1.14	V
$V_{EN-VCC-L}$	EN input level required to turn off internal LDO	Falling threshold	0.3			V
$V_{EN-H}$	EN input level required to start switching	Rising threshold	1.157	1.231	1.30	V
$V_{EN-HYS}$	Hysteresis below $V_{EN-H}$	Hysteresis below $V_{EN-H}$ ; falling		110		mV
$I_{LKG-EN}$	Enable input leakage current	$V_{EN} = 3.3\text{ V}$		0.2		nA
<b>POWER GOOD (PGOOD PIN)</b>						
$V_{PG-LOW-UP}$	$V_{OUT}$ rising (fault)	% of FB voltage		107%		
$V_{PG-HIGH-DN}$	$V_{OUT}$ falling (good)	% of FB voltage		105%		
$V_{PG-HIGH-UP}$	$V_{OUT}$ rising (good)	% of FB voltage		95%		
$V_{PG-LOW-DN}$	$V_{OUT}$ falling (fault)	% of FB voltage		93%		
$R_{PG}$	Power-good flag $R_{DSON}$	$V_{EN} = 0\text{ V}$		35		$\Omega$
$V_{IN-PG}$	Minimum input voltage for proper PGOOD function	$I_{PG} = 50\text{ }\mu\text{A}$ , $EN = 0\text{ V}$			2	V
<b>PERFORMANCE</b>						
$\eta$	Efficiency	$V_{OUT} = 3.3\text{ V}$ , $I_{OUT} = 0.75\text{ A}$ , $T_A = 25^\circ\text{C}$		81%		
$\eta$	Efficiency	$V_{OUT} = 5.0\text{ V}$ , $I_{OUT} = 0.75\text{ A}$ , $T_A = 25^\circ\text{C}$		86%		
<b>SOFT START</b>						
$t_{SS}$	Internal soft-start time			4.5		ms
<b>SWITCHING FREQUENCY</b>						
$f_{SW}$	Switching frequency	$I_{OUT} = 0.75\text{ A}$ , $T_A = 25^\circ\text{C}$	340	400 <sup>(2)</sup>	460	kHz
$f_{SW}$ ss device	Switching frequency for spread spectrum device only	$I_{OUT} = 0.75\text{ A}$ , $T_A = 25^\circ\text{C}$	320	400	480	kHz

(1) The recommended minimum  $V_{IN}$  is 4.2 V or ( $V_{OUT} + 600\text{ mV}$ ), whichever is greater.

(2) The typical switching frequency of this device will change based on operating conditions. See the Switching Frequency section for more information.

## 7.6 Typical Characteristics ( $V_{IN} = 12\text{ V}$ )

$T_A = 25^\circ\text{C}$ , unless otherwise noted.

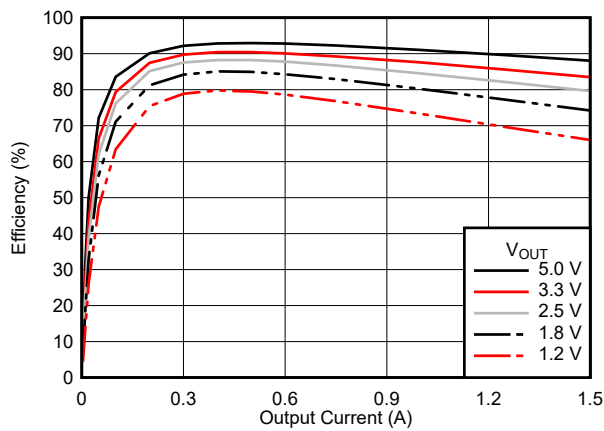


图 7-1. Efficiency

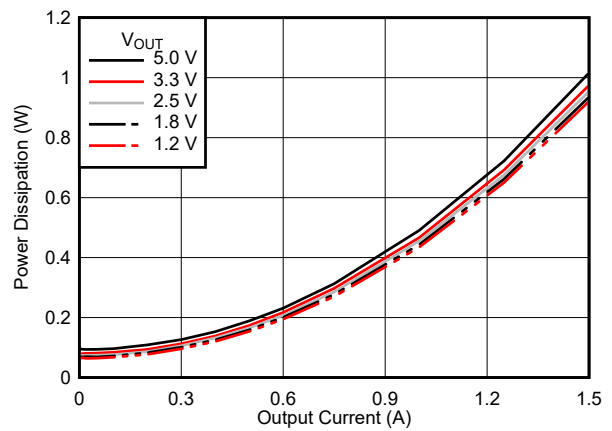
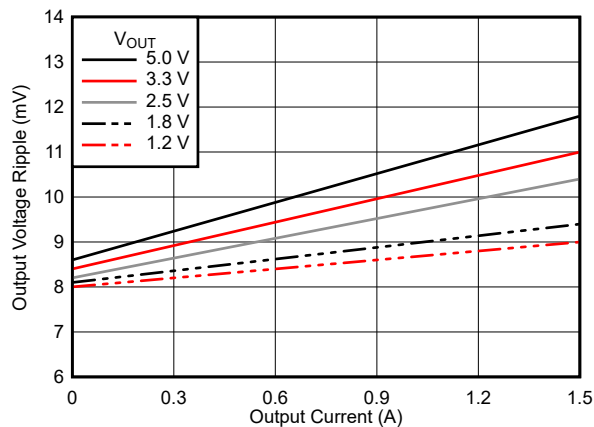
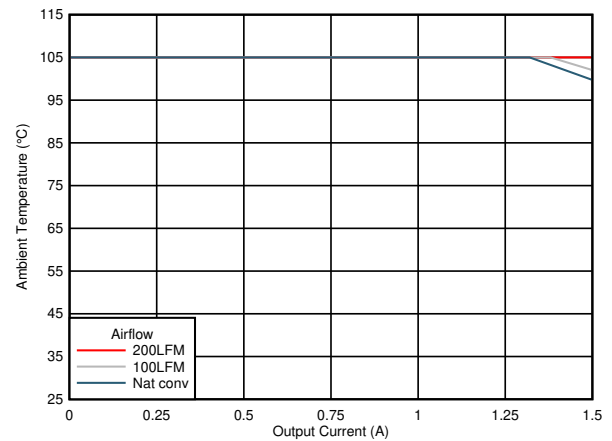


图 7-2. Power Dissipation



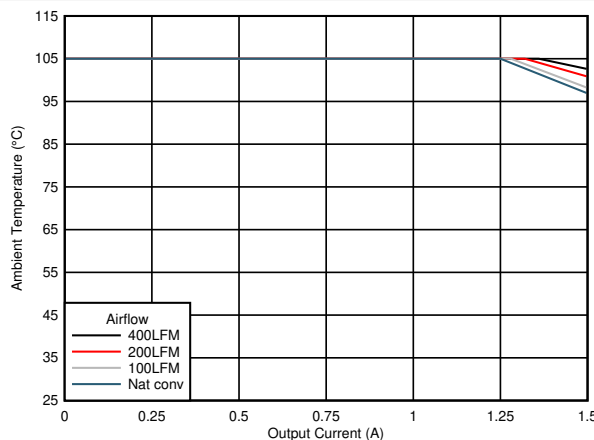
$C_{OUT} = 4 \times 100\ \mu\text{F}$ , 10-V, ceramic

图 7-3. Output Voltage Ripple



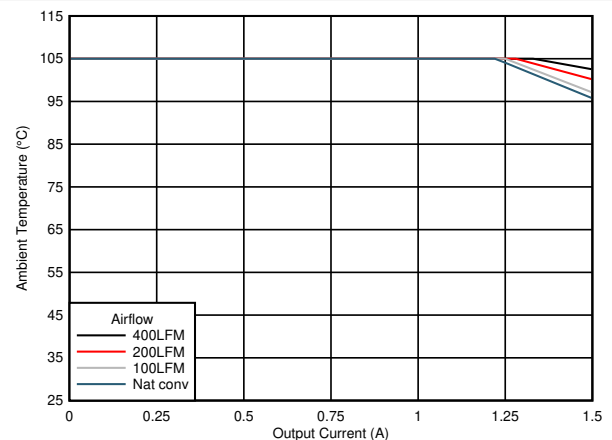
Device soldered to a 63.5-mm  $\times$  82.5-mm, 4-layer PCB

图 7-4. Safe Operating Area ( $V_{OUT} = 1.2\text{ V}$ )



Device soldered to a 63.5-mm  $\times$  82.5-mm, 4-layer PCB

图 7-5. Safe Operating Area ( $V_{OUT} = 3.3\text{ V}$ )



Device soldered to a 63.5-mm  $\times$  82.5-mm, 4-layer PCB

图 7-6. Safe Operating Area ( $V_{OUT} = 5.0\text{ V}$ )



## 7.7 Typical Characteristics ( $V_{IN} = 24\text{ V}$ )

$T_A = 25^\circ\text{C}$ , unless otherwise noted.

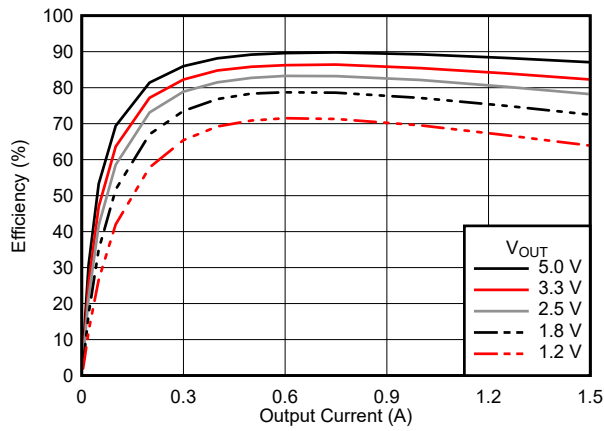


图 7-7. Efficiency

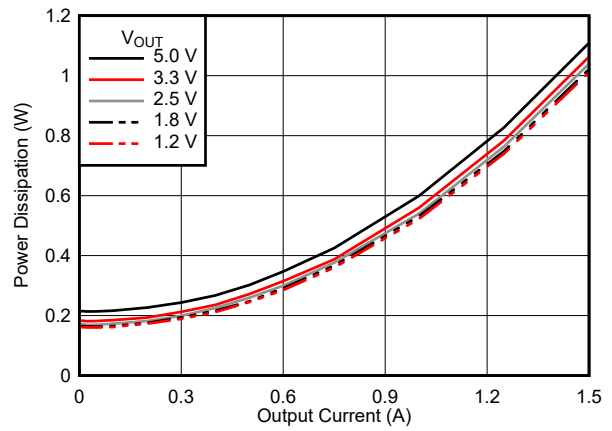
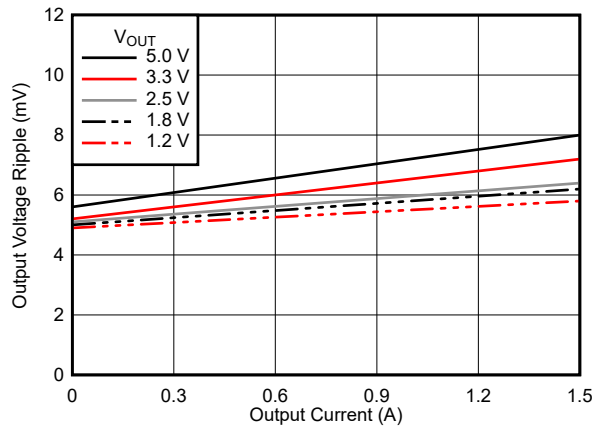
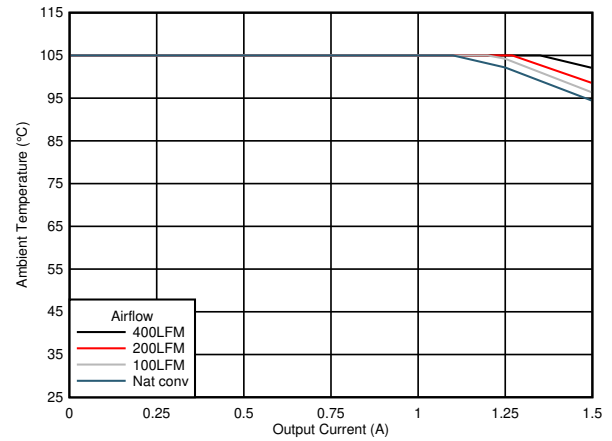


图 7-8. Power Dissipation



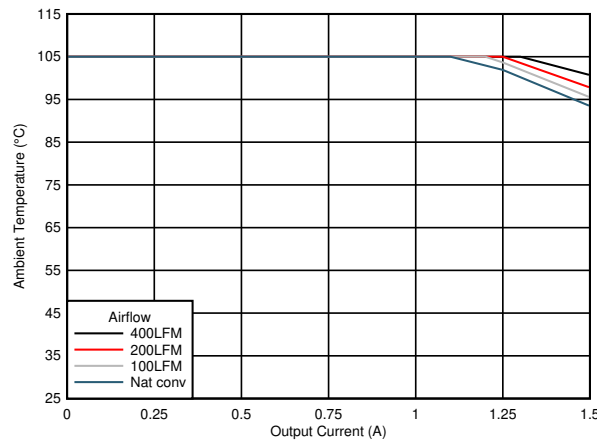
$C_{OUT} = 4 \times 100\text{-}\mu\text{F}$ , 10-V, ceramic

图 7-9. Output Voltage Ripple



Device soldered to a 63.5-mm  $\times$  82.5-mm, 4-layer PCB

图 7-10. Safe Operating Area ( $V_{OUT} = 1.8\text{ V}$ )



Device soldered to a 63.5-mm  $\times$  82.5-mm, 4-layer PCB

图 7-11. Safe Operating Area ( $V_{OUT} = 5.0\text{ V}$ )

## 7.8 Typical Characteristics ( $V_{IN} = 48\text{ V}$ )

$T_A = 25^\circ\text{C}$ , unless otherwise noted.

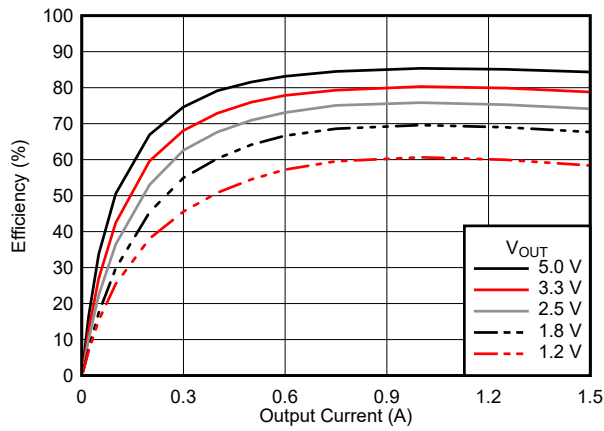


图 7-12. Efficiency

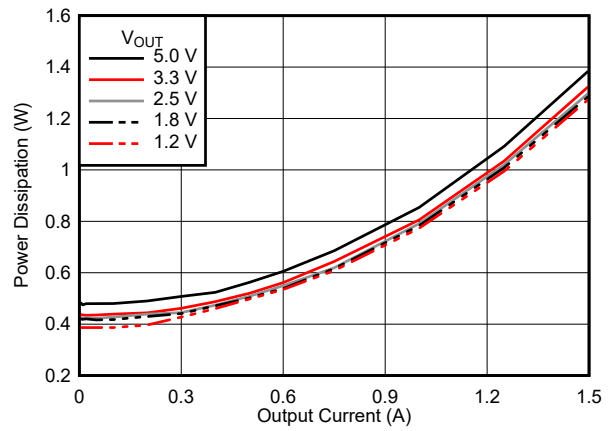
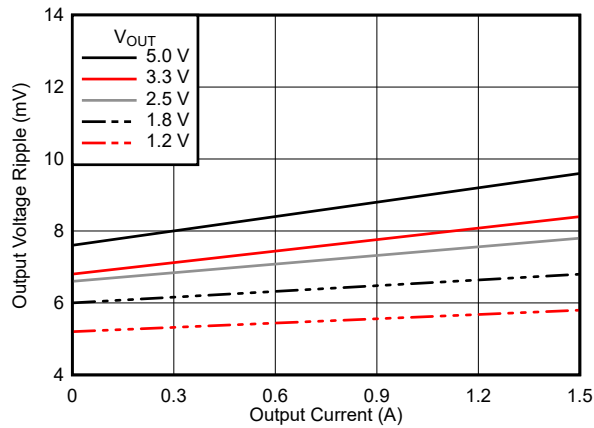
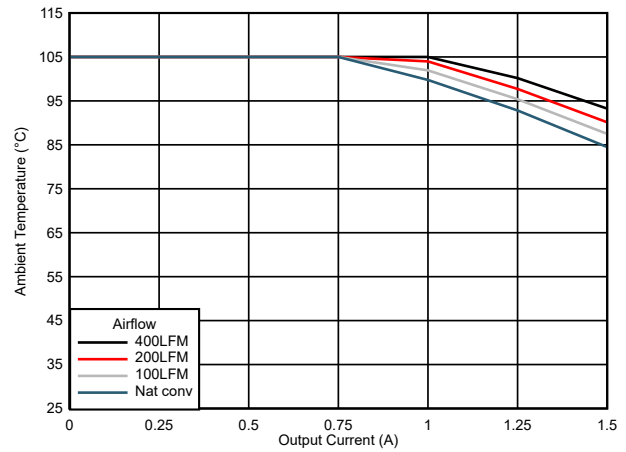


图 7-13. Power Dissipation



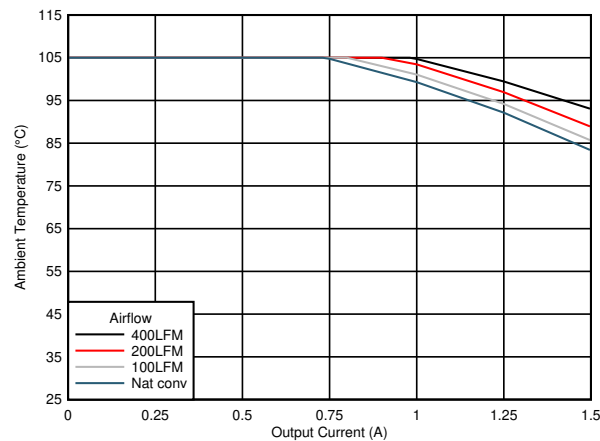
$C_{OUT} = 4 \times 100\text{-}\mu\text{F}$ , 10-V, ceramic

图 7-14. Output Voltage Ripple



Device soldered to a 63.5-mm  $\times$  82.5-mm, 4-layer PCB

图 7-15. Safe Operating Area ( $V_{OUT} = 3.3\text{ V}$ )



Device soldered to a 63.5-mm  $\times$  82.5-mm, 4-layer PCB

图 7-16. Safe Operating Area ( $V_{OUT} = 5.0\text{ V}$ )

## 7.9 Typical Characteristics ( $V_{IN} = 60\text{ V}$ )

$T_A = 25^\circ\text{C}$ , unless otherwise noted.

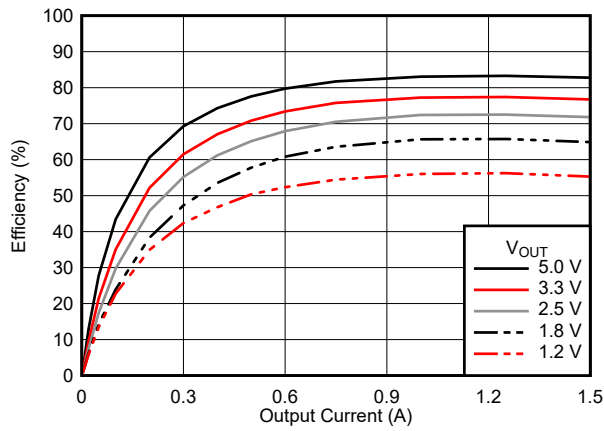


图 7-17. Efficiency

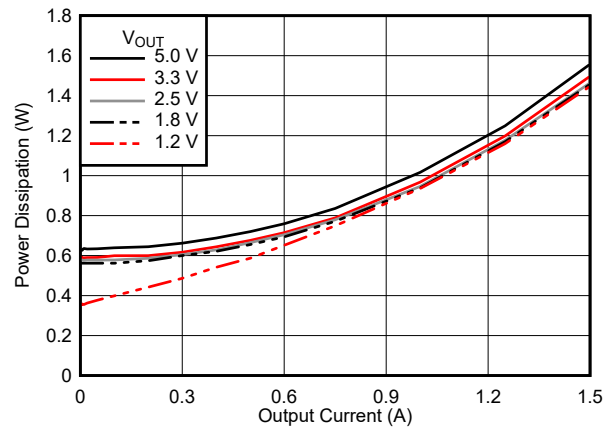
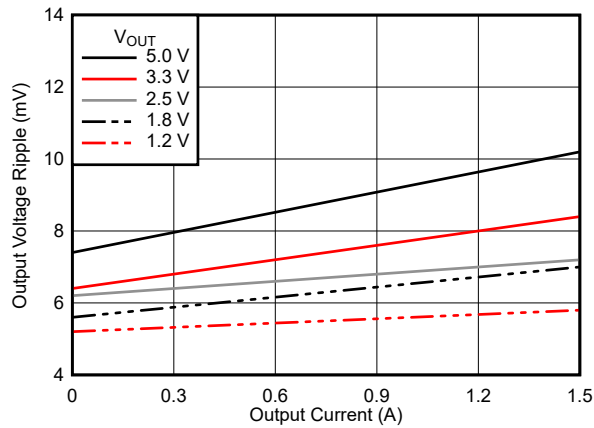
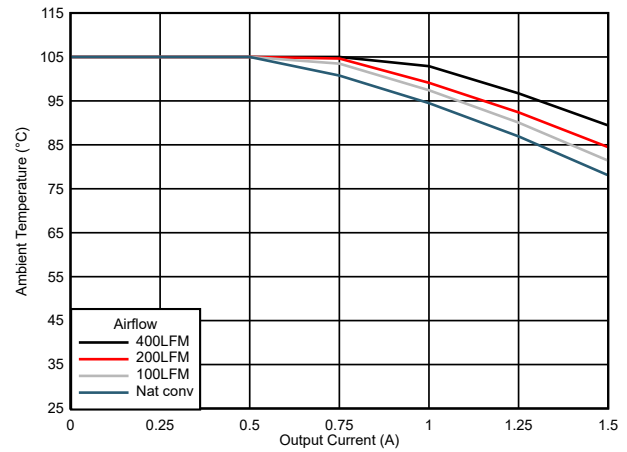


图 7-18. Power Dissipation



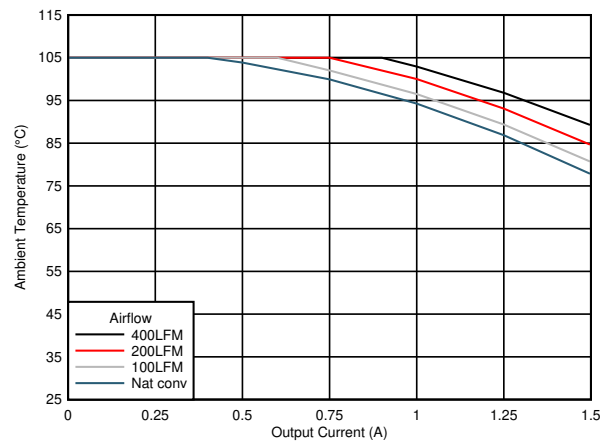
$C_{OUT} = 4 \times 100\text{-}\mu\text{F}$ , 10-V, ceramic

图 7-19. Output Voltage Ripple



Device soldered to a 63.5-mm  $\times$  82.5-mm, 4-layer PCB

图 7-20. Safe Operating Area ( $V_{OUT} = 3.3\text{ V}$ )



Device soldered to a 63.5-mm  $\times$  82.5-mm, 4-layer PCB

图 7-21. Safe Operating Area ( $V_{OUT} = 5.0\text{ V}$ )

## 8 Detailed Description

### 8.1 Overview

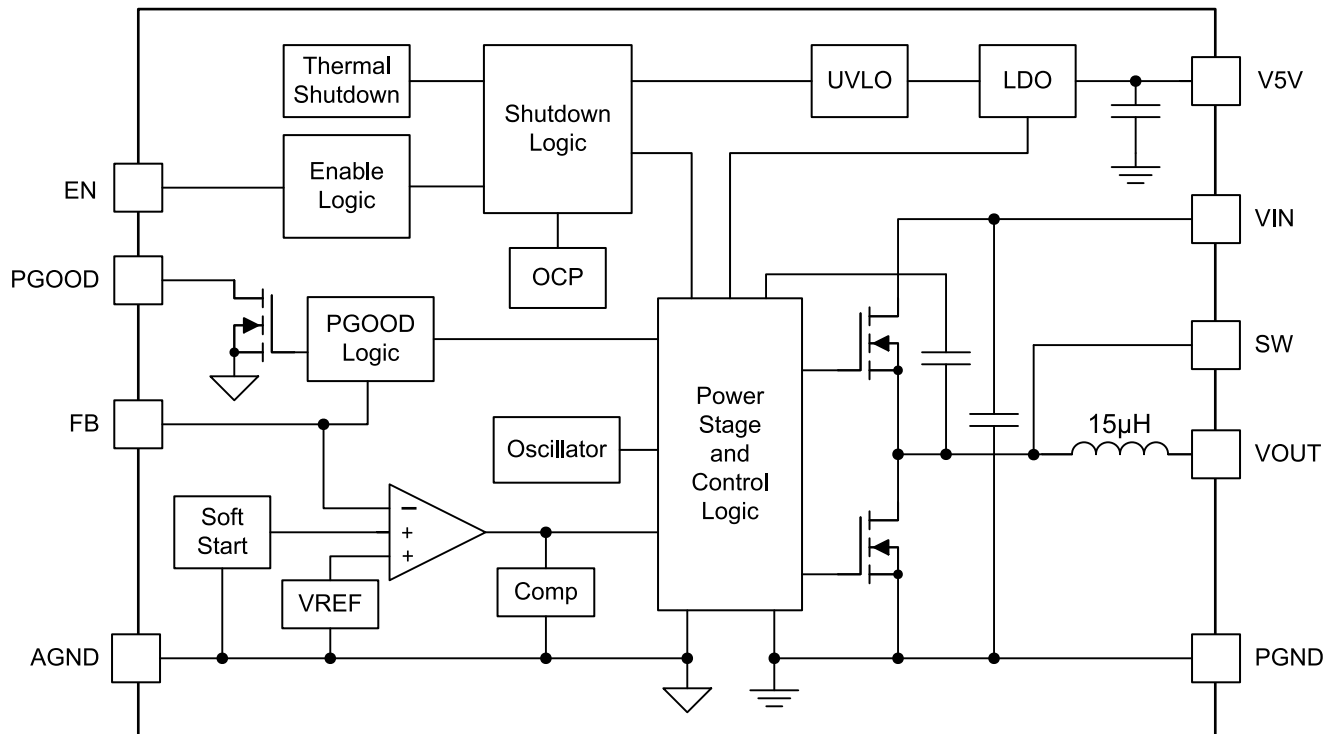
The TPSM5601R5 converter is an easy-to-use, synchronous buck, DC-DC power module that operates from a 4.2-V to 60-V supply voltage. The device is intended for step-down conversions from 5-V, 12-V, 24-V, and 48-V unregulated, semi-regulated, or fully-regulated supply rails. With an integrated power controller, inductor, and MOSFETs, the TPSM5601R5 delivers up to 1.5-A DC load current, with high efficiency and ultra-low input quiescent current, in a very small solution size. Although designed for simple implementation, this device offers flexibility to optimize its usage according to the target application. Control-loop compensation is not required, reducing design time and external component count.

The TPSM5601R5 incorporates several features for comprehensive system requirements, including the following:

- Open-drain power-good circuit for power-rail sequencing and fault reporting
- Monotonic start-up into prebiased loads
- Precision enable with customizable hysteresis for programmable line undervoltage lockout (UVLO)
- Overcurrent and thermal shutdown with automatic recovery

Additionally, the TPSM5601R5S offers frequency spread-spectrum operation. These features enable a flexible and easy-to-use platform for a wide range of applications. The pin arrangement is designed for simple PCB layout, requiring as few as four external components.

### 8.2 Functional Block Diagram



## 8.3 Feature Description

### 8.3.1 Adjustable Output Voltage (FB)

The TPSM5601R5 has an adjustable output voltage range of 1.0 V to 6 V. Setting the output voltage requires two resistors,  $R_{FBT}$  and  $R_{FBB}$  (see 图 8-1). Connect  $R_{FBT}$  between VOUT, at the regulation point, and the FB pin. Connect  $R_{FBB}$  between the FB pin and AGND (pin 10). The recommended value of  $R_{FBT}$  is 10 k $\Omega$ . The value for  $R_{FBB}$  can be calculated using 方程式 1.

$$R_{FBB} = \frac{1.0}{V_{OUT} - 1.0} \times R_{FBT} \quad (1)$$

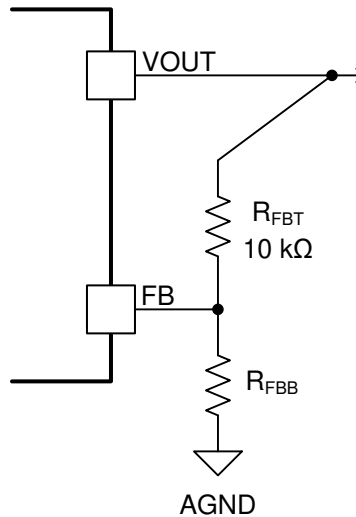


图 8-1. FB Resistor Divider

表 8-1. Standard  $R_{FBB}$  Values

VOUT (V)	$R_{FBB}$ (k $\Omega$ ) <sup>(1)</sup>
1.0	open
1.2	49.9
1.5	20.0
1.8	12.4
2.0	10.0
2.5	6.65
3.0	4.99
3.3	4.32
5.0	2.49
6.0	2.0

(1)  $R_{FBT} = 10$  k $\Omega$

Select an  $R_{FBT}$  value of 10 k $\Omega$  for most applications. A larger  $R_{FBT}$  consumes less DC current, which is mandatory if light-load efficiency is critical. However,  $R_{FBT}$  larger than 1 M $\Omega$  is not recommended as the feedback path becomes more susceptible to noise. High feedback resistance generally requires more careful layout of the feedback path. It is important to keep the feedback trace as short as possible while keeping the feedback trace away from the noisy area of the PCB. For more layout recommendations, see 节 11.

### 8.3.2 Minimum Input Capacitance

The TPSM5601R5 requires a minimum input capacitance of  $9.4 \mu\text{F}$  ( $2 \times 4.7 \mu\text{F}$ ) of ceramic type. High-quality, ceramic-type X5R or X7R capacitors with sufficient voltage rating are required. Place the input capacitors, as close as possible to both VIN pins of the device, between VIN and PGND as shown in 节 11.1. Applications with transient load requirements can benefit from adding additional bulk capacitance to the input as well.

### 8.3.3 Minimum Output Capacitance

The TPSM5601R5 requires a minimum amount of ceramic output capacitance for stability depending on the output voltage setting. 图 8-2 shows the amount of required output capacitance, which is the amount of *effective* capacitance. The effects of DC bias and temperature variation must be considered when using ceramic capacitance. For ceramic capacitors, the package size, voltage rating, and dielectric material contribute to differences between the standard rated value and the actual effective value of the capacitance. Additional output capacitance above the minimum can be added to reduce output voltage ripple and to improve transient response. When adding additional capacitance above the minimum, the capacitance can be ceramic type, low-ESR polymer type, or a combination of the two.

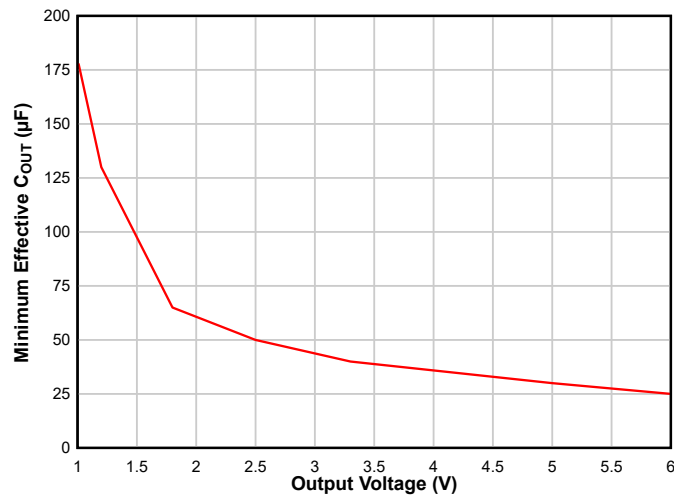


图 8-2. Minimum Required Output Capacitance

### 8.3.4 Precision Enable (EN), Undervoltage Lockout (UVLO), and Hysteresis (HYS)

The EN pin provides precision ON and OFF control for the TPSM5601R5. Once the EN pin voltage exceeds the threshold voltage, the device starts operation. The simplest way to enable the device is to connect EN directly to VIN. This allows the device to start up when  $V_{IN}$  is within its valid operating range. An external logic signal can also be used to drive the EN input to toggle the output on and off and for system sequencing or protection. *This input must not be allowed to float.*

The TPSM5601R5 implements internal undervoltage lockout (UVLO) circuitry on the VIN pin. The device is disabled when the VIN pin voltage is below the internal VIN UVLO threshold. The internal VIN UVLO rising threshold is 3.8 V (typical) with a typical hysteresis of 500 mV.

If an application requires a higher UVLO threshold, the EN input supports adjustable UVLO by connecting a resistor divider from VIN to the EN pin. Applying a voltage greater than or equal to 1.14 V causes the device to enter Standby mode, powering the internal LDO, but not producing an output voltage. Increasing the EN voltage to 1.231 V (typical) fully enables the device, allowing it to enter Start-up mode and start the soft-start period. When the EN input is brought below 1.121 V (110-mV hysteresis), the regulator stops running and enters Standby mode. Further decrease in the EN voltage to below 0.3 V completely shuts down the device.

The TPSM5601R5 utilizes a reference-based soft start that prevents output voltage overshoots and large inrush currents as the regulator is starting up. The rise time of the output voltage is about 4 ms.

### 8.3.5 Power Good (PGOOD)

The TPSM5601R5 provides a PGOOD signal to indicate when the output voltage is within regulation. Use the PGOOD signal for output monitoring, fault protection, or start-up sequencing of downstream converters. PGOOD is an open-drain output that requires a pullup resistor to a DC supply not greater than 18 V. V5V or VOUT can be used as the pullup voltage source. Typical range of pullup resistance is 10 k $\Omega$  to 100 k $\Omega$ . If necessary, use a resistor divider to decrease the voltage from a higher voltage pullup rail. If this function is not needed, the PGOOD pin must be grounded.

When the output voltage exceeds 95% (rising) or decreases below 105% (falling) of the setpoint, the internal PGOOD switch turns off and PGOOD can be pulled high by the external pullup. If the FB voltage falls below 93% or rises above 107% of the setpoint, the internal PGOOD switch turns on, and PGOOD is pulled low to indicate that the output voltage is out of regulation.

Note that during initial power up, a delay of about 4 ms (typical) is inserted from the time that EN is asserted to the time that the power-good flag goes high. This delay only occurs during start-up and is not encountered during normal operation of the power-good function.

### 8.3.6 Spread Spectrum Operation

Spread spectrum is a factory option in the TPSM5601R5S variant. The purpose of the spread spectrum is to eliminate peak emissions at specific frequencies by spreading emissions across a wider range of frequencies than a part with fixed frequency operation. In most systems, low frequency conducted emissions from the first few harmonics of the switching frequency can be easily filtered. A more difficult design criterion is reduction of emissions at higher harmonics which fall in the FM band. These harmonics often couple to the environment through electric fields around the switch node. The TPSM5601R5S device with triangular spread spectrum uses a  $\pm 4\%$  spreading rate (typical) with the modulation rate set at 16 kHz (typical). The spread spectrum is only available while the internal clock is free running at its natural frequency. Any of the following conditions override spread spectrum, turning it off:

- At high input voltages/low output voltage ratio when the device operates at minimum on time the internal clock is slowed disabling spread spectrum.
- The clock is slowed during dropout.

### 8.3.7 Overcurrent Protection (OCP)

The TPSM5601R5 is protected from overcurrent conditions using cycle-by-cycle current limiting for overload conditions and Hiccup mode for short circuits. The current is compared every switching cycle to the current limit threshold. During an overcurrent condition, the output voltage decreases.

### 8.3.8 Thermal Shutdown

Thermal shutdown is an integrated self-protection used to limit junction temperature and prevent damage related to overheating. Thermal shutdown turns off the device when the junction temperature exceeds 170°C (typical) to prevent further power dissipation and temperature rise. Junction temperature decreases after shutdown and the TPSM5601R5 restarts when the junction temperature falls to 158°C (typical).

## 8.4 Device Functional Modes

### 8.4.1 Active Mode

The TPSM5601R5 is in Active mode when VIN is above the turn-on threshold and the EN pin voltage is above the EN high threshold. Connect the EN pin to VIN to allow the device to start up when a valid input voltage is applied. This allows self start-up of the TPSM5601R5 when the input voltage is in the operation range of 4.2 V to 60 V. Connecting a resistor divider between VIN, EN, and AGND adjusts the UVLO to delay the turn on until VIN is closer to its regulated voltage.

### 8.4.2 Standby Mode

Start-up and shutdown are controlled by the EN input. This input features precision thresholds, allowing the use of an external voltage divider to provide an adjustable input UVLO. Applying a voltage of greater than or equal to 1.14 V causes the device to enter Standby mode, powering the internal LDO, but not producing an output voltage. Increasing the EN voltage to 1.231 V (typical) fully enables the device, allowing it to enter Start-up mode

and start the soft-start period. When the EN input is brought below 1.121 V (110-mV hysteresis), the regulator stops running and enters Standby mode. Further decrease in the EN voltage to below 0.3 V completely shuts down the device.

#### 8.4.3 Shutdown Mode

The EN pin provides ON and OFF control for the TPSM5601R5. When  $V_{EN}$  is below the EN low threshold, the device is in shutdown mode. Both the internal LDO and the switching regulator are off. The quiescent current in shutdown mode drops to 5  $\mu$ A at  $V_{IN} = 24$  V.



## 9 Applications and Implementation

### 备注

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

### 9.1 Application Information

The TPSM5601R5 only requires a few external components to convert from a wide range of supply voltages to a fixed output voltage. To expedite and streamline the process of designing a TPSM5601R5, WEBENCH® online software is available to generate complete designs, leveraging iterative design procedures and access to comprehensive component databases. The following section describes the design procedure to configure the TPSM5601R5 power module.

As mentioned previously, the TPSM5601R5 also integrates several optional features to meet system design requirements, including precision enable, UVLO, and PGOOD indicator. The application circuit detailed below shows TPSM5601R5 configuration options suitable for several application use cases. Refer to the [TPSM5601R5EVM User's Guide](#) for more detail.

### 9.2 Typical Application

图 9-1 shows the schematic diagram of a 24-V input, 5-V output, 1.5-A converter.

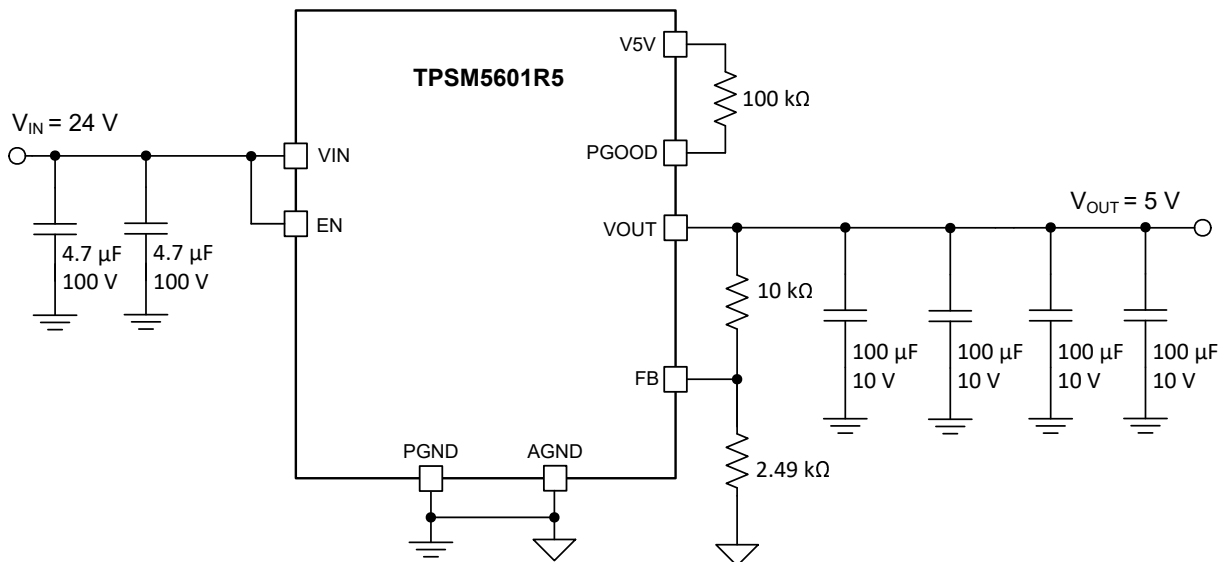


图 9-1. TPSM5601R5 Typical Schematic

#### 9.2.1 Design Requirements

For this design example, use the parameters listed in 表 9-1 as the input parameters and follow the design procedures in 节 9.2.2.

表 9-1. Design Example Parameters

DESIGN PARAMETER	VALUE
Input voltage $V_{IN}$	24 V typical
Output voltage $V_{OUT}$	5 V
Output current rating	1.5 A

## 9.2.2 Detailed Design Procedure

### 9.2.2.1 Custom Design With WEBENCH® Tools

[Click here](#) to create a custom design using the TPSM5601R5 device with the WEBENCH® Power Designer.

1. Start by entering the input voltage ( $V_{IN}$ ), output voltage ( $V_{OUT}$ ), and output current ( $I_{OUT}$ ) requirements.
2. Optimize the design for key parameters such as efficiency, footprint, and cost using the optimizer dial.
3. Compare the generated design with other possible solutions from Texas Instruments.

The WEBENCH Power Designer provides a customized schematic along with a list of materials with real-time pricing and component availability.

In most cases, these actions are available:

- Run electrical simulations to see important waveforms and circuit performance.
- Run thermal simulations to understand board thermal performance.
- Export customized schematic and layout into popular CAD formats.
- Print PDF reports for the design, and share the design with colleagues.

Get more information about WEBENCH tools at [www.ti.com/WBENCH](http://www.ti.com/WBENCH).

### 9.2.2.2 Output Voltage Setpoint

The output voltage of the TPSM5601R5 device is externally adjustable using a resistor divider. The recommended value of  $R_{FBT}$  is 10 k $\Omega$ . The value for  $R_{FBB}$  can be selected from [表 8-1](#) or calculated using [方程式 2](#):

$$R_{FBB} = \frac{1.0}{V_{OUT} - 1.0} \times R_{FBT} \quad (2)$$

For the desired output voltage of 5 V, the formula yields a value of 2.5 k $\Omega$ . Choose the closest available standard value of 2.49 k $\Omega$  for  $R_{FBB}$ .

### 9.2.2.3 Input Capacitors

The TPSM5601R5 requires a minimum input capacitance of 2 × 4.7- $\mu$ F ceramic type. High-quality ceramic type X5R or X7R capacitors with sufficient voltage rating are recommended. The voltage rating of input capacitors must be greater than the maximum input voltage.

For this design, 2 × 4.7- $\mu$ F, 100-V ceramic capacitors are selected.

### 9.2.2.4 Output Capacitor Selection

The TPSM5601R5 requires a minimum amount of output capacitance for proper operation. The minimum amount of required output varies depending on the output voltage. See [图 8-2](#) for the required output capacitance. Additional output capacitance can be added to reduce ripple voltage or for applications with transient load requirements.

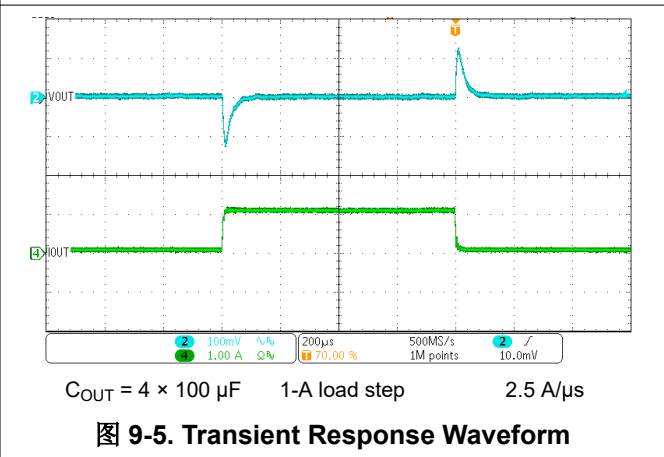
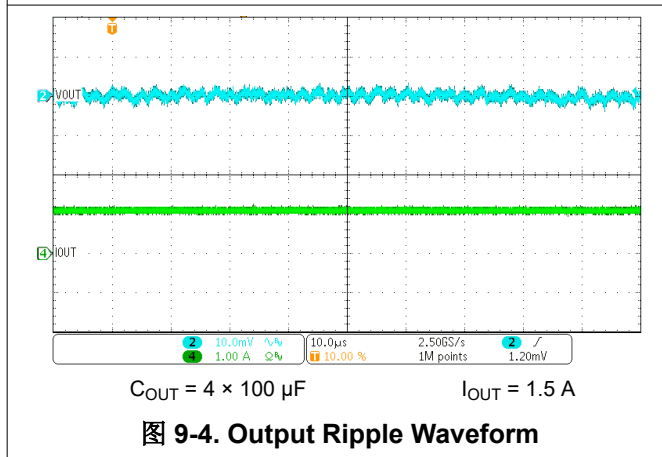
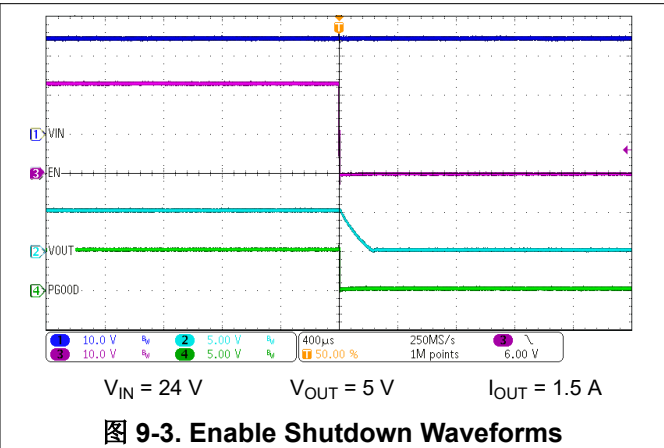
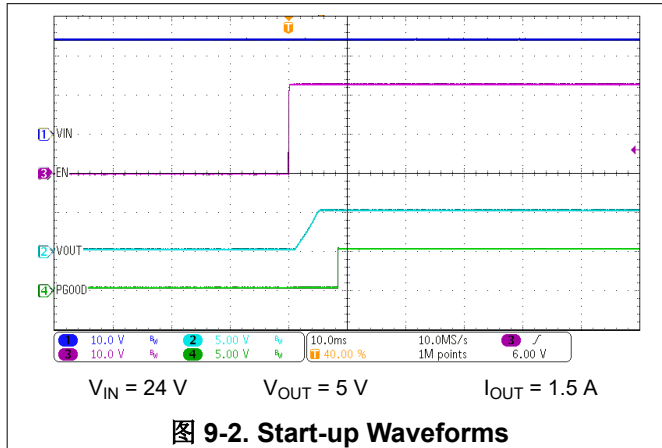
For this design example, 4 × 100- $\mu$ F, 10-V, ceramic capacitors are used.

### 9.2.2.5 Power Good Signal

Applications requiring a power good signal to indicate that the output voltage is present and in regulation must use a pullup resistor between the PGOOD pin and a valid voltage source.

For this design a 100-k $\Omega$  resistor is placed between the PGOOD pin and the V5V pin (the internal 5-V LDO output).

### 9.2.3 Application Curves



## 10 Power Supply Recommendations

The TPSM5601R5 is designed to operate from an input voltage supply range between 4.2 V and 60 V. This input supply must be able to provide the maximum input current and maintain a voltage above the set UVLO voltage. Ensure that the resistance of the input supply rail is low enough that an input current transient does not cause a high enough drop at the TPSM5601R5 supply rail to cause a false UVLO fault triggering and system reset. If the input supply is located more than a few inches from the TPSM5601R5, additional bulk capacitance can be required in addition to the ceramic input capacitance. A 47-µF electrolytic capacitor is a typical choice for this function, whereby the capacitor ESR provides a level of damping against input filter resonances. A typical ESR of 0.5 Ω provides enough damping for most input circuit configurations.

## 11 Layout

The performance of any switching power supply depends as much upon the layout of the PCB as the component selection. Use the following guidelines to design a PCB with the best power conversion performance, optimal thermal performance, and minimal generation of unwanted EMI.

### 11.1 Layout Guidelines

To achieve optimal electrical and thermal performance, an optimized PCB layout is required. 图 11-1 and 图 11-2 show a typical PCB layout. Some considerations for an optimized layout are:

- Use large copper areas for power planes (VIN, VOUT, and PGND) to minimize conduction loss and thermal stress.
- *Connect all PGND pins together using copper plane.*
- Connect AGND pin to the PGND copper at a single point near the pin.
- Place ceramic input and output capacitors close to the device pins to minimize high frequency noise.
- Locate additional output capacitors between the ceramic capacitor and the load.
- Place  $R_{FBT}$  and  $R_{FBB}$  as close as possible to their respective pins.
- Use multiple vias to connect the power planes to internal layers.

### 11.2 Layout Example

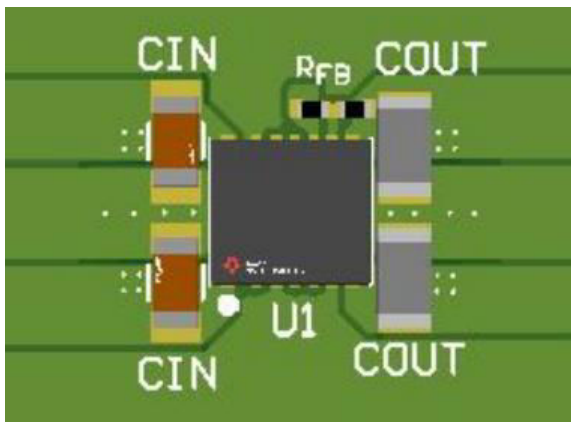


图 11-1. Typical Layout

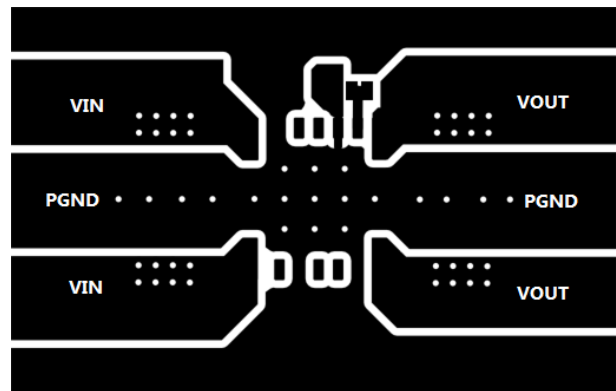


图 11-2. Typical Top-Layer

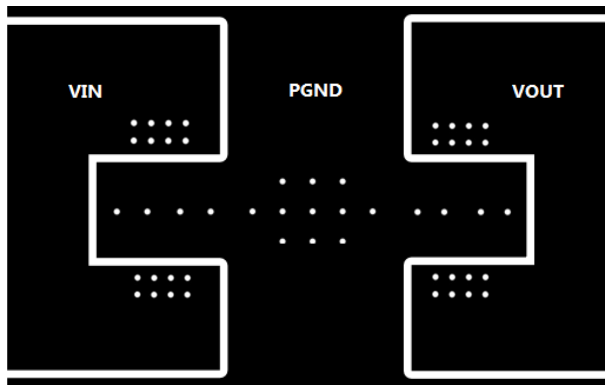


图 11-3. Typical Mid-Layer

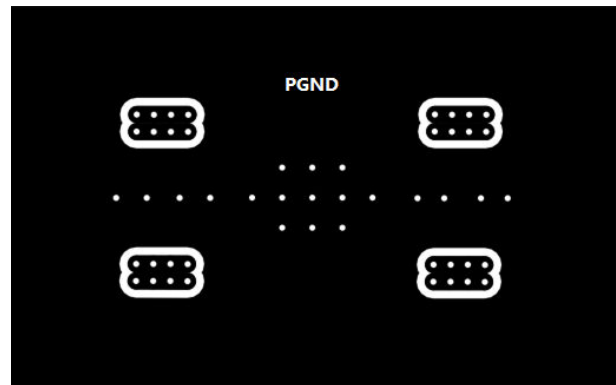


图 11-4. Typical PGND-Layer

### 11.2.1 Theta JA Versus PCB Area

The amount of PCB copper as well as airflow effects the thermal performance of the device. 图 11-5 shows the effects of copper area and airflow on the junction-to-ambient thermal resistance ( $R_{\theta JA}$ ) of the TPSM5601R5. The junction-to-ambient thermal resistance versus PCB area is plotted for a 4-layer PCB.

To determine the required copper area for an application:

1. Determine the maximum power dissipation of the device in the application by referencing the power dissipation graphs in the *Typical Characteristics*.
2. Calculate the maximum  $\theta_{JA}$  using 方程式 3 and the maximum ambient temperature of the application.

$$\theta_{JA} = \frac{(125^{\circ}\text{C} - T_{A(\text{max})})}{P_{D(\text{max})}} \quad (^{\circ}\text{C}/\text{W}) \quad (3)$$

3. Reference 图 11-5 to determine the minimum required PCB area for the application conditions.

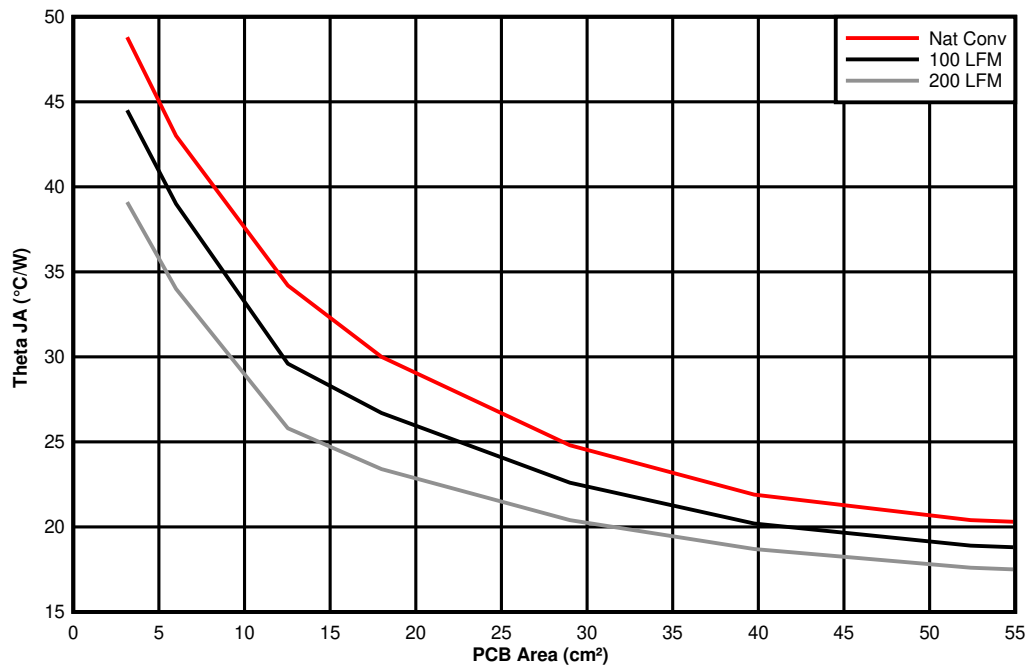


图 11-5.  $\theta_{JA}$  vs PCB Area

### 11.2.2 Package Specifications

表 11-1. Package Specifications Table

TPSM5601R5		VALUE	UNIT
Weight		429	mg
Flammability	Meets UL 94 V-O		
MTBF Calculated Reliability	Per Bellcore TR-332, 50% stress, $T_A = 40^{\circ}\text{C}$ , ground benign	87.7	MHrs

### 11.2.3 EMI

The TPSM5601R5 is compliant with EN55011 radiated emissions. 图 11-6 through 图 11-9 show typical examples of radiated emission plots for the TPSM5601R5H (1.5-A, 1-MHz version). Expect similar results for the TPSM5601R5. The graphs include the plots of the antenna in the horizontal and vertical positions.

#### 11.2.3.1 EMI Plots

EMI plots were measured using the standard TPSM5601R5HEVM.

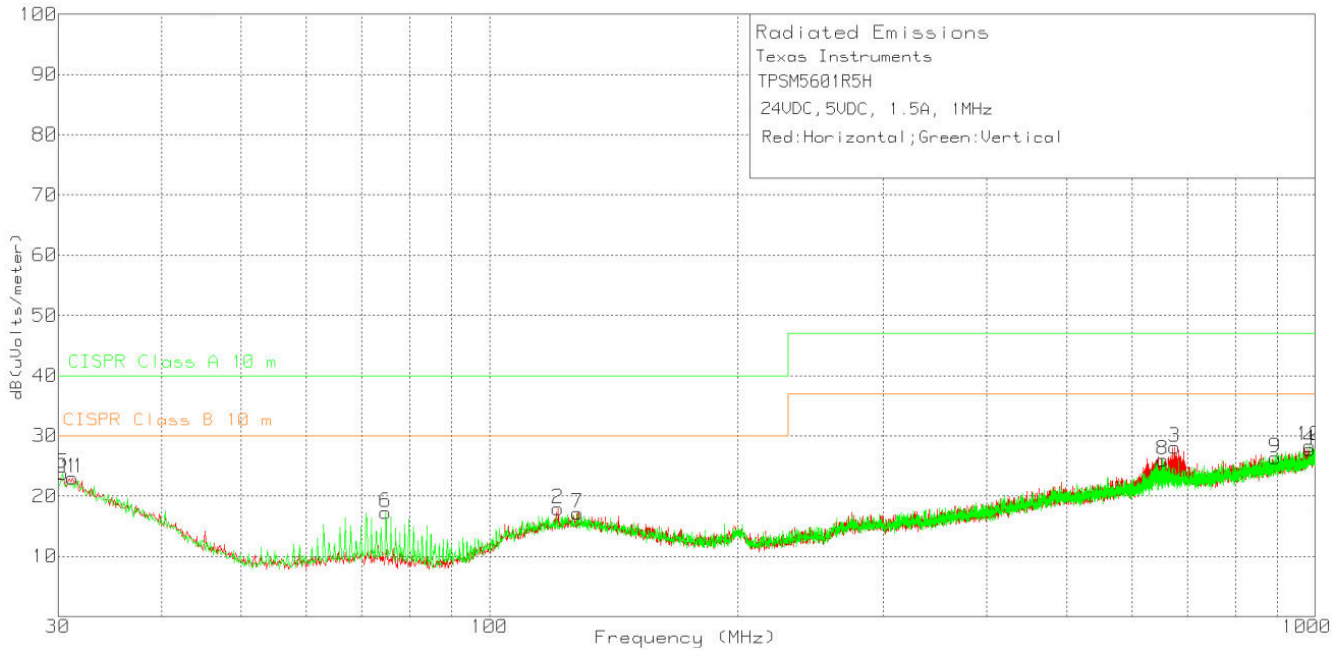


图 11-6. Radiated Emissions 24-V Input, 5-V Output, 1.5-A Load

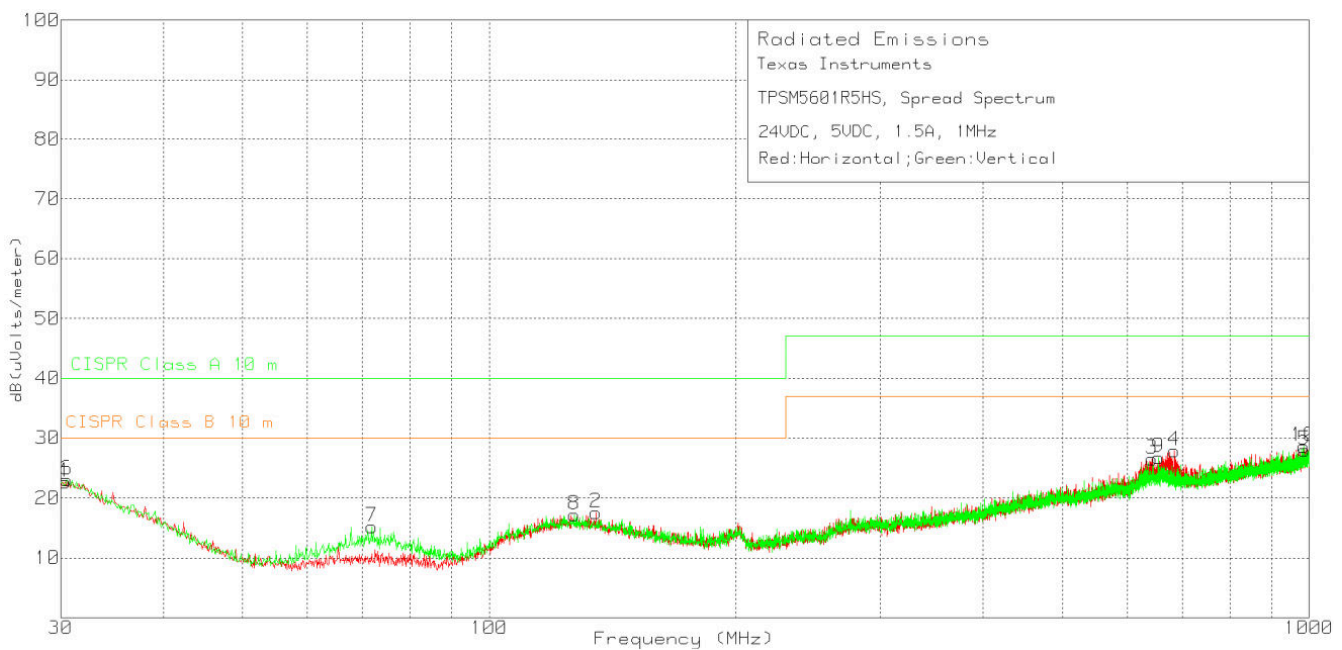


图 11-7. Radiated Emissions 24-V Input, 5-V Output, 1.5-A Load (Spread Spectrum)

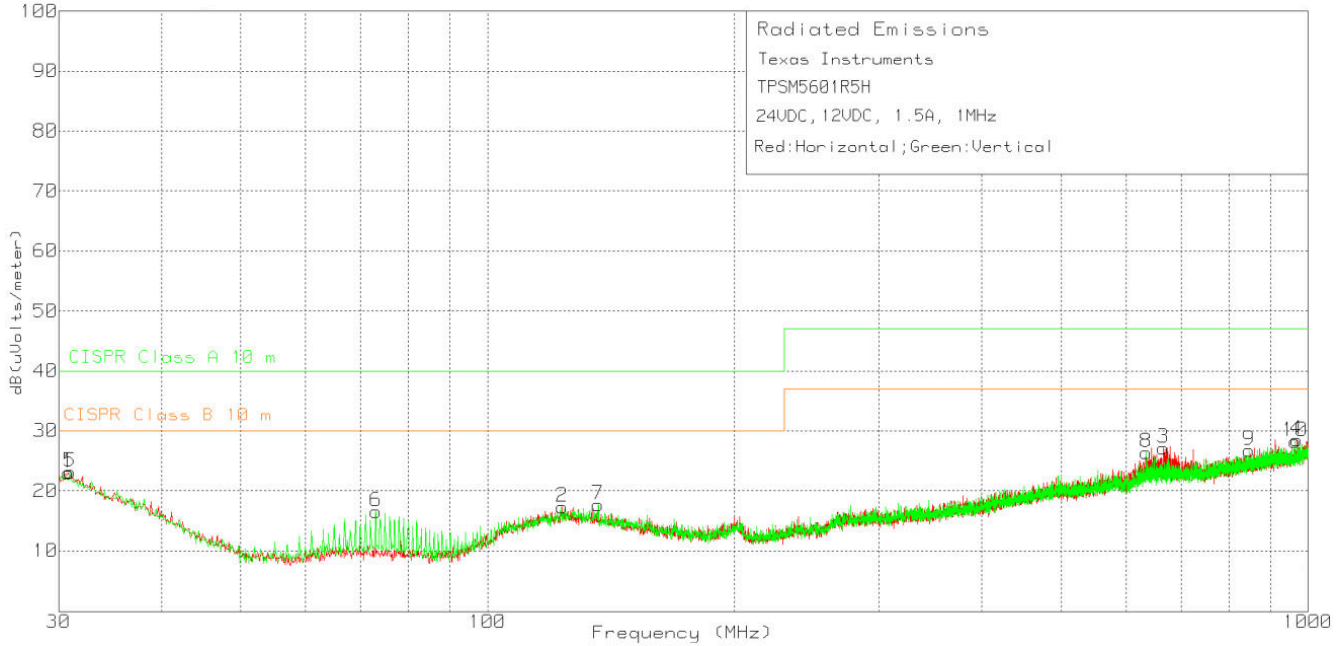


图 11-8. Radiated Emissions 24-V Input, 12-V Output, 1.5-A Load

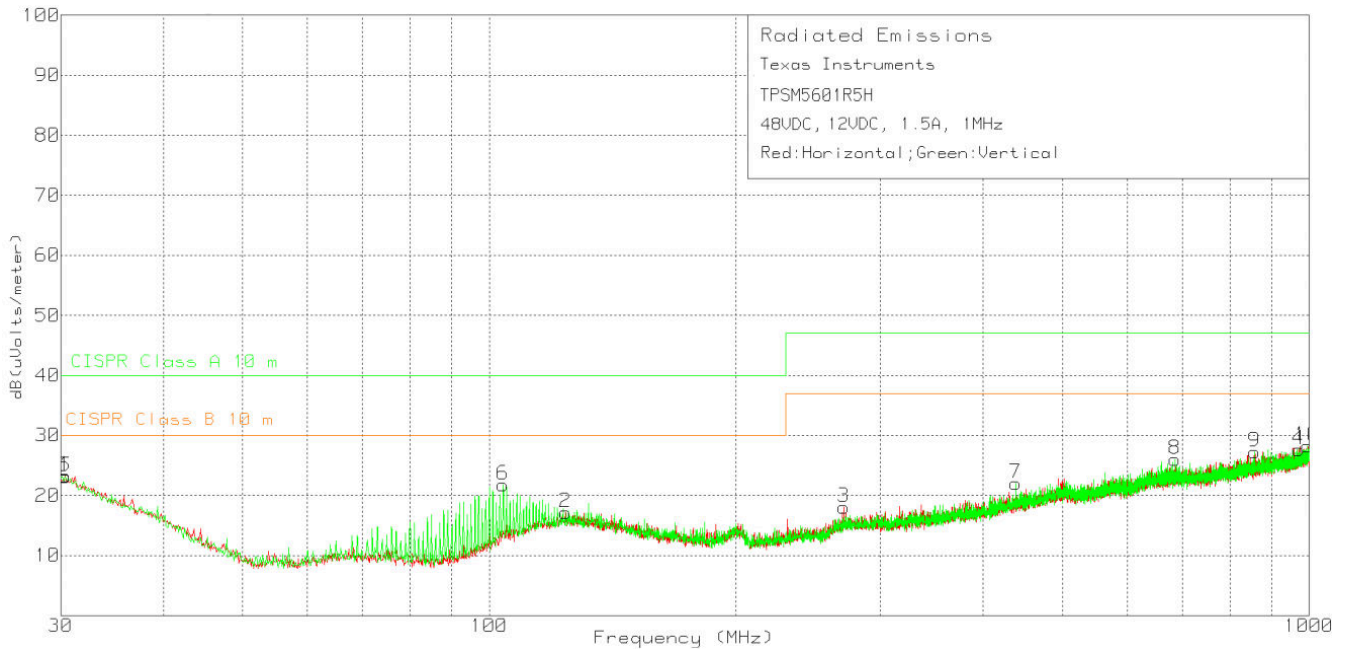


图 11-9. Radiated Emissions 48-V Input, 12-V Output, 1.5-A Load

## 12 Device and Documentation Support

### 12.1 Device Support

#### 12.1.1 第三方产品免责声明

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#### 12.1.2 Development Support

For development support, see the following:

- For TI's reference design library, visit [TI Designs](#).
- To view a related device of this product, see the [TPSM5601R5Hx](#).

##### 12.1.2.1 Custom Design With WEBENCH® Tools

[Click here](#) to create a custom design using the TPSM5601R5 device with WEBENCH® Power Designer.

1. Start by entering the input voltage ( $V_{IN}$ ), output voltage ( $V_{OUT}$ ), and output current ( $I_{OUT}$ ) requirements.
2. Optimize the design for key parameters such as efficiency, footprint, and cost using the optimizer dial.
3. Compare the generated design with other possible solutions from Texas Instruments.

The WEBENCH Power Designer provides a customized schematic along with a list of materials with real-time pricing and component availability.

In most cases, these actions are available:

- Run electrical simulations to see important waveforms and circuit performance.
- Run thermal simulations to understand board thermal performance.
- Export customized schematic and layout into popular CAD formats.
- Print PDF reports for the design, and share the design with colleagues.

Get more information about WEBENCH tools at [www.ti.com/WEBENCH](http://www.ti.com/WEBENCH).

### 12.2 Documentation Support

#### 12.2.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, [TPSM5601R5EVM User's Guide](#)
- Texas Instruments, [Using the TPSM5601R5Hx in an Inverting Buck-Boost Topology Application Report](#)
- Texas Instruments, [Using New Thermal Metrics Application Report](#)
- Texas Instruments, [Semiconductor and IC Package Thermal Metrics Application Report](#)

### 12.3 接收文档更新通知

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### 12.4 支持资源

[TI E2E™ 支持论坛](#) 是工程师的重要参考资料，可直接从专家获得快速、经过验证的解答和设计帮助。搜索现有解答或提出自己的问题可获得所需的快速设计帮助。

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### 12.5 Trademarks

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## 12.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

## 12.7 术语表

[TI 术语表](#) 本术语表列出并解释了术语、首字母缩略词和定义。

## 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this datasheet, refer to the left-hand navigation.

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**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPSM5601R5RDAR	ACTIVE	B3QFN	RDA	15	1000	RoHS & Green	NIPDAU	Level-3-245C-168 HR	-40 to 125	5601R5	<a href="#">Samples</a>
TPSM5601R5SRDAR	ACTIVE	B3QFN	RDA	15	1000	RoHS & Green	NIPDAU	Level-3-245C-168 HR	-40 to 125	5601R5S	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

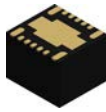
(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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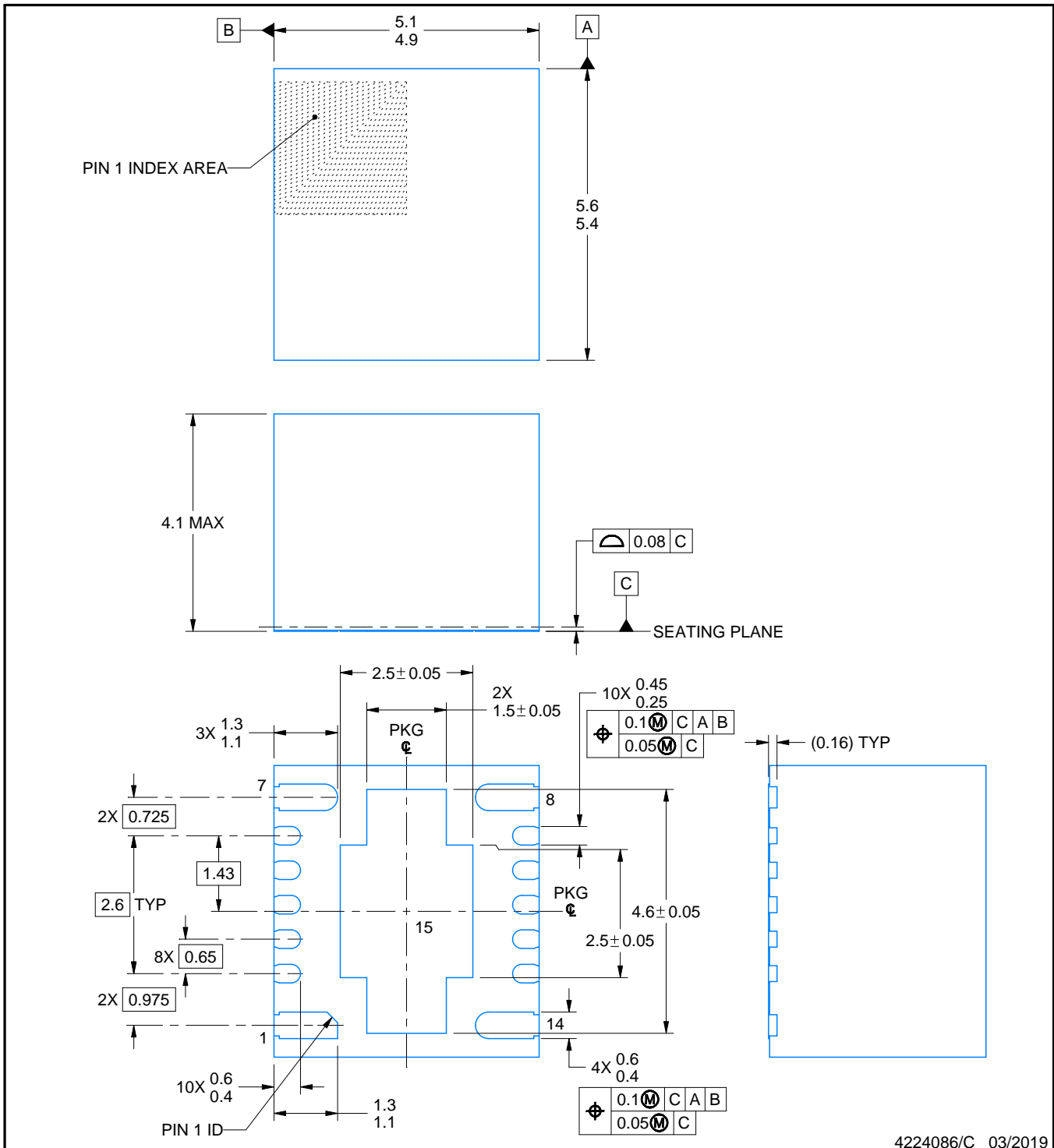


# PACKAGE OUTLINE

RDA0015A

B3QFN - 4.1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



4224086/C 03/2019

## NOTES:

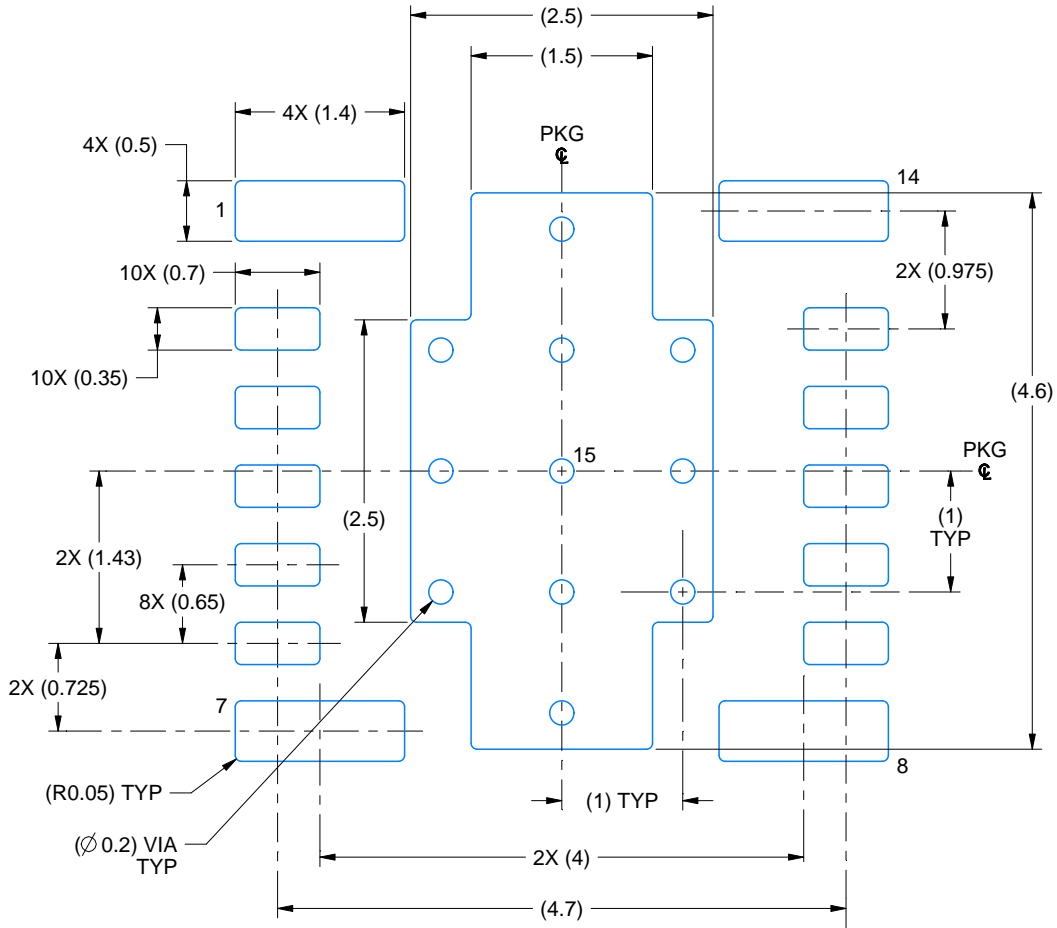
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.

# EXAMPLE BOARD LAYOUT

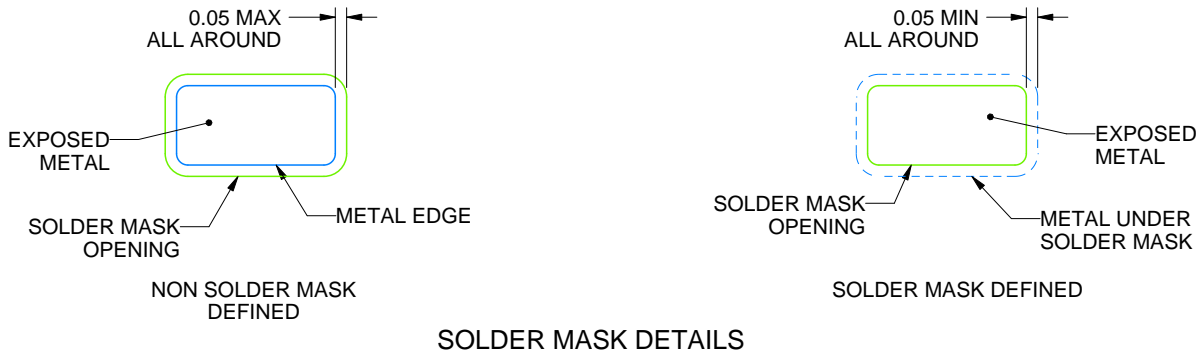
RDA0015A

B3QFN - 4.1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE: 16X



SOLDER MASK DETAILS

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NOTES: (continued)

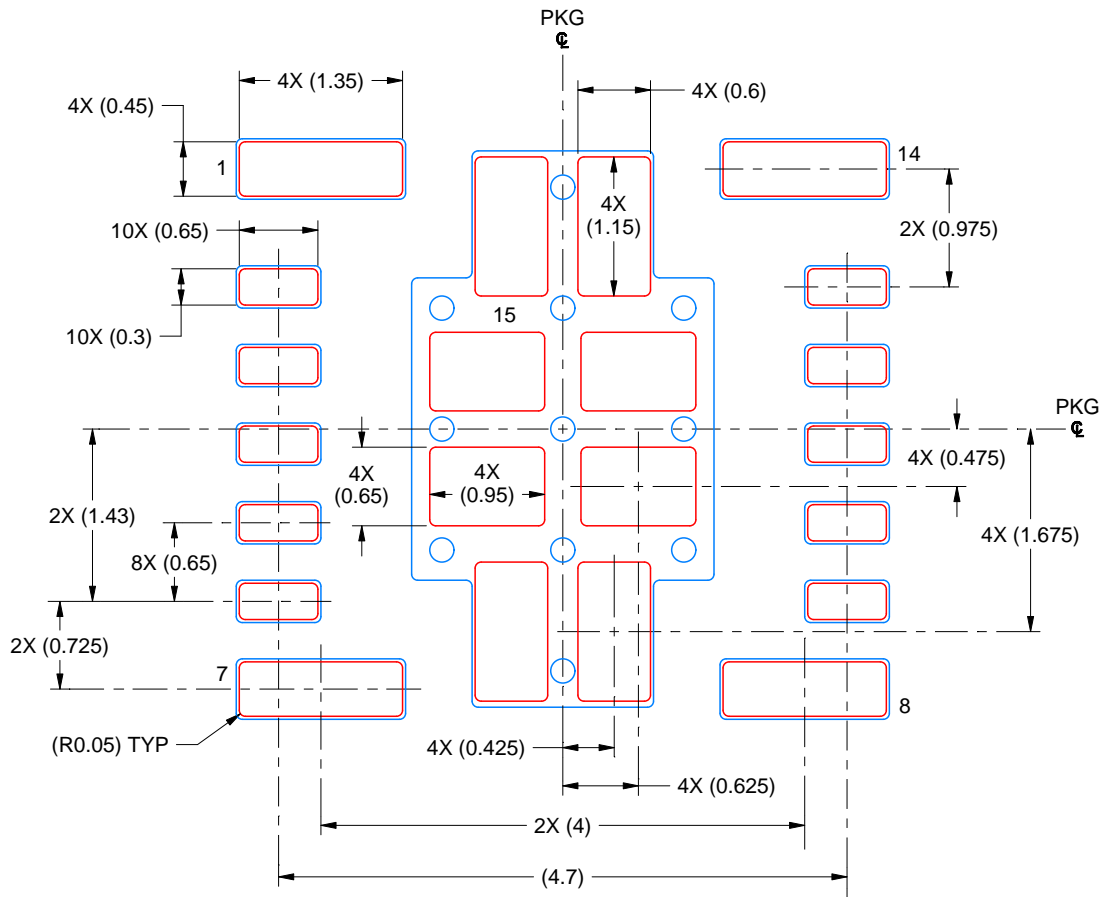
- This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 ([www.ti.com/lit/slua271](http://www.ti.com/lit/slua271)).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

# EXAMPLE STENCIL DESIGN

RDA0015A

B3QFN - 4.1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD 15:  
56% PRINTED SOLDER COVERAGE BY AREA  
SCALE: 16X

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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