

## 120V 启动电压、4A 峰值电流的高频高端/低端驱动器

查询样品: [UCC27210](#), [UCC27211](#)

### 特性

- 可驱动两个高端/低端配置中的 **N**通道 **MOSFET** 并具有独立输入
- 最大启动电压: **120V DC**
- **4A** 吸收、**4A** 供电输出电流
- **0.9-Ω** 上拉/下拉电阻
- 输入引脚电压范围 **-10 V** 至 **20 V** 并且独立于电源电压之外
- **TTL** 或者伪-**CMOS** 兼容输入版本
- **8-V** 至 **17-V VDD** 运行范围, (最大值 **20 V ABS**)
- **7.2 ns** 上升和 **5.5 ns** 下降时间 (**1000pF** 负载时)
- 快速传播延迟(典型值 **18-ns**)
- **2ns** 延迟匹配
- 用于高端和低端驱动器的对称欠压闭锁功能
- 采用全工业标准 (**SOIC-8**, **PowerPAD™** **SOIC-8**, **4-mm x 4-mm SON-8** 和 **4-mm x 4-mm SON-10**)封装
- 额定温度范围 **-40** 至 **140 °C**

### 应用范围

- 针对电信、数据通信和商业市场的电源
- 半桥式和全桥式转换器
- 推-拉转换器
- 高电压同步降压型转换器
- 两开关正激式转换器
- 有源箝位正激式转换器
- **D** 类音频放大器

### 说明

UCC27210 和 UCC27211 驱动器是基于流行的 UCC27200 和 UCC27201 MOSFET 驱动器设计而成的, 但相对于之前的产品, 这两款器件的性能有很大提升。峰值输出上拉和下拉电流已经增加至 **4-A** 源电流/**4-A** 吸收电流, 而上拉/下拉电阻也已经减少到 **0.9 Ω**, 这样的话, 在 MOSFET 的米勒效应平台转换期间, 能够在保证最小切换损失的前提下驱动大功率 MOSFET。此输入结构现在能直接处理 **10 VDC**, 这增加了耐用性并且也在无需使用整流二极管的情况下实现到栅极驱动变压器的直接接口连接。此输入也与电源电压无关并且最大额定电压为 **20-V**。



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English Data Sheet: [SLUSAT7](#)



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

说明 (续)

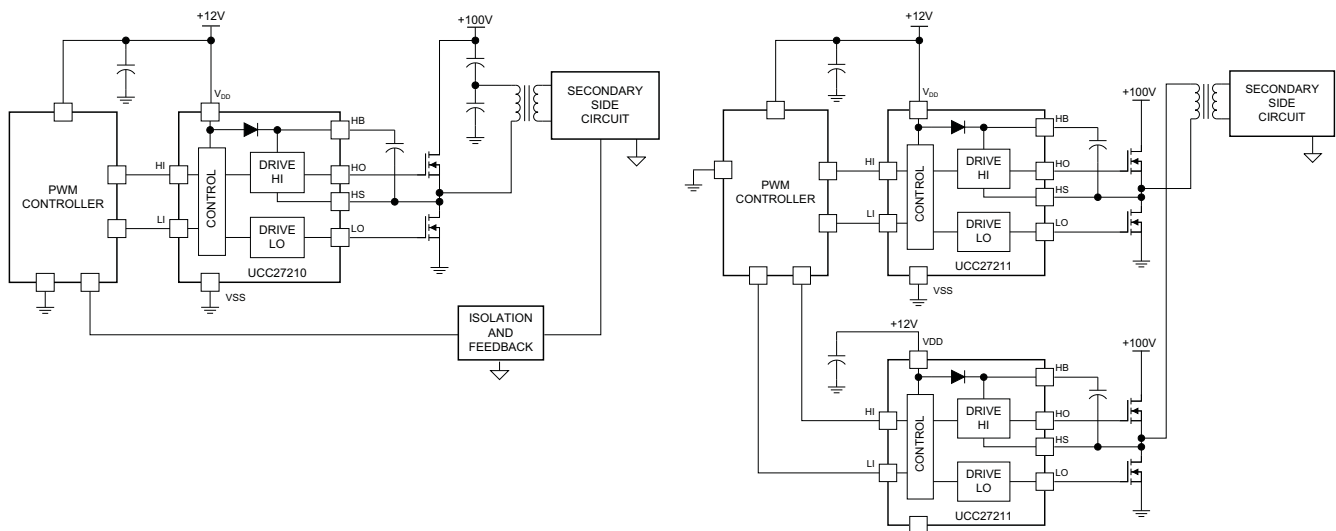
UCC27210/1 的开关结点 (HS 引脚) 能够处理最大-18 V电压, 这样就保护高端通道不受由寄生电感和杂散电容引起的固有负电压带来的损害。UCC27210 (伪-CMOS输入) 和 UCC27211 (TTL 输入) 已经增加了滞后时间, 这样就实现了到具有增强抗噪性能的模拟或者数字PWM控制器的接口连接。

低端和高端栅极驱动器是独立控制的, 它们之间打开和关闭的匹配时间为 2 ns。

片载额定 120-V自举二极管省去了对外部分立式二极管的需要。为高端和低端驱动器提供了欠压闭锁功能, 如果驱动电压低于规定的门限, 则提供对称的打开/关闭性能并将输出强制设为低电平。

这两款器件均采用 8 引脚 SOIC(D)、PowerPad™ SOIC-8 (DDA)、4-mm x 4-mm SON-8(DRM) 和 SON-10 (DPR) 封装。

Typical Application Diagrams



ORDERING INFORMATION (1)

TEMPERATURE RANGE $T_A = T_J$	INPUT COMPATIBILITY	PACKAGED DEVICES <sup>(1)</sup>			
		SOIC-8 (D) <sup>(2)</sup>	PowerPAD™ SOIC-8 (DDA) <sup>(2)</sup>	SON-8 (DRM) <sup>(3)</sup>	SON-10 (DPR) <sup>(4)</sup>
-40°C to 140°C	Pseudo CMOS	UCC27210D	UCC27210DDA	UCC27210DRM	UCC27210DPR
	TTL	UCC27211D	UCC27211DDA	UCC27211DRM	UCC27211DPR

- (1) These products are packaged in Lead (Pb)-Free and green lead finish of PdNiAu which is compatible with MSL level 1 at 255°C to 260°C peak reflow temperature to be compatible with either lead free or Sn/Pb soldering operations.
- (2) D (SOIC-8) and DDA (Power Pad™ SOIC-8) packages are available taped and reeled. Add R suffix to device type (e.g. UCC27210ADR/UCC27211ADR) to order quantities of 2,500 devices per reel.
- (3) DRM (SON-8) package comes either in a small reel of 250 pieces as part number UCC27210ADRMT/UCC27211ADRMT, or larger reels of 3000 pieces as part number UCC27210ADRMR/UCC27211ADRMR.
- (4) DPR (SON-10) package comes either in a small reel of 250 pieces as part number UCC27210ADPRT/UCC27211ADPRT, or large reels of 3000 pieces as part number UCC27210ADPRR/UCC27211ADPRR.

## ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
Supply voltage range, $V_{DD}^{(1)}$ , $V_{HB} - V_{HS}$		-0.3	20	V
Input voltages on LI and HI, $V_{LI}$ , $V_{HI}$		-10	20	
Output voltage on LO, $V_{LO}$	DC	-0.3	$V_{DD} + 0.3$	
	Repetitive pulse <100 ns <sup>(2)</sup>	-2	$V_{DD} + 0.3$	
Output voltage on HO, $V_{HO}$	DC	$V_{HS} - 0.3$	$V_{HB} + 0.3$	
	Repetitive pulse <100 ns <sup>(2)</sup>	$V_{HS} - 2$	$V_{HB} + 0.3$	
Voltage on HS, $V_{HS}$	DC	-1	115	
	Repetitive pulse <100 ns <sup>(2)</sup>	-18	115	
Voltage on HB, $V_{HB}$		-0.3	120	
ESD	Human Body Model (HBM)		2	
	Field Induced Charged Device Model (FICDM)		1	
Operating virtual junction temperature range, $T_J$		-40	150	°C
Storage temperature, $T_{STG}$		-65	150	
Lead temperature (soldering, 10 sec.)			300	

(1) All voltages are with respect to VSS unless otherwise noted. Currents are positive into, negative out of the specified terminal.

(2) Verified at bench characterization.

## RECOMMENDED OPERATING CONDITIONS

all voltages are with respect to  $V_{SS}$ ; currents are positive into and negative out of the specified terminal.  $-40^{\circ}\text{C} < T_J = T_A < 140^{\circ}\text{C}$  (unless otherwise noted)

PARAMETER	MIN	TYP	MAX	UNIT
Supply voltage range, $V_{DD}$ , $V_{HB}-V_{HS}$	8	12	17	V
Voltage on HS, $V_{HS}$	-1		105	
Voltage on HS, $V_{HS}$ (repetitive pulse <100 ns)	-15		110	
Voltage on HB, $V_{HB}$	$V_{HS} + 8$ , $V_{DD} - 1$		$V_{HS} + 17$ , 115	
Voltage slew rate on HS			50	V/ns
Operating junction temperature range	-40		140	°C

**THERMAL INFORMATION**

THERMAL METRIC		UCC27210/11 <sup>(1)</sup>		UNITS
		D	DDA	
		8 PINS	8 PINS	
$\theta_{JA}$	Junction-to-ambient thermal resistance <sup>(2)</sup>	111.8	37.7	°C/W
$\theta_{JCTop}$	Junction-to-case (top) thermal resistance <sup>(3)</sup>	56.9	47.2	
$\theta_{JB}$	Junction-to-board thermal resistance <sup>(4)</sup>	53.0	9.6	
$\psi_{JT}$	Junction-to-top characterization parameter <sup>(5)</sup>	7.8	2.8	
$\psi_{JB}$	Junction-to-board characterization parameter <sup>(6)</sup>	52.3	9.4	
$\theta_{JCbott}$	Junction-to-case (bottom) thermal resistance <sup>(7)</sup>	n/a	3.6	

- (1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).
- (2) The junction-to-ambient thermal resistance under natural convection is obtained in a simulation on a JEDEC-standard, high-K board, as specified in JESD51-7, in an environment described in JESD51-2a.
- (3) The junction-to-case (top) thermal resistance is obtained by simulating a cold plate test on the package top. No specific JEDEC-standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.
- (4) The junction-to-board thermal resistance is obtained by simulating in an environment with a ring cold plate fixture to control the PCB temperature, as described in JESD51-8.
- (5) The junction-to-top characterization parameter,  $\psi_{JT}$ , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining  $\theta_{JA}$ , using a procedure described in JESD51-2a (sections 6 and 7).
- (6) The junction-to-board characterization parameter,  $\psi_{JB}$ , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining  $\theta_{JA}$ , using a procedure described in JESD51-2a (sections 6 and 7).
- (7) The junction-to-case (bottom) thermal resistance is obtained by simulating a cold plate test on the exposed (power) pad. No specific JEDEC standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

**THERMAL INFORMATION**

THERMAL METRIC		UCC27210/11 <sup>(1)</sup>		UNITS
		DRM	DPR	
		8 PINS	10 PINS	
$\theta_{JA}$	Junction-to-ambient thermal resistance <sup>(2)</sup>	33.9	36.8	°C/W
$\theta_{JCTop}$	Junction-to-case (top) thermal resistance <sup>(3)</sup>	33.2	36.0	
$\theta_{JB}$	Junction-to-board thermal resistance <sup>(4)</sup>	11.4	14.0	
$\psi_{JT}$	Junction-to-top characterization parameter <sup>(5)</sup>	0.4	0.3	
$\psi_{JB}$	Junction-to-board characterization parameter <sup>(6)</sup>	11.7	14.2	
$\theta_{JCbott}$	Junction-to-case (bottom) thermal resistance <sup>(7)</sup>	2.3	3.4	

- (1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).
- (2) The junction-to-ambient thermal resistance under natural convection is obtained in a simulation on a JEDEC-standard, high-K board, as specified in JESD51-7, in an environment described in JESD51-2a.
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- (7) The junction-to-case (bottom) thermal resistance is obtained by simulating a cold plate test on the exposed (power) pad. No specific JEDEC standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

## ELECTRICAL CHARACTERISTICS

 $V_{DD} = V_{HB} = 12\text{ V}$ ,  $V_{HS} = V_{SS} = 0\text{ V}$ , no load on LO or HO,  $T_A = T_J = -40^\circ\text{C}$  to  $140^\circ\text{C}$ , (unless otherwise noted)

PARAMETER		TEST CONDITION	MIN	TYP	MAX	UNITS	
<b>Supply Currents</b>							
$I_{DD}$	$V_{DD}$ quiescent current	$V(\text{LI}) = V(\text{HI}) = 0\text{ V}$	0.05	0.085	0.17	mA	
$I_{DDO}$	$V_{DD}$ operating current	UCC27210	2.4	2.6	4.3		
		UCC27211	2.4	2.5	4.3		
$I_{HB}$	Boot voltage quiescent current	$V(\text{LI}) = V(\text{HI}) = 0\text{ V}$	0.015	0.065	0.1		
$I_{HBO}$	Boot voltage operating current	$f = 500\text{ kHz}$ , $C_{\text{LOAD}} = 0$	1.5	2.5	4		
$I_{HBS}$	HB to $V_{SS}$ quiescent current	$V(\text{HS}) = V(\text{HB}) = 115\text{ V}$		0.0005	0.13	$\mu\text{A}$	
$I_{HBSO}$	HB to $V_{SS}$ operating current	$f = 500\text{ kHz}$ , $C_{\text{LOAD}} = 0$		0.07	0.9	mA	
<b>Input</b>							
$V_{\text{HIT}}$	Input voltage threshold	UCC27210	4.2	5.0	5.8	V	
$V_{\text{LIT}}$	Input voltage threshold		2.4	3.2	4.0		
$V_{\text{IHYS}}$	Input voltage hysteresis		1.8				
$R_{\text{IN}}$	Input pull-down resistance		102				k $\Omega$
$V_{\text{HIT}}$	Input voltage threshold	UCC27211	1.9	2.3	2.7	V	
$V_{\text{LIT}}$	Input voltage threshold		1.3	1.6	1.9		
$V_{\text{IHYS}}$	Input voltage hysteresis		700				mV
$R_{\text{IN}}$	Input pull-down resistance		68				k $\Omega$
<b>Under-Voltage Lockout (UVLO)</b>							
$V_{\text{DDR}}$	$V_{DD}$ turn-on threshold		6.2	7.0	7.8	V	
$V_{\text{DDHYS}}$	Hysteresis			0.5			
$V_{\text{HBR}}$	$V_{\text{HB}}$ turn-on threshold		5.6	6.7	7.9		
$V_{\text{HBHYS}}$	Hysteresis			1.1			
<b>Bootstrap Diode</b>							
$V_{\text{F}}$	Low-current forward voltage	$I_{\text{VDD-HB}} = 100\ \mu\text{A}$		0.65	0.8	V	
$V_{\text{FI}}$	High-current forward voltage	$I_{\text{VDD-HB}} = 100\ \text{mA}$		0.85	0.95		
$R_{\text{D}}$	Dynamic resistance, $\Delta V_{\text{F}}/\Delta I$	$I_{\text{VDD-HB}} = 100\ \text{mA}$ and $80\ \text{mA}$	0.3	0.5	0.85	$\Omega$	
<b>LO Gate Driver</b>							
$V_{\text{LOL}}$	Low-level output voltage	$I_{\text{LO}} = 100\ \text{mA}$	0.05	0.09	0.15	V	
$V_{\text{LOH}}$	High level output voltage	$I_{\text{LO}} = -100\ \text{mA}$ , $V_{\text{LOH}} = V_{\text{DD}} - V_{\text{LO}}$	0.1	0.16	0.27		
	Peak pull-up current <sup>(1)</sup>	$V_{\text{LO}} = 0\ \text{V}$		3.7		A	
	Peak pull-down current <sup>(1)</sup>	$V_{\text{LO}} = 12\ \text{V}$		4.5			
<b>HO GATE Driver</b>							
$V_{\text{HOL}}$	Low-level output voltage	$I_{\text{HO}} = 100\ \text{mA}$	0.05	0.09	0.15	V	
$V_{\text{HOH}}$	High-level output voltage	$I_{\text{HO}} = -100\ \text{mA}$ , $V_{\text{HOH}} = V_{\text{HB}} - V_{\text{HO}}$	0.1	0.16	0.27		
	Peak pull-up current <sup>(1)</sup>	$V_{\text{HO}} = 0\ \text{V}$		3.7		A	
	Peak pull-down current <sup>(1)</sup>	$V_{\text{HO}} = 12\ \text{V}$		4.5			

(1) Ensured by design.

**ELECTRICAL CHARACTERISTICS (continued)**

$V_{DD} = V_{HB} = 12\text{ V}$ ,  $V_{HS} = V_{SS} = 0\text{ V}$ , no load on LO or HO,  $T_A = T_J = -40^\circ\text{C}$  to  $140^\circ\text{C}$ , (unless otherwise noted)

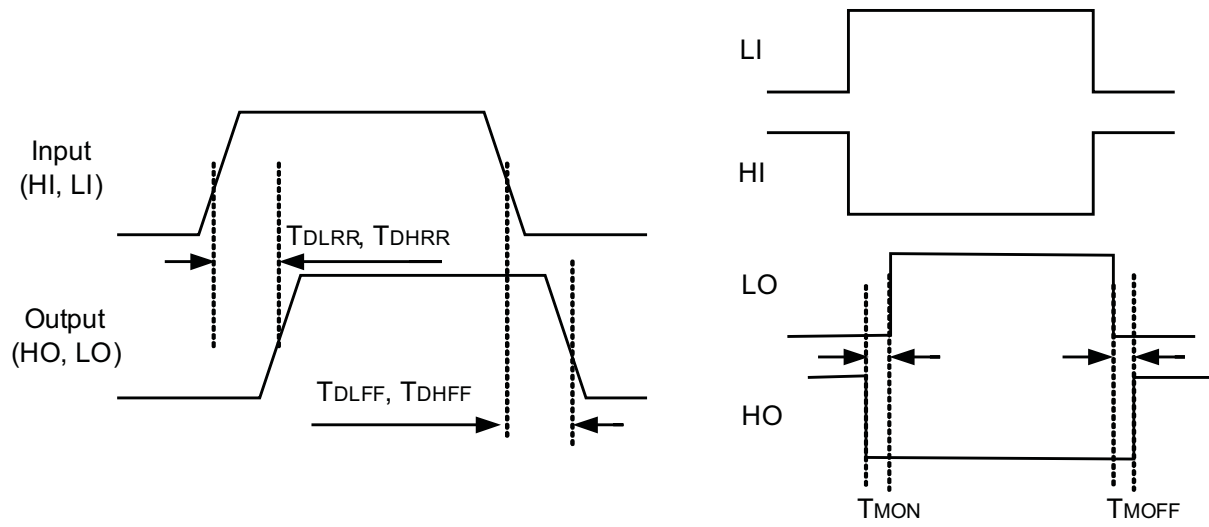
PARAMETER		TEST CONDITION	MIN	TYP	MAX	UNITS	
<b>Switching Parameters: Propagation Delays</b>							
$T_{DLFF}$	$V_{LI}$ falling to $V_{LO}$ falling	UCC27210, $C_{LOAD} = 0$	17	21	37	ns	
$T_{DHFF}$	$V_{HI}$ falling to $V_{HO}$ falling		17	21	37		
$T_{DLRR}$	$V_{LI}$ rising to $V_{LO}$ rising		18	24	46		
$T_{DHRR}$	$V_{HI}$ rising to $V_{HO}$ rising		18	24	46		
$T_{DLFF}$	$V_{LI}$ falling to $V_{LO}$ falling	UCC27211, $C_{LOAD} = 0$	10	17	30		
$T_{DHFF}$	$V_{HI}$ falling to $V_{HO}$ falling		10	17	30		
$T_{DLRR}$	$V_{LI}$ rising to $V_{LO}$ rising		10	18	40		
$T_{DHRR}$	$V_{HI}$ rising to $V_{HO}$ rising		10	18	40		
<b>Switching Parameters: Delay Matching</b>							
$T_{MON}$	From HO OFF to LO ON	UCC27210	$T_J = 25^\circ\text{C}$		3	11	ns
			$T_J = -40^\circ\text{C}$ to $140^\circ\text{C}$		3	14	
$T_{MOFF}$	From LO OFF to HO ON		$T_J = 25^\circ\text{C}$		3	11	ns
			$T_J = -40^\circ\text{C}$ to $140^\circ\text{C}$		3	14	
$T_{MON}$	From HO OFF to LO ON	UCC27211	$T_J = 25^\circ\text{C}$		2	9.5	ns
			$T_J = -40^\circ\text{C}$ to $140^\circ\text{C}$		2	14	
$T_{MOFF}$	From LO OFF to HO ON		$T_J = 25^\circ\text{C}$		2	9.5	ns
			$T_J = -40^\circ\text{C}$ to $140^\circ\text{C}$		2	14	
<b>Switching Parameters: Output Rise and Fall Time</b>							
$t_R$	LO rise time	$C_{LOAD} = 1000\text{ pF}$ , from 10% to 90%			7.2	ns	
$t_R$	HO rise time				7.2		
$t_F$	LO fall time	$C_{LOAD} = 1000\text{ pF}$ , from 90% to 10%			5.5		
$t_F$	HO fall time				5.5		
$t_R$	LO, HO	$C_{LOAD} = 0.1\text{ }\mu\text{F}$ , (3 V to 9 V)			0.36	0.6	$\mu\text{s}$
$t_F$	LO, HO	$C_{LOAD} = 0.1\text{ }\mu\text{F}$ , (9 V to 3 V)			0.15	0.4	
<b>Switching Parameters: Miscellaneous</b>							
	Minimum input pulse width that changes the output				50	ns	
	Bootstrap diode turn-off time <sup>(2)(3)</sup>	$I_F = 20\text{ mA}$ , $I_{REV} = 0.5\text{ A}$ <sup>(4)</sup>			20		

(2) Ensured by design.

(3)  $I_F$ : Forward current applied to bootstrap diode,  $I_{REV}$ : Reverse current applied to bootstrap diode.

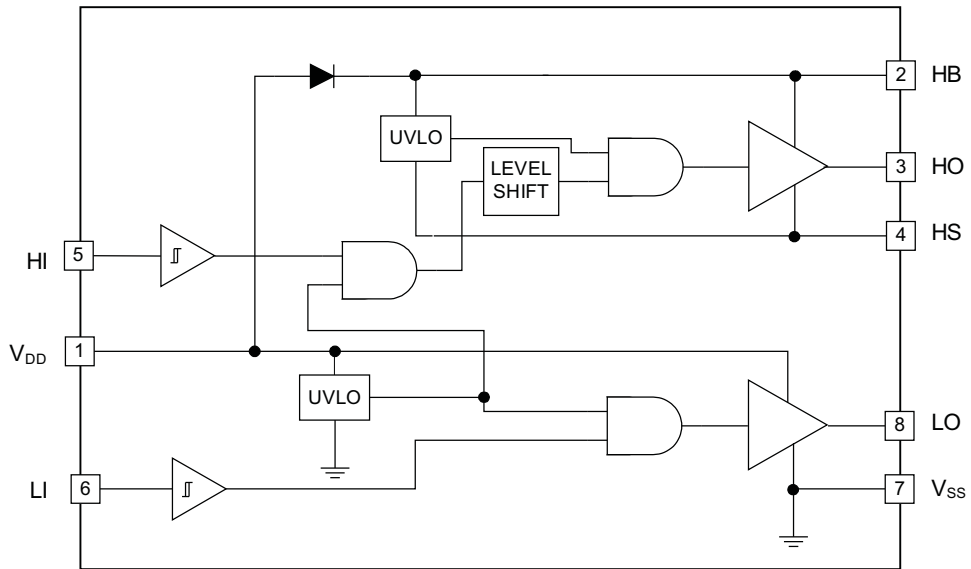
(4) Typical values for  $T_A = 25^\circ\text{C}$ .

Timing Diagrams

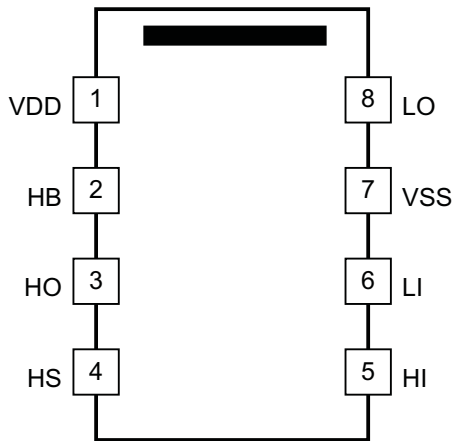


**DEVICE INFORMATION**

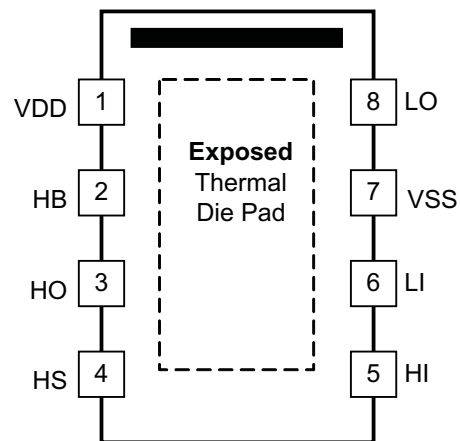
**Functional Block Diagram**



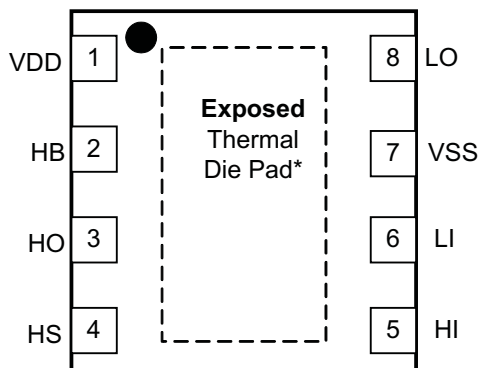
**SOIC-8 (D)**  
TOP VIEW



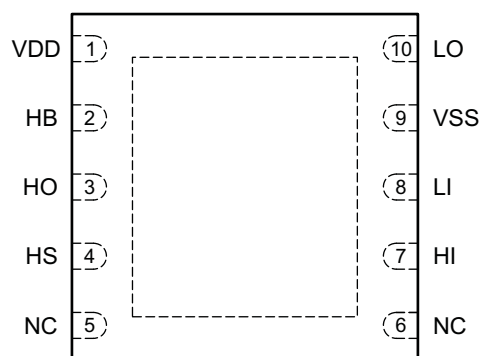
**Power Pad™ SOIC-8 (DDA)**  
TOP VIEW



**SON-8 (DRM)**  
TOP VIEW



**SON-10 (DPR)**  
TOP VIEW





**TERMINAL FUNCTIONS**

PIN NAME	PIN		DESCRIPTION
	D/DDA/DRM	DPR	
VDD	1	1	Positive supply to the lower-gate driver. De-couple this pin to V <sub>SS</sub> (GND). Typical decoupling capacitor range is 0.22 $\mu$ F to 1.0 $\mu$ F.
HB	2	2	High-side bootstrap supply. The bootstrap diode is on-chip but the external bootstrap capacitor is required. Connect positive side of the bootstrap capacitor to this pin. Typical range of HB bypass capacitor is 0.022 $\mu$ F to 0.1 $\mu$ F. The capacitor value is dependant on the gate charge of the high-side MOSFET and should also be selected based on speed and ripple criteria
HO	3	3	High-side output. Connect to the gate of the high-side power MOSFET.
HS	4	4	High-side source connection. Connect to source of high-side power MOSFET. Connect the negative side of bootstrap capacitor to this pin.
HI	5	7	High-side input.
LI	6	8	Low-side input.
VSS	7	9	Negative supply terminal for the device which is generally grounded.
LO	8	10	Low-side output. Connect to the gate of the low-side power MOSFET.
N/C	-	5/6	Not Connected.
PowerPAD™ <sup>(1)</sup>	Pad	Pad	Utilized on the DDA, DRM and DPR packages only. Electrically referenced to V <sub>SS</sub> (GND). Connect to a large thermal mass trace or GND plane to dramatically improve thermal performance.

- (1) The PowerPAD™ is not directly connected to any leads of the package. However it is electrically and thermally connected to the substrate which is the ground of the device.

TYPICAL CHARACTERISTICS

QUIESCENT CURRENT  
vs  
SUPPLY VOLTAGE

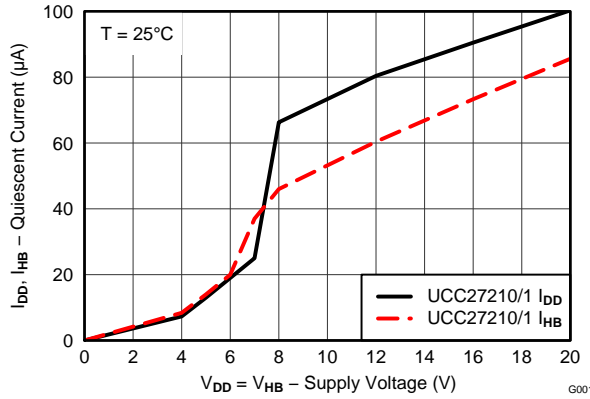


Figure 1.

UCC27210 IDD OPERATING CURRENT  
vs  
FREQUENCY

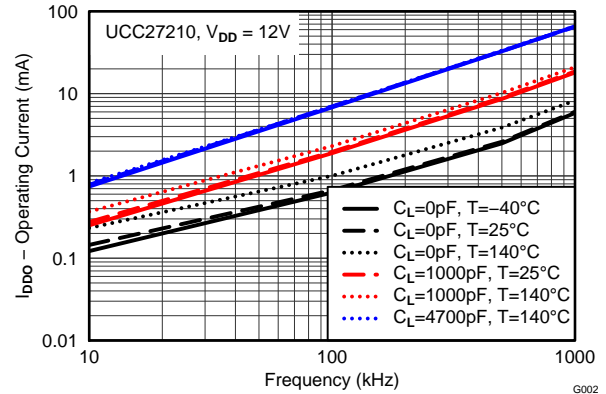


Figure 2.

UCC27211 IDD OPERATING CURRENT  
vs  
FREQUENCY

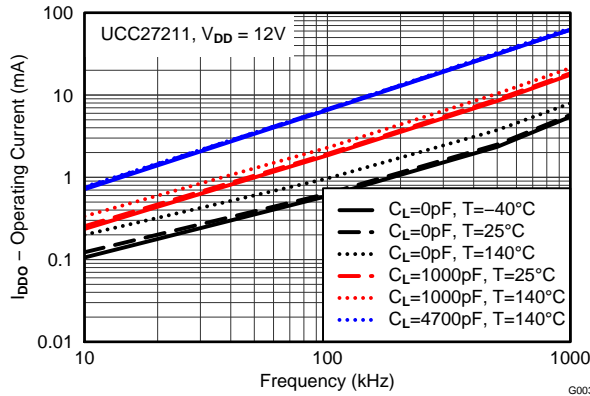


Figure 3.

BOOT VOLTAGE OPERATING CURRENT  
vs  
FREQUENCY (HB to HS)

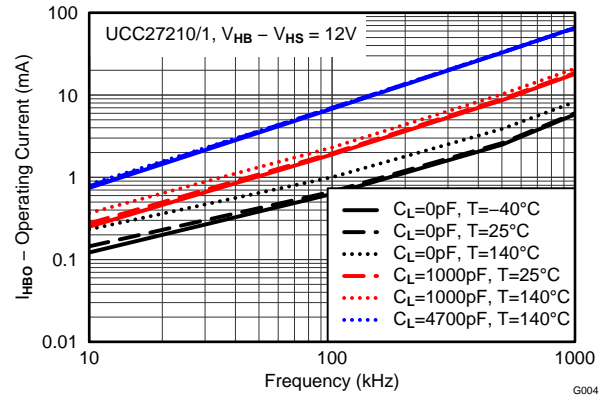


Figure 4.

UCC27210/11 INPUT THRESHOLD  
vs  
SUPPLY VOLTAGE

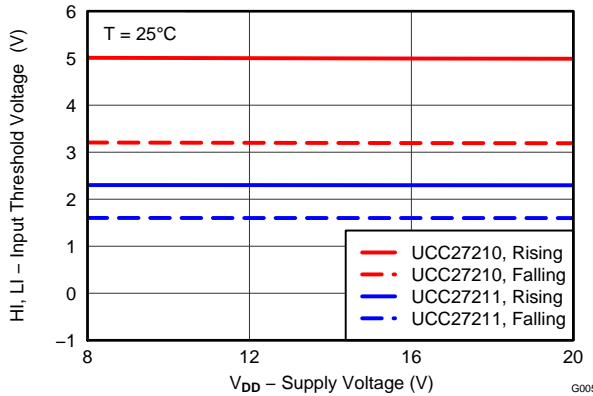


Figure 5.

UCC27210/11 INPUT THRESHOLDS  
vs  
TEMPERATURE

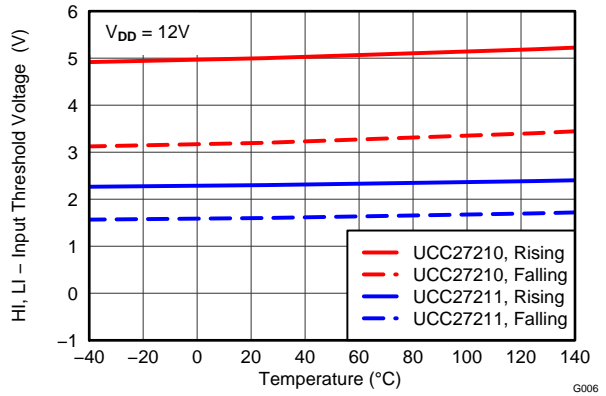


Figure 6.

**TYPICAL CHARACTERISTICS (continued)**

**LO AND HO HIGH LEVEL OUTPUT VOLTAGE vs TEMPERATURE**

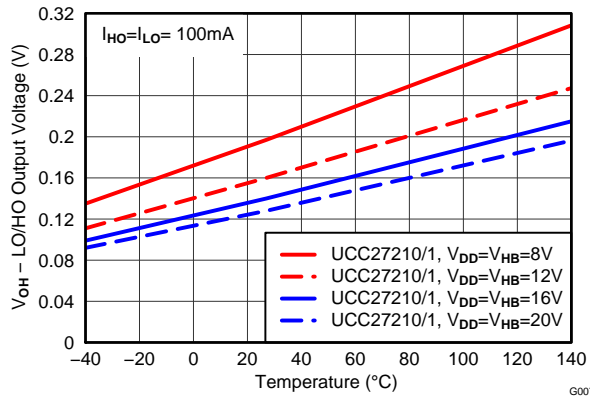


Figure 7.

**LO AND HO LOW LEVEL OUTPUT VOLTAGE vs TEMPERATURE**

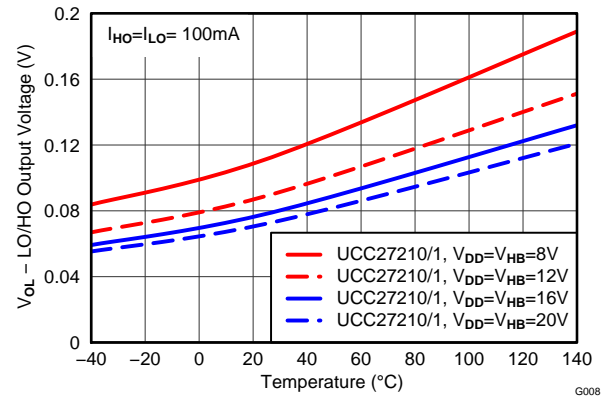


Figure 8.

**UNDervOLTAGE LOCKOUT THRESHOLD vs TEMPERATURE**

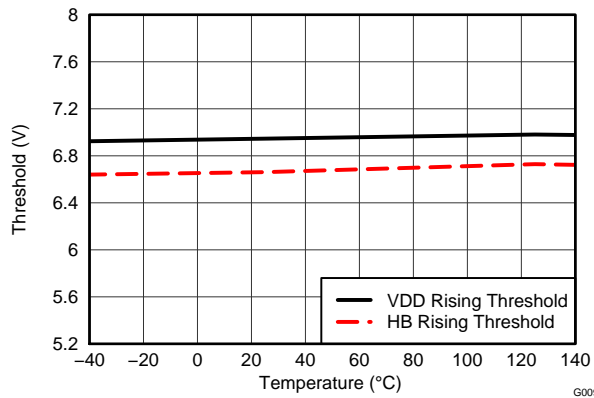


Figure 9.

**UNDervOLTAGE LOCKOUT THRESHOLD HYSTERESIS vs TEMPERATURE**

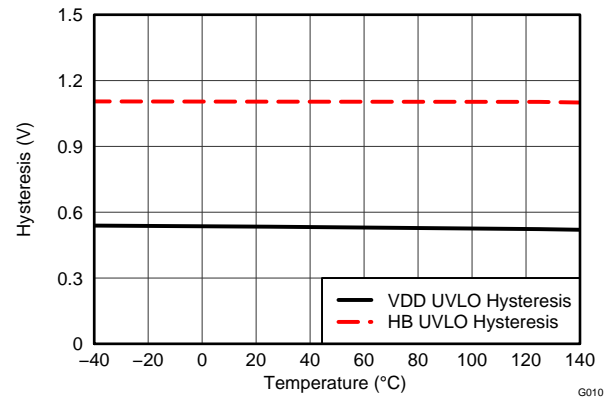


Figure 10.

**UCC27210 PROPAGATION DELAYS vs TEMPERATURE**

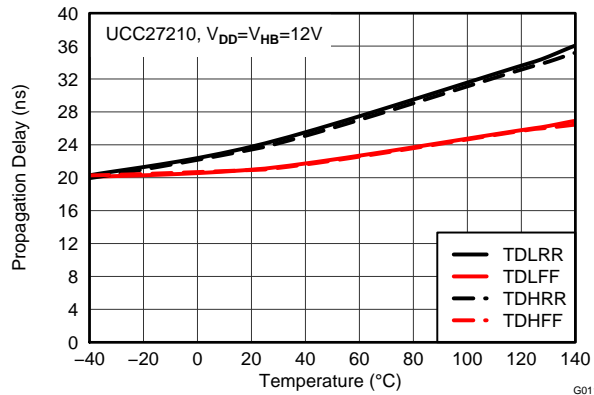


Figure 11.

**UCC27211 PROPAGATION DELAYS vs TEMPERATURE**

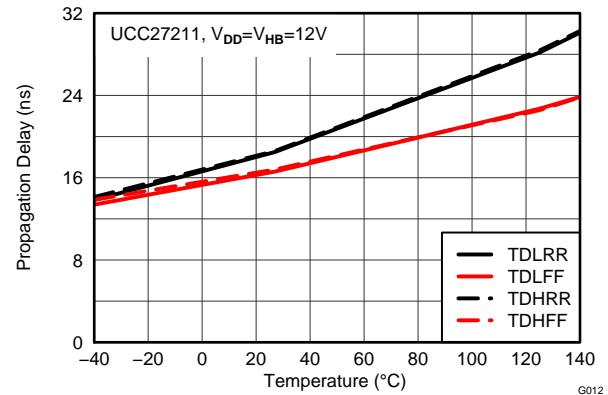
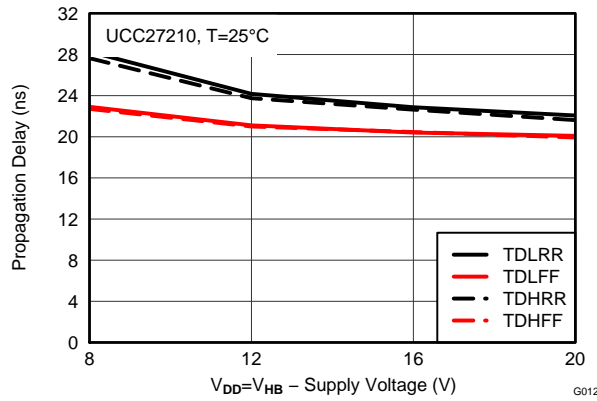


Figure 12.

**TYPICAL CHARACTERISTICS (continued)**

**UCC27210 PROPAGATION DELAYS**

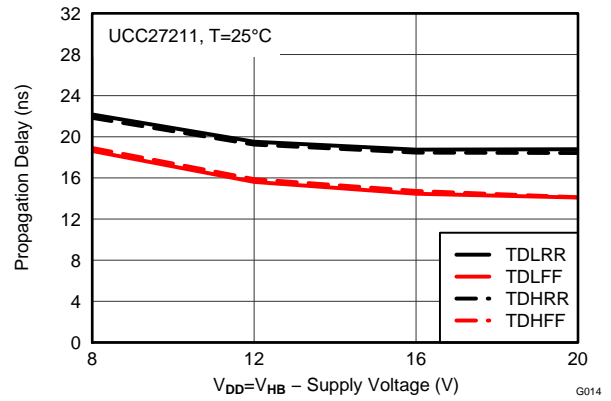
**vs  
SUPPLY VOLTAGE**



**Figure 13.**

**UCC27211 PROPAGATION DELAYS**

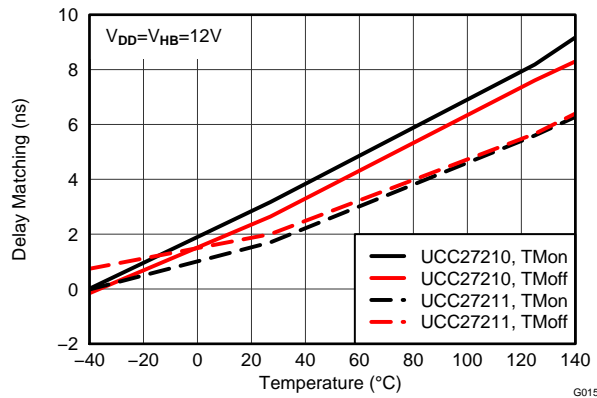
**vs  
SUPPLY VOLTAGE**



**Figure 14.**

**DELAY MATCHING**

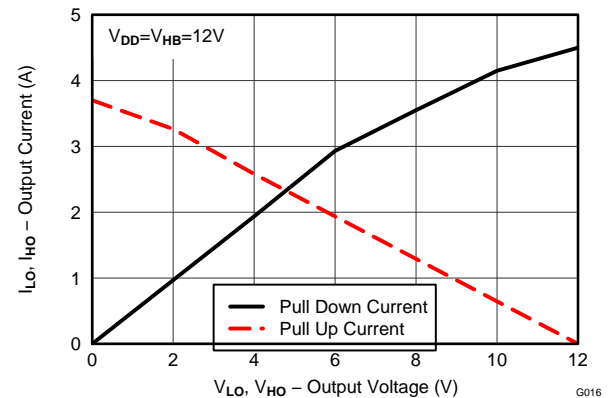
**vs  
TEMPERATURE**



**Figure 15.**

**OUTPUT CURRENT**

**vs  
OUTPUT VOLTAGE**



**Figure 16.**

TYPICAL CHARACTERISTICS (continued)

DIODE CURRENT  
vs  
DIODE VOLTAGE

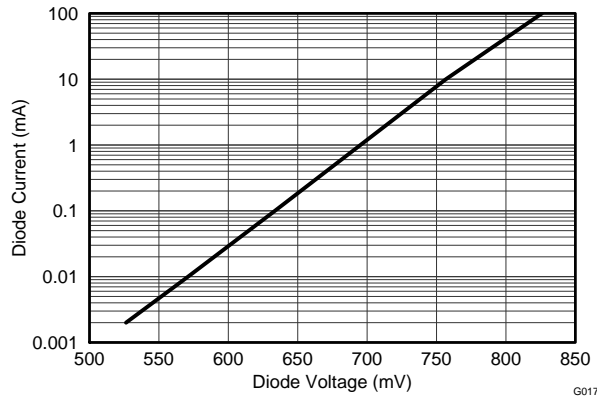


Figure 17.

NEGATIVE 10-V INPUT

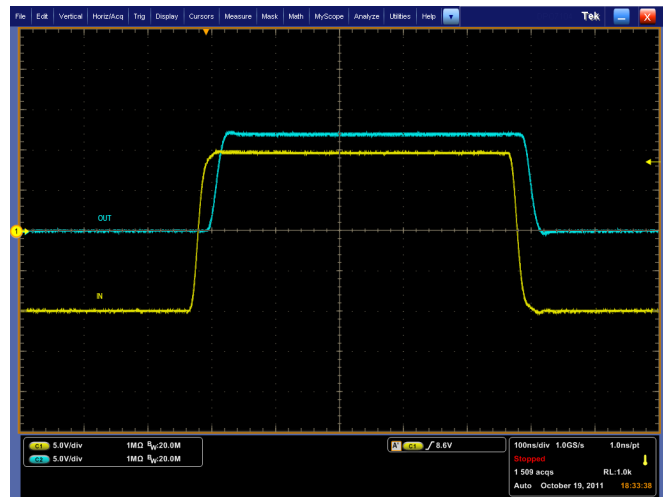


Figure 18.

STEP INPUT

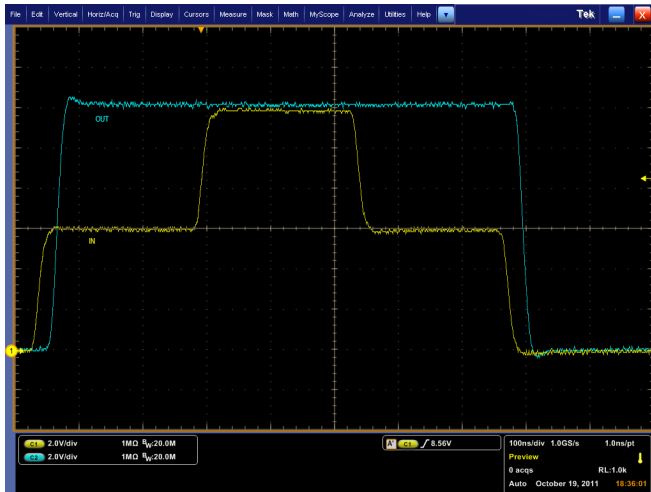


Figure 19.

SYMMETRICAL UVLO



Figure 20.

## APPLICATION INFORMATION

### Functional Description

The UCC27210/11 represent Texas Instruments' latest generation of high voltage gate drivers which are designed to drive both the high-side and low-side of N-Channel MOSFETs in a half-/full-bridge or synchronous buck configuration. The floating high-side driver is capable of operating with supply voltages of up to 120 V. This allows for N-Channel MOSFET control in half-bridge, full-bridge, push pull, two-switch forward and active clamp forward converters.

The UCC27210/11 feature 4-A source/sink capability, industry best-in-class switching characteristics and a host of other features listed in the table below. These features combine to ensure efficient, robust and reliable operation in high-frequency switching power circuits.

**Table 1. UCC27210/11 Highlights**

FEATURE	BENEFIT
4-A source and sink current with 0.9-Ω output resistance	High peak current ideal for driving large power MOSFETs with minimal power loss (fast-drive capability at Miller plateau)
Input pins (HI and LI) can directly handle -10 VDC up to 20 VDC	Increased robustness and ability to handle under/overshoot. Can interface directly to gate-drive transformers without having to use rectification diodes
120-V internal boot diode	Provides voltage margin to meet telecom 100-V surge requirements
Switch node (HS pin) able to handle -18 V maximum for 100 ns	Allows the high-side channel to have extra protection from inherent negative voltages caused parasitic inductance and stray capacitance.
Robust ESD circuitry to handle voltage spikes	Excellent immunity to large dV/dT conditions
18-ns propagation delay with 7.2-ns / 5.5-ns rise/fall Times	Best-in-class switching characteristics and extremely low-pulse transmission distortion
2-ns (typ) delay matching between channels	Avoids transformer volt-second offset in bridge
Symmetrical UVLO circuit	Ensures high-side and low-side shut down at the same time
CMOS optimized threshold or TTL optimized thresholds with increased hysteresis	Complementary to analog or digital PWM controllers. Increased hysteresis offers added noise immunity

In UCC27210/11, the high side and low side each have independent inputs which allow maximum flexibility of input control signals in the application. The boot diode for the high-side driver bias supply is internal to the UCC27210 and UCC27211. The UCC27210 is the Pseudo-CMOS compatible input version and the UCC27211 is the TTL or logic compatible version. The high-side driver is referenced to the switch node (HS) which is typically the source pin of the high-side MOSFET and drain pin of the low-side MOSFET. The low-side driver is referenced to  $V_{SS}$  which is typically ground. The functions contained are the input stages, UVLO protection, level shift, boot diode, and output driver stages.

## Input Stages

The input stages provide the interface to the PWM output signals. The input impedance of the UCC27210 is 100 k $\Omega$  nominal and input capacitance is approximately 2 pF. The 100 k $\Omega$  is a pull-down resistance to  $V_{SS}$  (ground). The UCC27210 Pseudo-CMOS input structure has been designed to provide large hysteresis and at the same time to allow interfacing to a multitude of analog or digital PWM controllers. In some CMOS designs, the input thresholds are determined as a percentage of VDD. By doing so, the high-level input threshold can become unreasonably high and unusable. The UCC27210 recognizes the fact that VDD levels are trending downward and it therefore provides a rising threshold with 5.0 V (typ) and falling threshold with 3.2 V (typ). The input hysteresis of the UCC27210 is 1.8 V (typ).

The input stages of the UCC27211 have impedance of 70 k $\Omega$  nominal and input capacitance is approximately 2 pF. Pull-down resistance to  $V_{SS}$  (ground) is 70 k $\Omega$ . The logic level compatible input provides a rising threshold of 2.3 V and a falling threshold of 1.6 V.

## Under Voltage Lockout (UVLO)

The bias supplies for the high-side and low-side drivers have UVLO protection.  $V_{DD}$  as well as  $V_{HB}$  to  $V_{HS}$  differential voltages are monitored. The  $V_{DD}$  UVLO disables both drivers when  $V_{DD}$  is below the specified threshold. The rising  $V_{DD}$  threshold is 7.0 V with 0.5-V hysteresis. The VHB UVLO disables only the high-side driver when the  $V_{HB}$  to  $V_{HS}$  differential voltage is below the specified threshold. The  $V_{HB}$  UVLO rising threshold is 6.7 V with 1.1-V hysteresis.

## Level Shift

The level shift circuit is the interface from the high-side input to the high-side driver stage which is referenced to the switch node (HS). The level shift allows control of the HO output referenced to the HS pin and provides excellent delay matching with the low-side driver.

## Boot Diode

The boot diode necessary to generate the high-side bias is included in the UCC27210/11 family of drivers. The diode anode is connected to  $V_{DD}$  and cathode connected to  $V_{HB}$ . With the  $V_{HB}$  capacitor connected to HB and the HS pins, the  $V_{HB}$  capacitor charge is refreshed every switching cycle when HS transitions to ground. The boot diode provides fast recovery times, low diode resistance, and voltage rating margin to allow for efficient and reliable operation.

## Output Stages

The output stages are the interface to the power MOSFETs in the power train. High slew rate, low resistance and high peak current capability of both output drivers allow for efficient switching of the power MOSFETs. The low-side output stage is referenced from  $V_{DD}$  to  $V_{SS}$  and the high side is referenced from  $V_{HB}$  to  $V_{HS}$ .

## Layout Recommendations

To improve the switching characteristics and efficiency of a design, the following layout rules should be followed.

- Locate the driver as close as possible to the MOSFETs.
- Locate the  $V_{DD}$  and  $V_{HB}$  (bootstrap) capacitors as close as possible to the driver.
- Pay close attention to the GND trace. Use the thermal pad of the DDA and DRM package as GND by connecting it to the VSS pin (GND). The GND trace from the driver goes directly to the source of the MOSFET but should not be in the high current path of the MOSFET(S) drain or source current.
- Use similar rules for the HS node as for GND for the high-side driver.
- Use wide traces for LO and HO closely following the associated GND or HS traces. 60 to 100-mils width is preferable where possible.
- Use as least two or more vias if the driver outputs or SW node needs to be routed from one layer to another. For GND the number of vias needs to be a consideration of the thermal pad requirements as well as parasitic inductance.
- Avoid LI and HI (driver input) going close to the HS node or any other high  $dV/dT$  traces that can induce significant noise into the relatively high impedance leads.

Keep in mind that a poor layout can cause a significant drop in efficiency versus a good PCB layout and can even lead to decreased reliability of the whole system.

## Example Component Placement

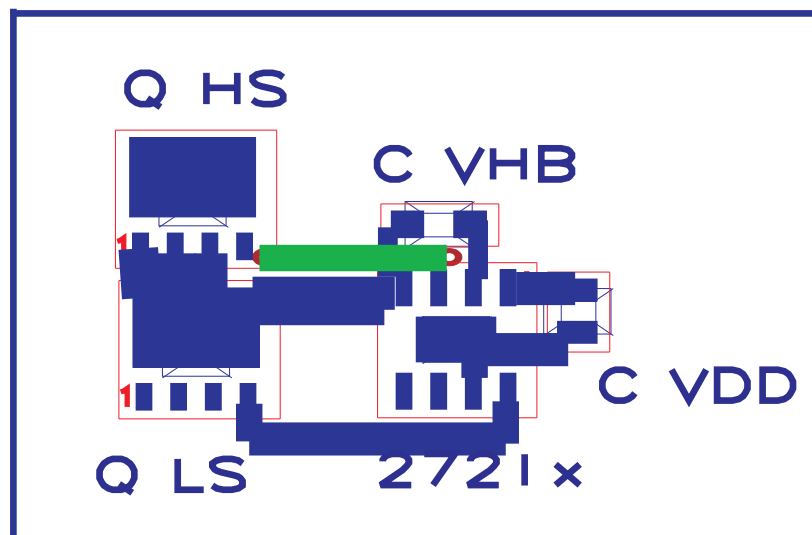


Figure 21. UCC27210/11 Component Placement

## Additional References

These references and links to additional information may be found at [www.ti.com](http://www.ti.com)

- Additional layout guidelines for PCB land patterns may be found in, *QFN/SON PCB Attachment*, Application Brief (Texas Instrument's Literature Number SLUA271)
- Additional thermal performance guidelines may be found in, *PowerPAD™ Thermally Enhanced Package Application Report*, Application Report (Texas Instrument's Literature Number SLMA002A)
- Additional thermal performance guidelines may be found in, *PowerPAD™ Made Easy*, Application Report (Texas Instrument's Literature Number SLMA004)



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## REVISION HISTORY

Changes from Revision A (November, 2011) to Revision B	Page
• Changed ordering information notes to reflect corrected part number. ....	<a href="#">2</a>

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**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
UCC27210D	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 140	27210	<a href="#">Samples</a>
UCC27210DDA	ACTIVE	SO PowerPAD	DDA	8	75	RoHS & Green	NIPDAUAG	Level-1-260C-UNLIM	-40 to 140	27210	<a href="#">Samples</a>
UCC27210DDAR	ACTIVE	SO PowerPAD	DDA	8	2500	RoHS & Green	NIPDAUAG	Level-1-260C-UNLIM	-40 to 140	27210	<a href="#">Samples</a>
UCC27210DPRR	ACTIVE	WSON	DPR	10	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 140	UCC 27210	<a href="#">Samples</a>
UCC27210DPRT	ACTIVE	WSON	DPR	10	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 140	UCC 27210	<a href="#">Samples</a>
UCC27210DR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 140	27210	<a href="#">Samples</a>
UCC27210DRMR	ACTIVE	VSON	DRM	8	3000	RoHS & Green	NIPDAUAG	Level-1-260C-UNLIM	-40 to 140	27210	<a href="#">Samples</a>
UCC27210DRMT	ACTIVE	VSON	DRM	8	250	RoHS & Green	NIPDAUAG	Level-1-260C-UNLIM	-40 to 140	27210	<a href="#">Samples</a>
UCC27211D	OBSOLETE	SOIC	D	8		TBD	Call TI	Call TI	-40 to 140	27211	
UCC27211DDA	OBSOLETE	SO PowerPAD	DDA	8		TBD	Call TI	Call TI	-40 to 140	27211	
UCC27211DDAR	ACTIVE	SO PowerPAD	DDA	8	2500	RoHS & Green	NIPDAUAG	Level-1-260C-UNLIM	-40 to 150	27211	<a href="#">Samples</a>
UCC27211DPRR	ACTIVE	WSON	DPR	10	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 150	UCC 27211	<a href="#">Samples</a>
UCC27211DPRT	OBSOLETE	WSON	DPR	10		TBD	Call TI	Call TI	-40 to 140	UCC 27211	
UCC27211DR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 150	27211	<a href="#">Samples</a>
UCC27211DRMR	ACTIVE	VSON	DRM	8	3000	RoHS & Green	NIPDAUAG	Level-1-260C-UNLIM	-40 to 150	27211	<a href="#">Samples</a>
UCC27211DRMT	OBSOLETE	VSON	DRM	8		TBD	Call TI	Call TI	-40 to 140	27211	

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of  $\leq 1000$ ppm threshold. Antimony trioxide based flame retardants must also meet the  $\leq 1000$ ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
UCC27210DDAR	SO PowerPAD	DDA	8	2500	330.0	12.8	6.4	5.2	2.1	8.0	12.0	Q1
UCC27210DPRR	WSON	DPR	10	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
UCC27210DPRT	WSON	DPR	10	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
UCC27210DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
UCC27210DRMR	VSON	DRM	8	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
UCC27210DRMT	VSON	DRM	8	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
UCC27211DDAR	SO PowerPAD	DDA	8	2500	330.0	12.8	6.4	5.2	2.1	8.0	12.0	Q1
UCC27211DPRR	WSON	DPR	10	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
UCC27211DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
UCC27211DRMR	VSON	DRM	8	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
UCC27210DDAR	SO PowerPAD	DDA	8	2500	364.0	364.0	27.0
UCC27210DPRR	WSON	DPR	10	3000	346.0	346.0	33.0
UCC27210DPRT	WSON	DPR	10	250	182.0	182.0	20.0
UCC27210DR	SOIC	D	8	2500	356.0	356.0	35.0
UCC27210DRMR	VSON	DRM	8	3000	356.0	356.0	35.0
UCC27210DRMT	VSON	DRM	8	250	210.0	185.0	35.0
UCC27211DDAR	SO PowerPAD	DDA	8	2500	364.0	364.0	27.0
UCC27211DPRR	WSON	DPR	10	3000	346.0	346.0	33.0
UCC27211DR	SOIC	D	8	2500	356.0	356.0	35.0
UCC27211DRMR	VSON	DRM	8	3000	356.0	356.0	35.0

**TUBE**


\*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
UCC27210D	D	SOIC	8	75	506.6	8	3940	4.32
UCC27210DDA	DDA	HSOIC	8	75	517	7.87	635	4.25



Images above are just a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.

DDA (R-PDSO-G8)

PowerPAD™ PLASTIC SMALL-OUTLINE



4202561/F 12/11

- NOTES:
- All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5-1994.
  - This drawing is subject to change without notice.
  - Body dimensions do not include mold flash or protrusion not to exceed 0,15.
  - This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at [www.ti.com](http://www.ti.com) <<http://www.ti.com>>.
  - See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
  - This package complies to JEDEC MS-012 variation BA

PowerPAD is a trademark of Texas Instruments.



DDA (R-PDSO-G8)

PowerPAD™ PLASTIC SMALL OUTLINE

## THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at [www.ti.com](http://www.ti.com).

The exposed thermal pad dimensions for this package are shown in the following illustration.



Exposed Thermal Pad Dimensions

4206322-6/L 05/12

NOTE: A. All linear dimensions are in millimeters

PowerPAD is a trademark of Texas Instruments



- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate designs.
  - D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at [www.ti.com](http://www.ti.com) <<http://www.ti.com>>. Publication IPC-7351 is recommended for alternate designs.
  - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
  - F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

PowerPAD is a trademark of Texas Instruments.

DRM (S-PVSON-N8)

PLASTIC SMALL OUTLINE NO-LEAD



- NOTES:
- All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
  - This drawing is subject to change without notice.
  - SON (Small Outline No-Lead) package configuration.
- The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.

### THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, Quad Flatpack No-Lead Logic Packages, Texas Instruments Literature No. SCBA017. This document is available at [www.ti.com](http://www.ti.com).

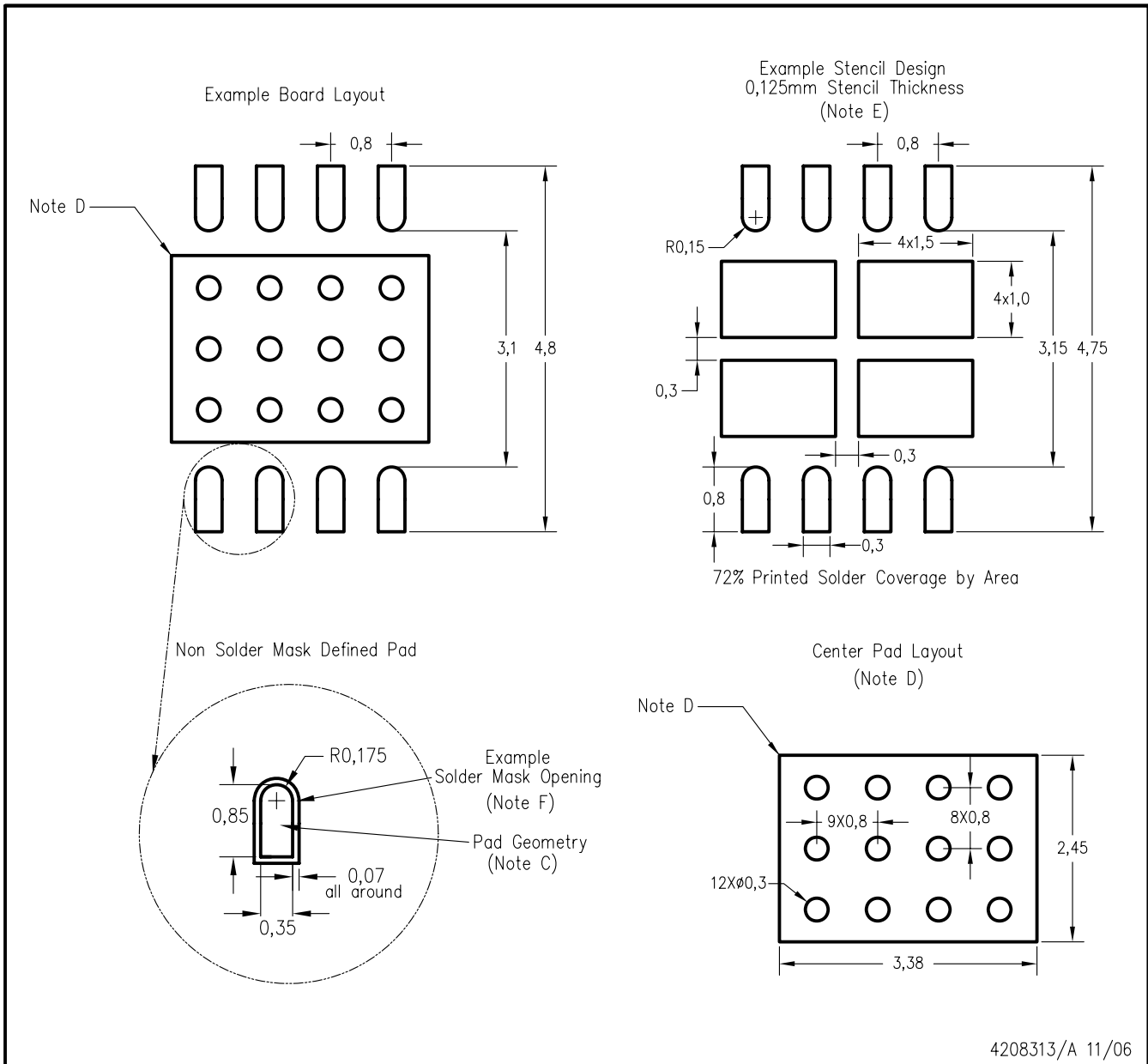
The exposed thermal pad dimensions for this package are shown in the following illustration.



NOTE: All linear dimensions are in millimeters

Exposed Thermal Pad Dimensions

DRM (S-PDSO-N8)



4208313/A 11/06

- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate designs.
  - D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN Packages, Texas Instruments Literature No. SCBA017, SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at [www.ti.com](http://www.ti.com) <<http://www.ti.com>>.
  - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
  - F. Customers should contact their board fabrication site for solder mask tolerances.



D0008A

# PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4214825/C 02/2019

NOTES:

1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed  $.006$  [0.15] per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MS-012, variation AA.

# EXAMPLE BOARD LAYOUT

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE  
 EXPOSED METAL SHOWN  
 SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



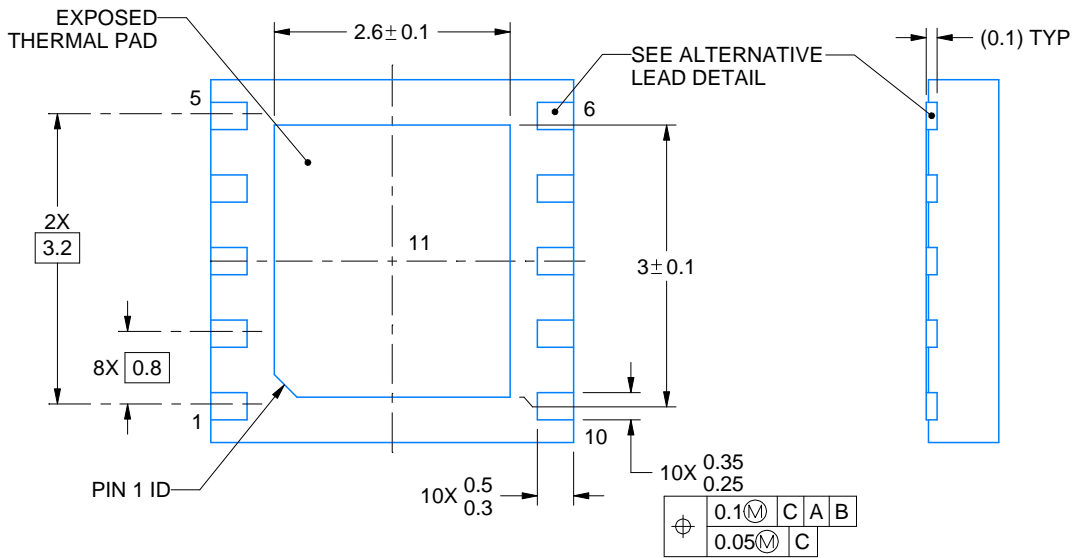
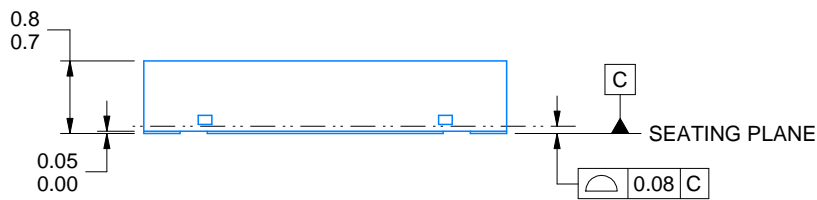
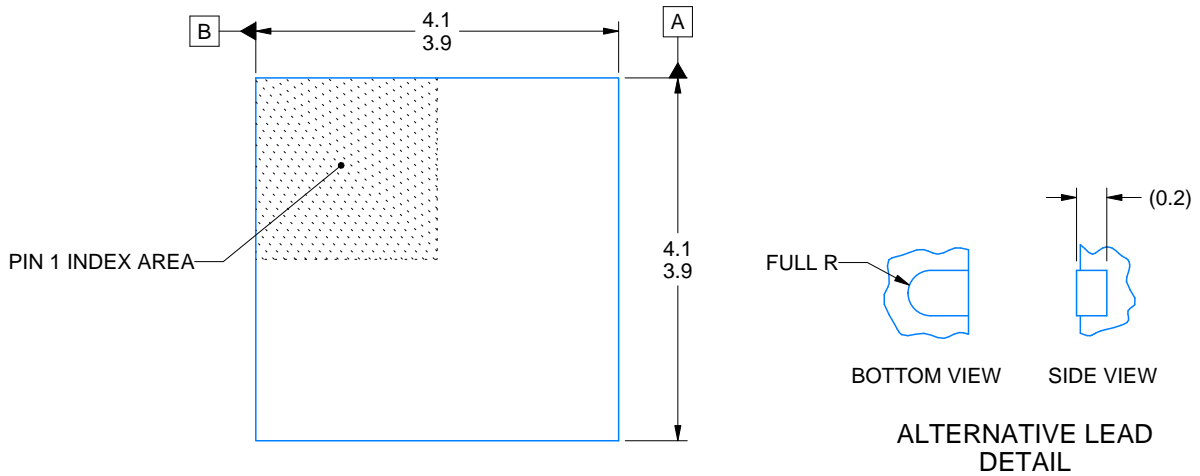
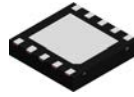
SOLDER PASTE EXAMPLE  
BASED ON .005 INCH [0.125 MM] THICK STENCIL  
SCALE:8X

4214825/C 02/2019

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.





4218856/B 01/2021

NOTES:

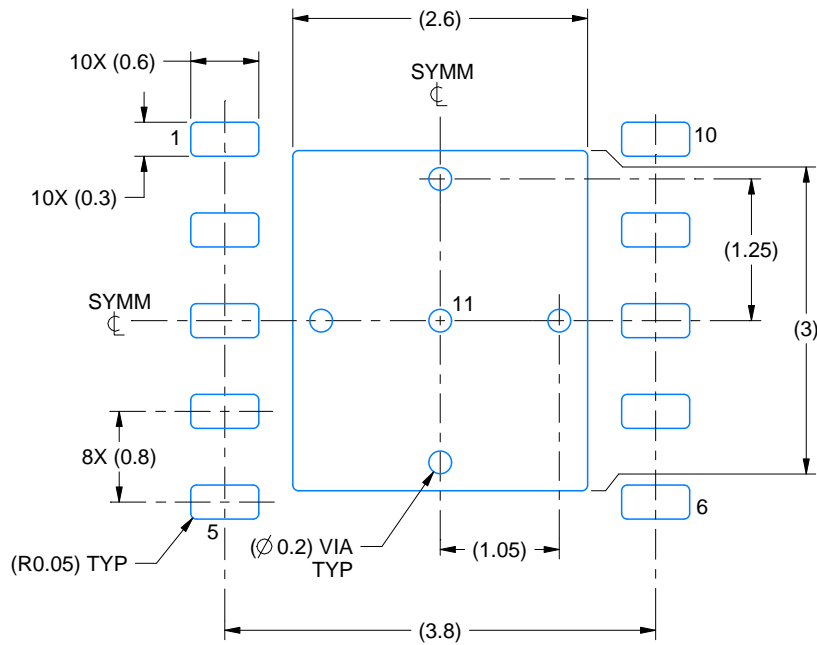
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

# EXAMPLE BOARD LAYOUT

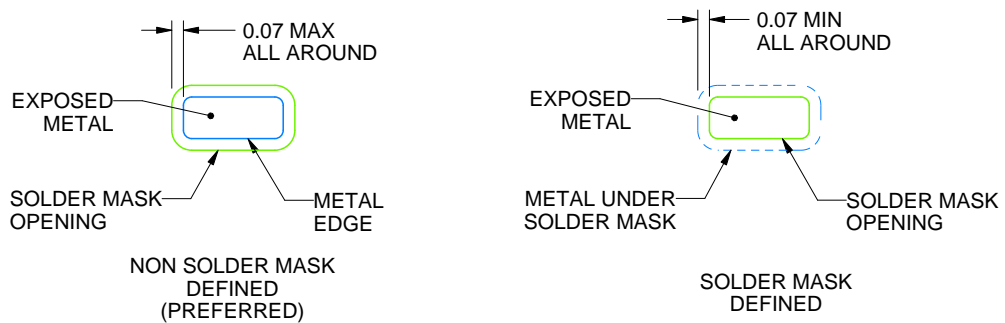
DPR0010A

WSON - 0.8 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:15X



SOLDER MASK DETAILS

4218856/B 01/2021

NOTES: (continued)

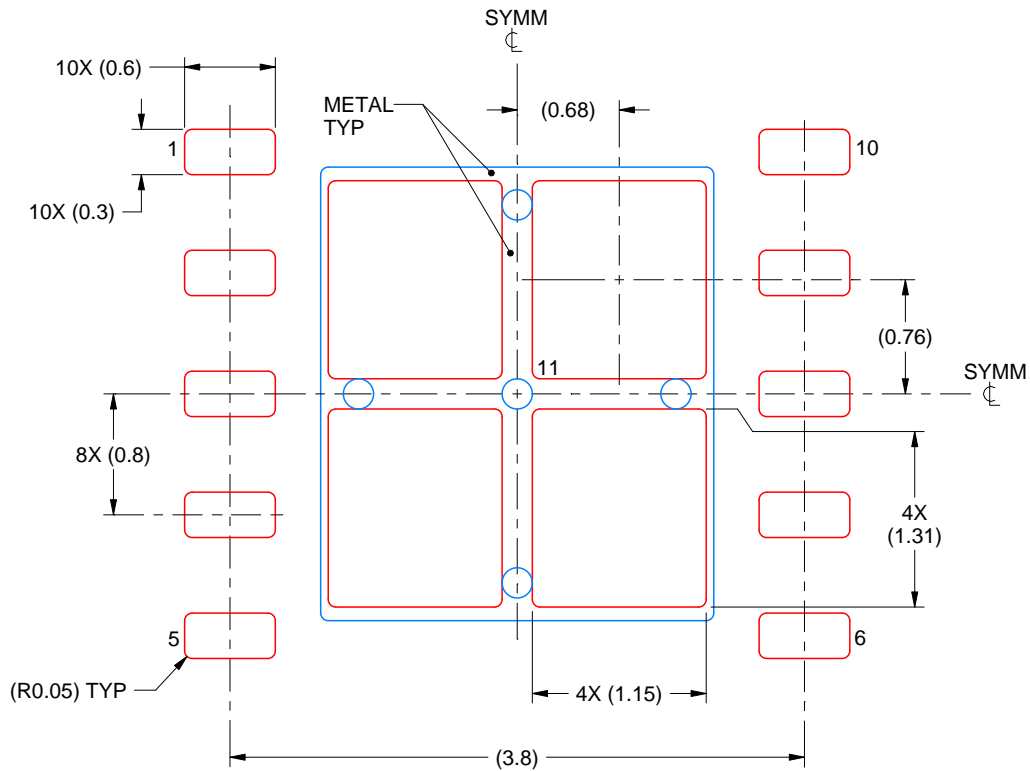
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 ([www.ti.com/lit/slua271](http://www.ti.com/lit/slua271)).

# EXAMPLE STENCIL DESIGN

DPR0010A

WSN - 0.8 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD 11:  
77% PRINTED SOLDER COVERAGE BY AREA  
SCALE:20X

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NOTES: (continued)

5. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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