

LMX1214 低ノイズ、高周波バッファおよびデバイダ

1 特長

- 300MHz から 18GHz までの周波数に対応するクロックバッファ
- 最大 8GHz のデバイダ出力周波数をサポート
- 超低ノイズ
 - ノイズフロア (6GHz 出力): -161dBc/Hz
 - 6GHz 出力時の 36fs の追加ジッタ (100Hz~ f_{CLK})
 - 5fs の追加ジッタ (100Hz~100MHz)
- 4 つの高周波クロック出力
 - 共有分周比は 1 (バイパス)、2、3、4、5、6、7、8
- 出力間スキュー標準値: 1ps
- 出力イネーブルおよびデバイダ I_2 、 I_3 、 I_4 の値を設定するためのピンモードオプションをサポート
- AUXCLK 出力
 - 個別の分周バンク上
 - 1、2、4 プリデバイダ
 - 1 (バイパス)、2、...、1023 ポストディバイダ
- 8 つのプログラム可能な出力電力レベル
- すべてのデバイダへの同期機能
- 2.5V の動作電圧
- 動作温度: -40°C ~ +85°C

2 アプリケーション

- 試験 & 測定:
 - オシロスコープ
 - ワイヤレス機器テスター

- 広帯域デジタイザ
- 航空宇宙 & 防衛:
 - レーダー
 - 電子戦
 - シーカー フロント エンド
 - 軍需品
 - フェーズド アレイ アンテナ / ビーム フォーミング
- 汎用:
 - データコンバータのクロック供給
 - クロックバッファの分配 / 分周

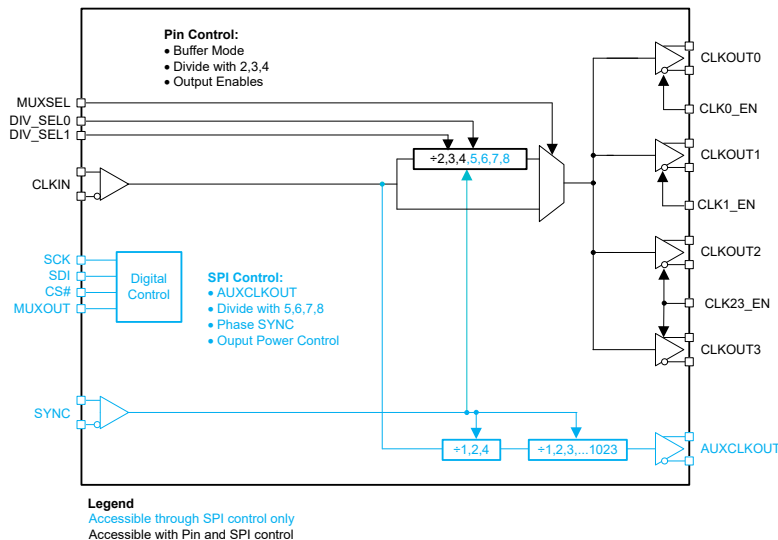
3 概要

LMX1214 は、高い出力周波数、超低ノイズフロア、超低スキュークロック分配を特長としています。このデバイスは、4 つの高周波出力クロックと、低周波数の補助クロック出力を備えています。このデバイスは、高周波クロックのバッファモードと分周モードの両方をサポートしています。このデバイスは、マルチチャネル、低スキュー、超低ノイズの局所発振器信号を複数のミキサに分配できます。

パッケージ情報

部品番号	パッケージ (1)	パッケージサイズ (2)
LMX1214	RHA (VQFN, 40)	6 mm × 6 mm

- 利用可能なすべてのパッケージについては、データシートの末尾にある注文情報を参照してください。
- パッケージサイズ (長さ × 幅) は公称値であり、該当する場合はピンも含まれます。



ブロック図



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4 Pin Configuration and Functions

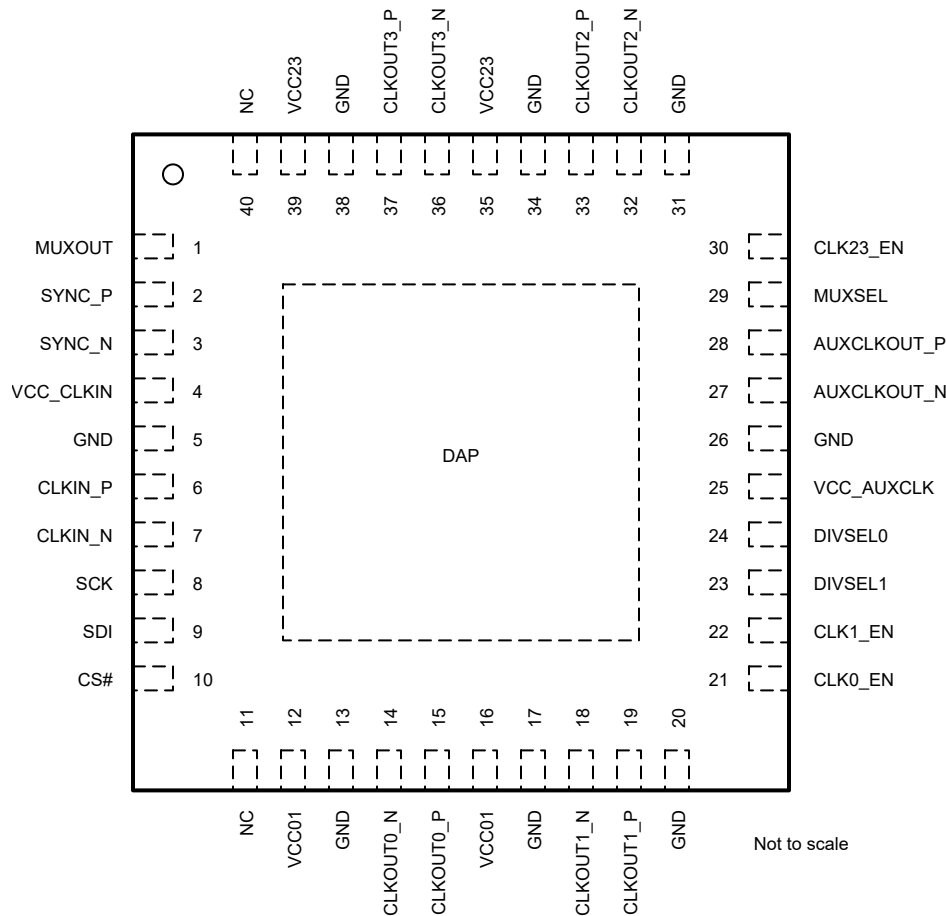


図 4-1. RHA0040C Package 40-Pin VQFN Top View

表 4-1. Pin Functions

NO.	NAME	TYPE ⁽¹⁾	DESCRIPTION
1	MUXOUT	O	SPI readback output (SDO).
2	SYNC_P	I	Differential synchronization input. Internal 100-Ω differential termination AC-coupled to GND with a 10-pf capacitor. Supports AC and DC coupling which can directly accept a common mode voltage of 1.2 V to 2 V.
3	SYNC_N		
4	VCC_CLKIN	PWR	Connect to a 2.5-V supply. Recommend a shunt RF wideband capacitor (typically 0.1 μF or smaller) close to the pin in parallel with larger capacitors (typically 1 μF and 10 μF). Large capacitors can be placed bit further away from the pin.
5	GND	GND	Ground these pins.
6	CLKIN_P	I	Differential reference input clock. Internal 100-Ω differential termination. AC-couple with a capacitor appropriate to the input frequency (typically 0.1 μF or smaller). If using single-ended, terminate unused pin with 50-Ω resistor AC-coupled to ground.
7	CLKIN_N		
8	SCK	I	SPI clock. High impedance CMOS input. Accepts up to 3.3 V.
9	SDI	I	SPI data input. High impedance CMOS input. Accepts up to 3.3 V.
10	CS#	I	SPI chip select. High impedance CMOS input. Accepts up to 3.3 V.
11	NC	—	Not connected

表 4-1. Pin Functions (続き)

NO.	NAME	TYPE ⁽¹⁾	DESCRIPTION
12	VCC01	PWR	Connect to a 2.5-V supply. Recommend a shunt RF wideband capacitor (typically 0.1 μ F or smaller) close to the pin in parallel with larger capacitors (typically 1 μ F and 10 μ F). Large capacitors can be placed bit further away from the pin.
13	GND	GND	Ground this pin
14	CLKOUT0_N	O	Differential clock output pairs. Each pin is an open-collector output with internally integrated 50- Ω resistor with programmable output swing. AC coupling required.
15	CLKOUT0_P		
16	VCC01	PWR	Connect to a 2.5-V supply. Recommend a shunt RF wideband capacitor (typically 0.1 μ F or smaller) close to the pin in parallel with larger capacitors (typically 1 μ F and 10 μ F). Large capacitors can be placed bit further away from the pin.
17	GND	GND	Ground this pin
18	CLKOUT1_N	O	Differential clock output pairs. Each pin is an open-collector output with internally integrated 50- Ω resistor with programmable output swing. AC coupling required.
19	CLKOUT1_P		
20	GND	GND	Ground this pin
21	CLK0_EN	I	CLKOUT0 output enable pin
22	CLK1_EN	I	CLKOUT1 output enable pin
23	DIVSEL1	I	Main clock out divider value 2, 3, and 4 selection control pins
24	DIVSEL0		
25	VCC_AUXCLK	PWR	Connect to a 2.5-V supply. Recommend a shunt RF wideband capacitor (typically 0.1 μ F or smaller) close to the pin in parallel with larger capacitors (typically 1 μ F and 10 μ F). Large capacitors can be placed bit further away from the pin.
26	GND	GND	Ground this pin
27	AUXCLKOUT_N	O	Differential clock output pair. Selectable CML, or LVDS format. Programmable common-mode voltage.
28	AUXCLKOUT_P		
29	MUXSEL	I	Clock out Bypass or Divider path MUX selection
30	CLK23_EN	I	CLKOUT2 & CLKOUT3 output enable pin
31	GND	GND	Ground this pin
32	CLKOUT2_N	O	Differential clock output pairs. Each pin is an open-collector output with internally integrated 50- Ω resistor with programmable output swing. AC coupling required.
33	CLKOUT2_P		
34	GND	GND	Ground this pin
35	VCC23	PWR	Connect to a 2.5-V supply. Recommend a shunt RF wideband capacitor (typically 0.1 μ F or smaller) close to the pin in parallel with larger capacitors (typically 1 μ F and 10 μ F). Large capacitors can be placed bit further away from the pin.
36	CLKOUT3_N	O	Differential clock output pairs. Each pin is an open-collector output with internally integrated 50- Ω resistor with programmable output swing. AC coupling required.
37	CLKOUT3_P		
38	GND	GND	Ground this pin
39	VCC23	PWR	Connect to a 2.5-V supply. Recommend a shunt RF wideband capacitor (typically 0.1 μ F or smaller) close to the pin in parallel with larger capacitors (typically 1 μ F and 10 μ F). Large capacitors can be placed bit further away from the pin.
40	NC	—	Not connected
GND	DAP	GND	Ground this pin

(1) I = Input; O = Output; GND = Ground; PWR = Power

5 Specifications

5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
V _{DD}	Power supply voltage	−0.3	2.75	V
V _{IN}	DC Input Voltage (SCK, SDI, CSB)	GND	3.6	V
V _{IN}	DC Input Voltage (SYNC)	GND	V _{DD} + 0.3	V
V _{IN}	AC Input Voltage (CLKIN)		2.1	V _{pp}
T _J	Junction temperature		150	°C
T _{stg}	Storage temperature	−65	150	°C

- (1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

5.2 ESD Ratings

			VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins ⁽¹⁾	±2500	V
		Charged device model (CDM), per ANSI/ESDA/JEDEC JS-002, all pins ⁽²⁾	±500	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
 (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V _{DD}	Supply voltage	2.4	2.5	2.6	V
T _A	Ambient temperature	−40		85	°C
T _J	Junction temperature			125	°C

5.4 Thermal Information

SYMBOL	THERMAL METRIC ⁽¹⁾	VALUE	UNIT
R _{θJA}	Junction-to-ambient thermal resistance	24.8	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	13.0	°C/W
R _{θJB}	Junction-to-board thermal resistance	6.9	°C/W
Ψ _{JT}	Junction-to-top characterization parameter	0.1	°C/W
Ψ _{JB}	Junction-to-board characterization parameter	6.9	°C/W
R _{θJC(bot)}	Junction-to-case (bottom) thermal resistance	0.5	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application note.

5.5 Electrical Characteristics

2.4 V ≤ VCC ≤ 2.6 V, −40°C ≤ T_A ≤ +85°C. Typical values are at VCC = 2.5 V, 25°C (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Current Consumption					

LMX1214

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2.4 V ≤ VCC ≤ 2.6 V, −40°C ≤ TA ≤ +85°C. Typical values are at VCC = 2.5 V, 25°C (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
I _{CC}	Supply current ⁽¹⁾	Powered up, all outputs on			530		mA
		Powered up, all outputs off			290		
		Powered down			12		
I _{ADD}	Additive output current	OUTx_PWR = 7			70		mA
	Divider current	Divide, CLK_DIV = 8			36		
I _{sync}	Supply current	with SYNC enable			685		mA
SYNC Pins							
V _{SYNC_single}	Voltage input range	DC Coupled single ended		0.6		1	V _{pp}
V _{SYNC}	Voltage input range	AC differential voltage		0.8		2	V _{pp}
V _{CM}	Input common mode	Differential 100 Ω Termination, DC coupled Set externally		1.2	1.3	2	V
Clock Input							
f _{IN}	Input frequency	Buffer Mode Only		0.3		18 ⁽²⁾	GHz
P _{IN}	Input power	Single-ended power at CLKIN_P or CLKIN_N		0		10	dBm
Clock Outputs							
f _{OUT}	Output frequency	Divide-by-2	Without Sync (pin mode)	0.15		8	GHz
f _{OUT}	Output frequency	Divide-by-2	With Sync	0.15		6.4	GHz
f _{OUT}	Output frequency	Buffer Mode		0.3		18 ⁽²⁾	
f _{OUT}	Output frequency	AUXCLK output		1		800	MHz
P _{OUT}	Output power	Single-Ended	f _{CLKLOUT} = 6 GHz OUTx_PWR = 7		6.5		dBm
			f _{CLKLOUT} = 12.8 GHz OUTx_PWR = 7		3.5		
			f _{CLKLOUT} = 18 GHz OUTx_PWR = 7		1.5		
Φ _{imb}	Output phase imbalance between P & N	Buffer mode				5	ps
t _{RISE}	Rise time (20% to 80%)	f _{CLKOUT} = 300 MHz			45		ps
t _{FALL}	Fall time (20% to 80%)	f _{CLKOUT} = 300 MHz			45		ps
t _{MUTE}	Output mute time	Falling edge of OE pin				30	μs
t _{UNMUTE}	Output unmute time	Rising edge of OE pin				30	μs
Propagation Delay and Skew							
t _{SKEW}	Magnitude of skew between outputs	CLKOUTx to CLKOUTy, not AUXCLK			1	10	ps
t _{DLY}	Propagation delay	T _A = 25°C	Bypass Mode		120		ps
			Divide Mode		125		
Δt _{DLY} /ΔT	Propagation delay variation over temperature	Bypass Mode			0.06		ps/C
Noise, Jitter, and Spurs							
J _{CKX}	Additive jitter	Additive Jitter. 12k to 100 MHz integration bandwidth.	Buffer Mode		5		fs, rms
Flicker	1/f flicker noise	Slew Rate > 8 V/ns, f _{CLK} = 6 GHz	Buffer Mode		−154		dBc/Hz
NFL	Noise Floor	f _{OUT} = 6 GHz; f _{Offset} = 100 MHz	Buffer Mode		−161		dBc/Hz
NFL			Divide-by-2		−160.5		

2.4 V ≤ VCC ≤ 2.6 V, −40°C ≤ TA ≤ +85°C. Typical values are at VCC = 2.5 V, 25°C (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
H2	Second harmonic	f _{OUT} = 6 GHz (differential), Buffer Mode			−25		dBc
		f _{OUT} = 6 GHz (single-ended), Buffer Mode			−12		
		f _{OUT} = 6 GHz, single-ended, Divide by 2			−13.5		
Coupling	AUXCLK to CLKOUT coupling	f _{OUT} = 6 GHz, Single-ended; f _{AUXCLK} = 300MHz			−70		dBc
NFL	Noise Floor	f _{AUXCLK} = 300MHz; LVDS mode			−152		dBc
NFL	Noise Floor	f _{AUXCLK} = 300MHz; CML mode			−151		dBc
P leakage	Leakage power Input to Output	Chip power down	Single Ended Input		−45		dBc
P leakage	Leakage power Input to Output	Pin OE = 0	Single Ended Input		−40		dBc
Digital Interface (SCK, SDI, CS#, MUXOUT,CLKx_EN,MUXSEL,DIVSELx)							
V _{IH}	High-level input voltage	SCK, SDI, CS#		1.4		3.3	V
	High-level input voltage	CLKx_EN, DIVSELx, MUXSEL		1.4		3.3	V
V _{IL}	Low-level input voltage	SCK, SDI, CS#		0		0.4	V
	Low-level input voltage	CLKx_EN, DIVSELx, MUXSEL		0		0.4	V
I _{IH}	High-level input current	SCK, SDI, CS#		−42		42	μA
	High-level input current	CLKx_EN, DIVSELx, MUXSEL		−42		42	μA
I _{IL}	Low-level input current	SCK, SDI, CS#		−25		25	μA
	Low-level input current	CLKx_EN, DIVSELx, MUXSEL		−25		25	μA
V _{OH}	High-level output voltage	MUXOUT	I _{OH} = 5 mA	1.4		V _{CC}	V
	High-level output voltage		I _{OH} = 0.1 mA	2.2		V _{CC}	V
V _{OL}	Low-level output voltage	MUXOUT	I _{OL} = 5 mA			0.45	V

- (1) Unless Otherwise Stated, f_{CLKIN}=6 GHz, CLK_MUX=Buffer, All clocks on with OUTx_PWR=7.
(2) SYNC and dividers supported up to 12.8GHz frequency

5.6 Timing Requirements

			MIN	NOM	MAX	UNIT
Timing Requirements						
f _{SPI}	SPI Read/Write Speed				2	MHz
t _{CE}	Clock to enable low time		20			ns
t _{CS}	Clock to data wait time		20			ns
t _{CH}	Clock to data hold time		20			ns
t _{CWH}	Clock pulse width high		100			ns
t _{CWL}	Clock pulse width low		100			ns
t _{CES}	Enable to clock setup time		20			ns
t _{EWH}	Enable pulse width high		50			ns
t _{CD}	Falling clock edge to data wait time		0		100	ns

5.7 Timing Diagram

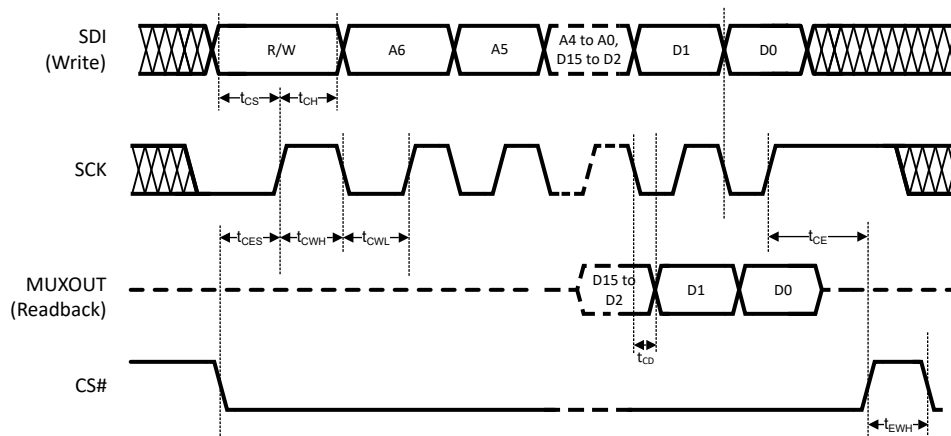


図 5-1. Serial Data Input Timing Diagram

There are several other considerations for writing on the SPI:

- The R/W bit must be set to 0.
- The data on SDI pin is clocked into a shift register on each rising edge on the SCK pin.
- The CS# must be held low for data to be clocked. The device ignores clock pulses if CS# is held high.
- Recommended SPI settings for this device are CPOL=0 and CPHA=0.
- When SCK and SDI lines are shared between devices, TI recommends to hold the CS# line high on the device that is not to be clocked.

There are several other considerations for SPI readback:

- The R/W bit must be set to 1.
- The MUXOUT pin is always be low for the address portion of the transaction.
- The data on MUXOUT is clocked out at the falling edge of SCK. In other words, the readback data is available at the MUXOUT pin t_{CD} after the clock falling edge.
- The data portion of the transition on the SDI line is always ignored.
- The MUXOUT pin does not automatically tri-state after a readback transaction completes. When sharing the SPI bus readback pin with other devices, set MUXOUT_EN=0 after all readback transactions from device are complete to manually tri-state the MUXOUT pin, permitting other devices to control the readback line.
- The values read back, even for R/W bits are not always the value written but rather an internal device state that takes into account the programmed value as well as other factors, such as pin states.

5.8 Typical Characteristics

If not otherwise, the following conditions can be assumed: Temperature = 25°C, Vcc = 2.5 V, OUTx_PWR = 7, CLKIN driven differentially with 8 dBm at each pin. Signal source used is SMA100B with ultra-low noise option B711.

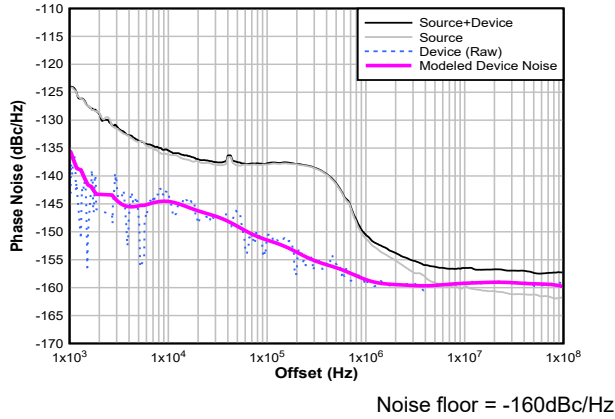


Figure 5-2. Buffer Phase Noise Plot at 6-GHz Output

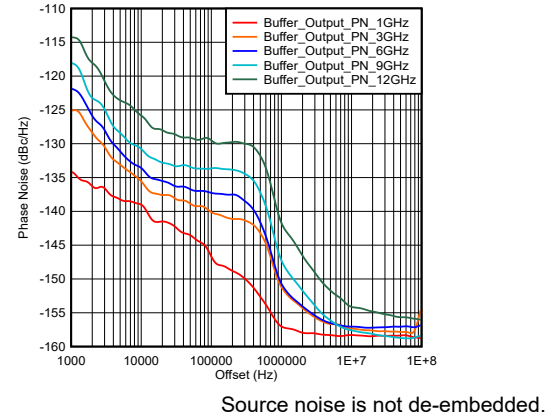


Figure 5-3. Buffer Output Phase Noise Plots vs Output Frequencies

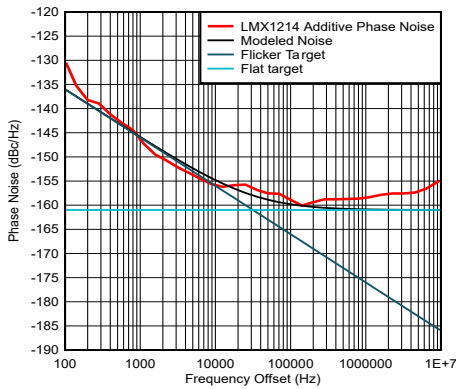


Figure 5-4. LMX1214 Flicker Noise at 6 GHz

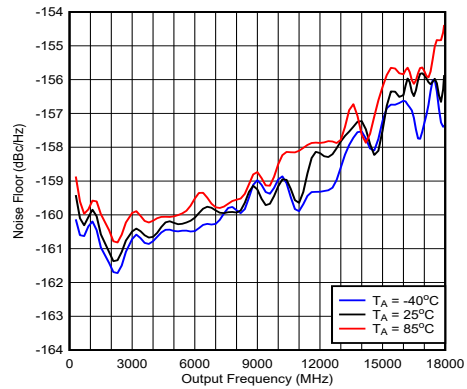


Figure 5-5. Noise Floor in Buffer Mode

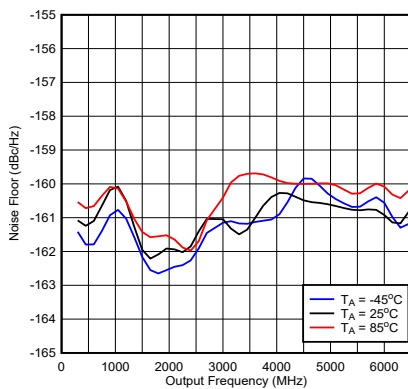


Figure 5-6. Noise Floor in Divider Mode

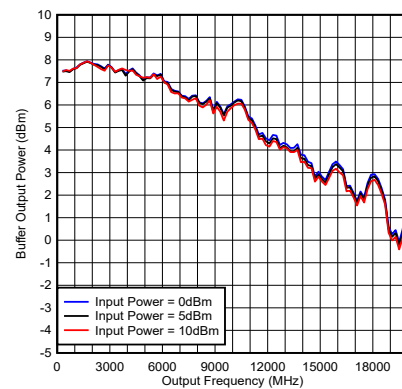
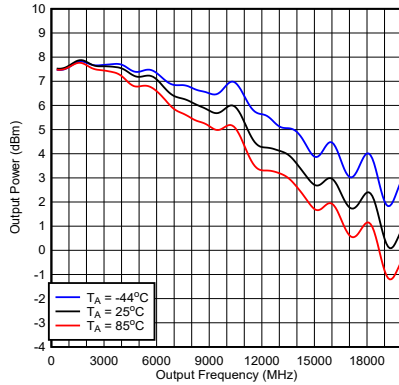
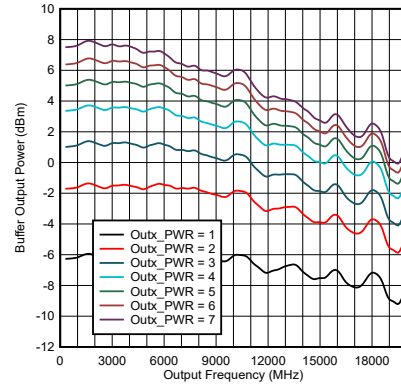


Figure 5-7. Buffer Mode Single-Ended Output Power vs Input Power

5.8 Typical Characteristics (continued)

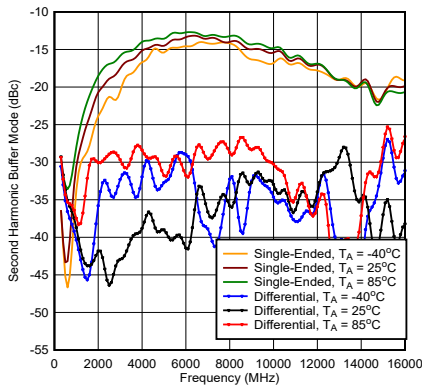


5-8. Buffer Mode Single-Ended Output Power Over Temperature

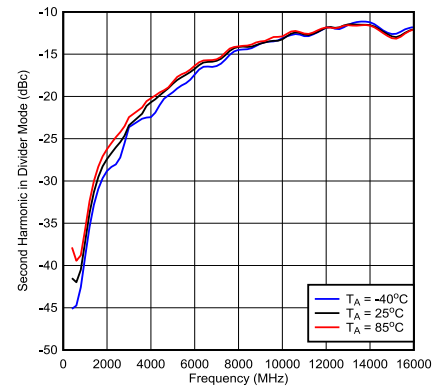


Applies to all modes except divider mode with odd divide (which have slightly lower power).

5-9. Buffer Mode Single-Ended Output Power vs Power Codes

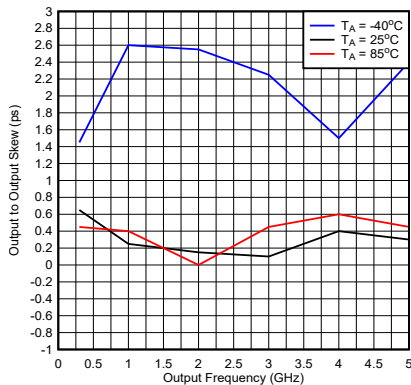


5-10. Second Harmonic in Buffer Mode

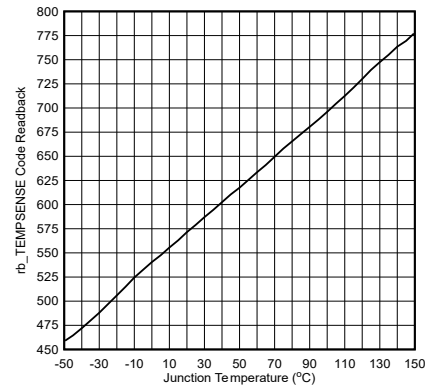


Single-Ended Input Power

5-11. Second Harmonic in Divider Mode



5-12. Output to Output Skew



Measured in power-down mode to make Junction Temperature = Ambient Temperature.

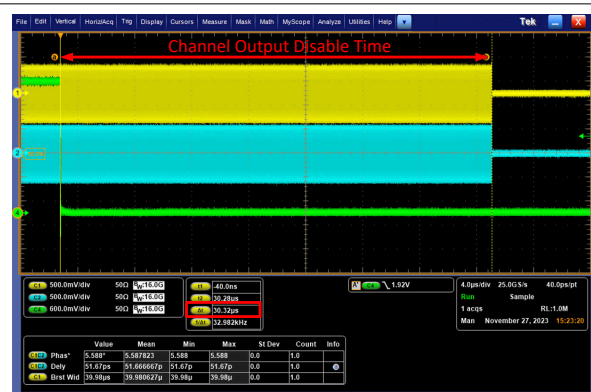
5-13. Temperature Sensor Readback

5.8 Typical Characteristics (continued)



at 300-MHz Output Frequency

図 5-14. Channel Enable Setting Time



at 300-MHz Output Frequency

図 5-15. Channel Disable Setting Time

6 Detailed Description

6.1 Overview

This device has four main clock outputs and another AUXCLK output. The main clock outputs are all the same frequency. This frequency can be the same or divided relative to the input clock. Each of these clock outputs has programmable power level. The AUXCLK output frequency is independent and typically lower frequency than the other four main clocks and has programmable output format (CML and LVDS) and power level.

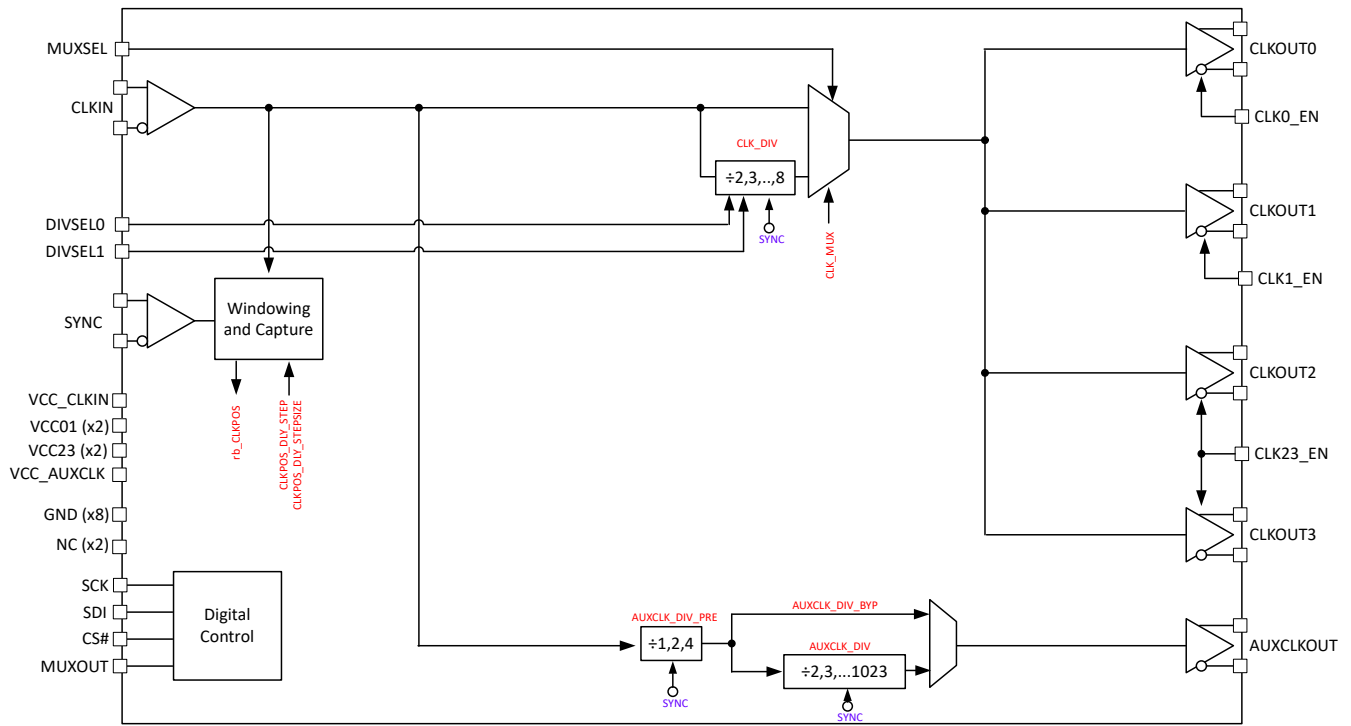
6.1.1 Range of Dividers

There are dividers that allow the main and AUXCLK outputs to be a divided value of the input clock. The main clock output has only one common divider, whereas the AUXCLK has a pre-divider and main divider.

表 6-1. Range of Dividers

CATEGORY		SPI PROGRAMMABLE RANGE	PIN PROGRAMMABLE RANGE	COMMENTS
Main Clocks	Buffer		Yes	
	Divider	2, 3, 4, ... 8	2,3,4	Odd divides (except 1) do not have 50% duty cycle
	Power Level	0,1,...,7	No	
AUXCLK	Divide	PreDivide	1, 2, 4	TotalDivide = PreDivide × Divide
		Divide	1, 2, 3, ... 1023	Odd divides (except 1) do not have 50% duty cycle

6.2 Functional Block Diagram



6-1. Functional Block Diagram

6.3 Feature Description

6.3.1 Power-On Reset

When the device is powered up, the power-on reset (POR) resets all registers to a default state as well as resets all state machines and dividers. For the power-on reset state, all the dividers are bypassed and the device performs as a 4-output buffer. Wait 100 μ s after the power supply rails before programming other registers to verify that this reset is finished. When there is no input clock present, the device power-on reset happens properly and functions well, but the device draws less current. The current changes after an input clock is added.

Performing a software power-on reset by writing RESET = 1 in the SPI bus is also possible and generally good practice. The RESET bit self-clears when the user writes to another register. The SPI bus can be used to override these states to the desired settings.

Although the device does have an automatic power-on reset, the device can be impacted by different ramp rates on the different supply pins, especially in the presence of a strong input clock signal. TI therefore recommends to do a software reset after POR. This can be done by programming RESET = 1. The reset bit can be cleared by programming any other register or setting RESET back to 0. Even at the maximum allowed SPI bus speed, the software reset event always completes before the subsequent SPI write.

6.3.2 Temperature Sensor

The junction temperature can be read back for purposes such as characterization or to make adjustments based on temperature. Such adjustments can include adjusting CLKOUTx_PWR to make the output power more stable or using external or digital delays to compensate for changes in propagation delay over temperature.

The junction temperature is typically higher than the ambient temperature due to power dissipation from the outputs and other functions on the device. 式 1 shows the relationship between the code read back and the junction temperature.

$$\text{Temperature} = 0.65 \times \text{Code} - 351 \quad (1)$$

式 1 is based on a best-fit line created from three devices from slow, nominal, and fast corner lots (nine parts total). The worst-case variation of the actual temperature from the temperature predicted by the best-fit line is 13°C, which works out to 20 codes.

6.3.3 Clock Outputs

This device has four main output clocks which share a common frequency. This does not include the additional lower frequency AUXCLK output.

6.3.3.1 Clock Output Buffers

The output buffers have a format that is open collector with an integrated pullup resistor, similar to CML.

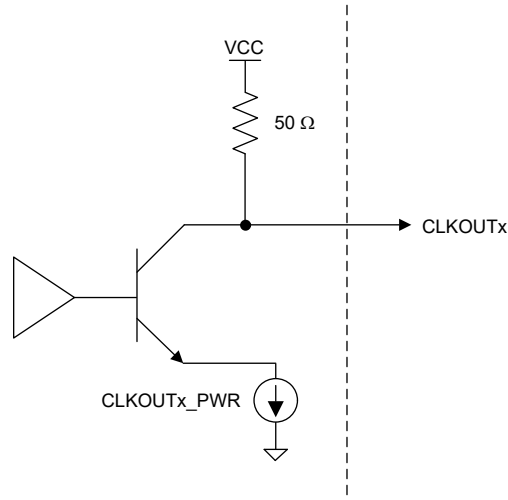


図 6-2. CLKOUT Output Buffer

The CLKOUTx_EN bits can enable the output buffers. The output power of the buffers can be individually set with CLKOUTx_PWR field.

表 6-2. Clock Output Power

CLKOUTx_EN	CLKOUTx_PWR	OUTPUT BUFFER
0	Don't Care	Powered Down
1	0	Minimum
	1	
	...	
	7	Maximum

6.3.3.2 Clock MUX

The four main clocks are the same frequency, but this frequency can be bypassed or divided. This is determined by the CLK_MUX word.

表 6-3. Clock MUX

CLK_MUX	OPTION	VALUES SUPPORTED
0	Buffer Mode	+1 (bypass)
1	Divider Mode	+2, 3, 4, 5, 6, 7, and 8

6.3.3.3 Clock Divider

Set the CLK_MUX to Divided to a divide value by 2, 3, 4, 5, 6, 7, or 8. This is set by the CLK_DIV word. When using the clock divider, any change to the input frequency requires the CLK_DIV_RST bit to be toggled from 1 to 0.

表 6-4. Clock Divider

CLK_DIV	DIVIDE	DUTY CYCLE
0	Reserved	n/a
1	2	50%
2	3	33%
3	4	50%
4	5	40%
5	6	50%
6	7	43%

表 6-4. Clock Divider (続き)

CLK_DIV	DIVIDE	DUTY CYCLE
7	8	50%

6.3.4 AUXCLK Output

The AUXCLK output can drive devices using lower frequency clocks, such as FPGAs. The AUXCLK output has the programmable output formats LVDS or CML.

6.3.4.1 AUXCLKOUT Output Format

The AUXCLKOUT output format can be programmed to LVDS and CML modes. Depending on the format, the common mode can be programmable or external components can be required.

表 6-5. AUXCLKOUT Formats and Properties

AUXCLKOUT_FMT	FORMAT	EXTERNAL COMPONENTS REQUIRED	OUTPUT LEVEL	COMMON MODE
0	LVDS	None	Fixed	Programmable through AUXCLKOUT_VCM
2	CML	Pullup Resistors 50 Ω to V _{CC}	Programmable through AUXCLKOUT_PWR	Not programmable

6.3.4.2 AUXCLK_DIV_PRE and AUXCLK_DIV Dividers

The AUXCLK_DIV_PRE divider and AUXCLK_DIV dividers are used for the AUXCLKOUT output. The AUXCLK_DIV_PRE divider is necessary to divide the frequency down to verify that the input to the AUXCLK_DIV divider is 3.2 GHz or less. When AUXCLK_DIV is not even and not bypassed, the duty cycle is not be 50%. Both the AUXCLKOUT dividers are synchronized by the SYNC feature, which allows synchronization across multiple devices.

表 6-6. Minimum N-Divider Restrictions

f _{CLKIN} (MHz)	AUXCLK_DIV_PRE	AUXCLK_DIV	TOTAL DIVIDE RANGE
f _{CLKIN} ≤ 3.2 GHz	+1,2,4	+1,2,3,...1023	[1, 2, ...1023] [2, 4, ... 2046] [4, 8, 4092]
3.2 GHz < f _{CLKIN} ≤ 6.4 GHz	+2,4	+1, 2, 3, ...1023	[4, ... 2046] [4, 8, 4092]
f _{CLKIN} > 6.4 GHz	+4	1, 2, 3, ...1023	[8, 4092]

6.3.5 SYNC Input Pins

The SYNC pins are used for the SYNC and windowing features. These pins can be DC or AC coupled and have internal dual 50-Ω, single-ended termination with programmable common-mode support. SYNC input is required, when the clock dividers (device clock and AUXCLK) need to be synchronized and multi-device sync needed. For buffer mode, sync input is not necessary.

The SYNC feature allows the user to synchronize the CLK_DIV, AUXCLK_DIV, and AUXCLK_DIV_PRE dividers so that the phase offset can be made consistent between power cycles. This allows users to synchronize multiple devices. Multi-device synchronization dividers can only be done through the SYNC pin, not the software.

6.3.5.1 SYNC Pins Common-Mode Voltage

The SYNC_P and SYNC_N pins can be driven either AC or DC coupled. When driven AC coupled, the common-mode voltage can be adjusted with the SYNC_VCM bit.

表 6-7. SYNC Pin Common-Mode Voltage

SYNC_VCM	COMMON-MODE VOLTAGE
0	1.3 V AC-coupled
1	1.1 V AC-coupled
2	1.5 V AC-coupled
3	No-Bias (DC Coupled)

6.3.5.2 Windowing Feature

The windowing feature can be used to internally calibrate the timing between the SYNC and CLKIN pins to optimize setup and hold timing and trim out any mismatches between the SYNC and CLKIN paths. This feature requires that the timing from the SYNC rising edge to the CLKIN rising edge is consistent. The timing from the SYNC rising edge to the CLKIN rising edges can be tracked with the rb_CLKPOS field. After the timing to the rising edge of the CLKIN pin is found, the SYNC rising edge can be internally adjusted with the SYNC_DLY and SYNC_DLY_STEP fields to optimize setup and hold times.

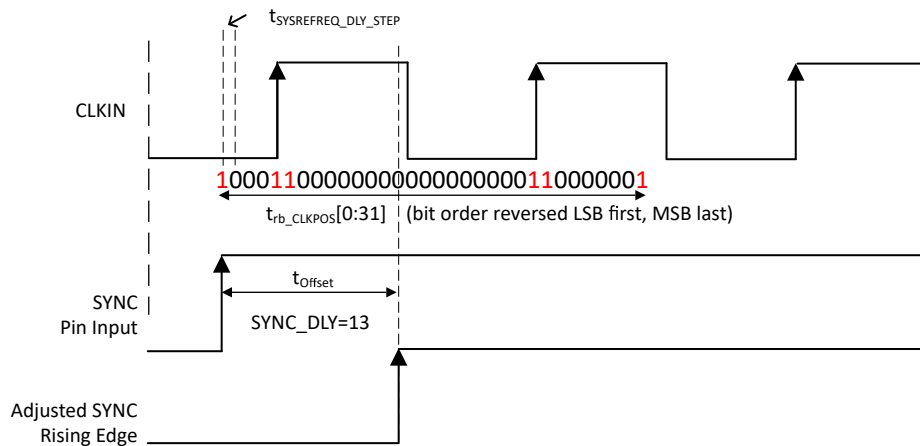


図 6-3. SYNC Internal Timing Adjustment

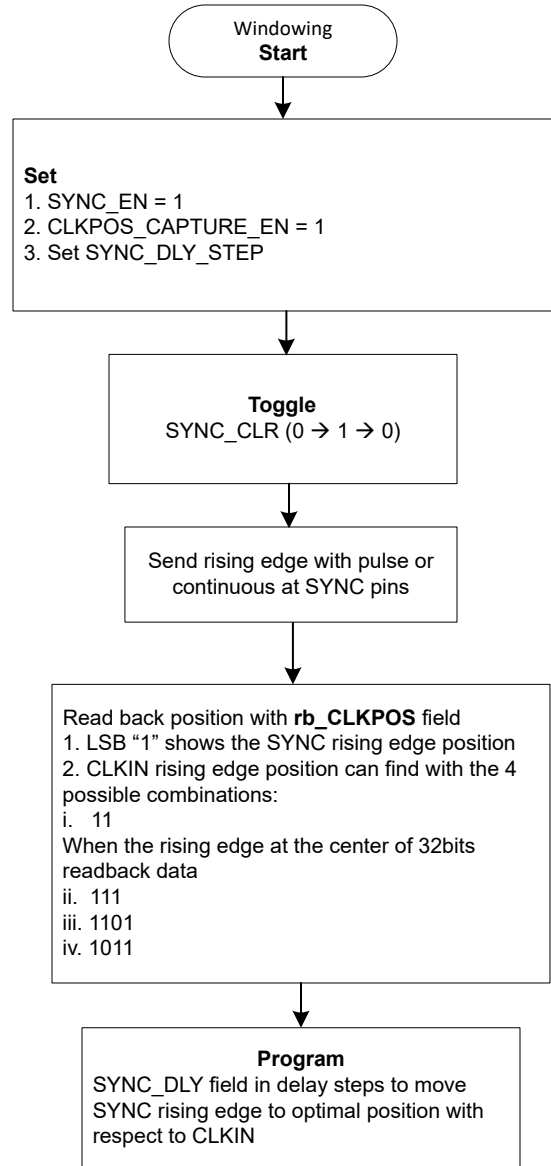


図 6-4. Flowchart for Windowing Operation

表 6-8. SYNC_DLY_STEP

INPUT FREQUENCY	RECOMMENDED SYNC_DLY_STEP	DELAY (ps)
$1.4 \text{ GHz} < f_{\text{CLKIN}} \leq 2.7 \text{ GHz}$	0	28
$2.4 \text{ GHz} < f_{\text{CLKIN}} \leq 4.7 \text{ GHz}$	1	15
$3.1 \text{ GHz} < f_{\text{CLKIN}} \leq 5.7 \text{ GHz}$	2	11
$f_{\text{CLKIN}} \geq 4.5 \text{ GHz}$	3	8

For the windowing feature:

1. The SYNC pins must be held high for a minimum time of $3/f_{\text{CLKIN}} + 1.6 \text{ ns}$ and only after this time rb_CLKPOS field is valid.
2. If the user infers multiple valid SYNC_DLY values from rb_CLKPOS registers to avoid setup-hold violations, TI recommends to choose the lowest valid SYNC_DLY to minimize variation over temperature.

For SYNC operation:

1. Only one SYNC pin rising edge is permitted per 75 input clock cycles
2. SYNC must stay high for more than six clock cycles

6.4 Device Functional Modes Configurations

The device can configure in high frequency clock buffer mode or divider mode. Each mode requires the below register configurations to function.

表 6-9. Device Functional Modes Settings

REGISTER ADDRESS	BIT	FIELD	FUNCTION	BUFFER	DIVIDER
R25	2:0	CLK_MUX	Select the mode	1	2
R25	5:3	CLK_DIV	Select the division value	x	CLK_DIV 0x1 = +2 0x2 = +3 0x3 = +4 0x4 = +5 0x5 = +6 0x6 = +7 0x7 = +8

6.4.1 Pin Mode Control

The LMX1214 supports pin mode for individual clock output control, mux selection and few dividers settings.

Below are the pin control options:

Output Channel Control

CLKOUT0 and CLKOUT1 have individual output control pins, whereas CLKOUT2 and CLKOUT3 uses a common control pin. These pins enable or disable the CLKOUT of particular channel outputs. Output channels can be controlled through SPI when CLKx_EN pin is high (1).

表 6-10. Output Channel Control Selection

CLKx_EN	CHANNEL OUTPUT STATE	SPI OVERRIDE CONTROL
0	Disable Channel Output	No
1	Enable Channel Output	Yes

Device Mode Selection

The device can function in buffer mode or divider mode, which can be set through the MUXSEL pin.

表 6-11. Device Mode Selection

MUXSEL	Mode Selection	SPI OVERRIDE CONTROL
0	Buffer Mode	Yes
1	Divider Mode	No

Divider Value Selection

When the divider mode is selected by MUXSEL pin logic or SPI control, the divider values (/2, /3 or /4) are selected by DIVSELx pin logic.

表 6-12. Divider Value Selection

DIVSEL1	DIVSEL0	DIVIDER VALUE
0	0	SPI Override
0	1	2
1	0	3
1	1	4

7 Register Map

	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	
R0	0	0	0	0	0	0	0	0	0	0	0	0	0	POWERDOW N	0	0	
R2	0	0	0	0	0	0	1	0	0	0	SMCLK K_EN	0	0	0	1	1	
R3	CLKO UT3_E N	CLKO UT2_E N	CLKO UT1_E N	CLKO UT0_E N	1	1	1	1	1	0	0	0	0	1	1	0	
R4	0	0	CLKOUT1_PWR			CLKOUT0_PWR			1	1	1	1	1	1	1	1	1
R5	0	0	1	1	0	1	1	0	1	1	CLKOUT3_PWR			CLKOUT2_PWR			
R7	0	1	0	AUXCLKOUT_ VCM		1	0	AUXCLK_DIV_ PWR_PRE		0	1	1	AUXCLKOUT_PWR			1	
R8	0	0	0	0	0	0	0	AUXCLK_DIV_PRE			AUXC LKOU T_EN	0	0	0	AUXCLKOUT_ FMT		
R9	SYNC_VCM		SYNC _EN	0	AUXC LK_DI V_BY P	0	AUXCLK_DIV										
R11	rb_CLKPOS																
R12	rb_CLKPOS[31:16]																
R13	0	0	0	0	0	0	0	0	0	0	0	0	0	0	SYNC_DLY_S TEP		
R14	0	0	0	0	0	0	0	0	0	0	0	0	0	0	1	SYNC _LATC H	
R15	0	0	0	0	1	0	1	1	0	SYNC_DLY						SYNC _CLR	
R23	TS_E N	1	MUXO UT_E N	0	0	0	0	0	0	0	0	0	0	0	0	0	
R24	0	0	0	0	rb_TS											TS_C NT_E N	
R25	0	0	0	0	0	0	1	0	0	CLK DIV_R ST	CLK_DIV			CLK_MUX			
R75	rb_CL KOUT 2_EN	rb_CL KOUT 1_EN	rb_CL KOUT 0_EN	rb_MU XSEL1	0	0	0	0	0	rb_DIV SEL1	rb_DIV SEL0	rb_CE	0	1	1	0	
R79	0	0	0	0	0	0	0	0	0	0	0	0	0	1	0	1	
R86	0	0	0	0	0	0	0	0	0	0	0	0	0	MUXO UT_E N_OV RD	0	0	
R90	0	0	0	0	0	0	0	0	0	AUXC LK_DI V_BY P3	AUXC LK_DI V_BY P2	0	0	0	0	0	

Programming is NOT required for buffer mode, but can be done to enable other features and enhancements

Registers	Enhancements
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R0, R3, R4, and R5	Outputs
R2, R75	Pin Modes
R7,R8,,R9,R79,R90	AUXCLK. R79 and R90 are only needed if using a divide by 1 if bypassing the pre-divider for the AUXCLK
R11,R12,R13,R14,R15	SYNC
R23, R24, R86	MUXOUT, Readback, and Temp Sensor
R25	Main Divide

7.1 Device Registers

表 7-1 lists the memory-mapped registers for the Device registers. All register offset addresses not listed in 表 7-1 must be considered as reserved locations and the register contents must not be modified.

表 7-1. DEVICE Registers

Offset	Acronym	Description	Section
0x0	R0		セクション 7.1.1
0x2	R2		セクション 7.1.2
0x3	R3		セクション 7.1.3
0x4	R4		セクション 7.1.4
0x5	R5		セクション 7.1.5
0x7	R7		セクション 7.1.6
0x8	R8		セクション 7.1.7
0x9	R9		セクション 7.1.8
0xB	R11		セクション 7.1.9
0xC	R12		セクション 7.1.10
0xD	R13		セクション 7.1.11
0xE	R14		セクション 7.1.12
0xF	R15		セクション 7.1.13
0x17	R23		セクション 7.1.14
0x18	R24		セクション 7.1.15
0x19	R25		セクション 7.1.16
0x4B	R75		セクション 7.1.17
0x4F	R79		セクション 7.1.18
0x56	R86		セクション 7.1.19
0x5A	R90		セクション 7.1.20

Complex bit access types are encoded to fit into small table cells. 表 7-2 shows the codes that are used for access types in this section.

表 7-2. Device Access Type Codes

Access Type	Code	Description
Read Type		
R	R	Read
Write Type		
W	W	Write
Reset or Default Value		
-n		Value after reset or the default value

7.1.1 R0 Register (Offset = 0x0) [Reset = 0x0000]

R0 is shown in 表 7-3.

Return to the [Summary Table](#).

表 7-3. R0 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-3	UNDISCLOSED	R	0x0	Program this field to 0x0.
2	POWERDOWN	R/W	0x0	Sets the device in a low-power state. The states of other registers are maintained.
1-0	UNDISCLOSED	R/W	0x0	Program this field to 0x0.

7.1.2 R2 Register (Offset = 0x2) [Reset = 0x0223]

R2 is shown in [表 7-4](#).

Return to the [Summary Table](#).

表 7-4. R2 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-11	UNDISCLOSED	R	0x0	Program this field to 0x0.
10-6	UNDISCLOSED	R/W	0x8	Program this field to 0x8.
5	SMCLK_EN	R/W	0x1	Enables the state machine clock generator. This is required for pin modes to function correctly and the part must be initialized with this bit enabled. However, this bit can later on be disabled to save current and prevent the state machine clock spur.
4-0	UNDISCLOSED	R/W	0x3	Program this field to 0x3.

7.1.3 R3 Register (Offset = 0x3) [Reset = 0xF0FE]

R3 is shown in [表 7-5](#).

Return to the [Summary Table](#).

表 7-5. R3 Register Field Descriptions

Bit	Field	Type	Reset	Description
15	CLKOUT3_EN	R/W	0x1	Enables CLKOUT3
14	CLKOUT2_EN	R/W	0x1	Enables CLKOUT2
13	CLKOUT1_EN	R/W	0x1	Enables CLKOUT1
12	CLKOUT0_EN	R/W	0x1	Enables CLKOUT0
11-0	UNDISCLOSED	R/W	0xFE	Program this field to 0xFE.

7.1.4 R4 Register (Offset = 0x4) [Reset = 0x36FF]

R4 is shown in [表 7-6](#).

Return to the [Summary Table](#).

表 7-6. R4 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-14	UNDISCLOSED	R	0x0	Program this field to 0x0.
13-11	CLKOUT1_PWR	R/W	0x6	Sets the output power of CLKOUT1. Larger values correspond to higher output power.
10-8	CLKOUT0_PWR	R/W	0x6	Sets the output power of CLKOUT0. Larger values correspond to higher output power.

表 7-6. R4 Register Field Descriptions (続き)

Bit	Field	Type	Reset	Description
7-0	UNDISCLOSED	R/W	0xFF	Program this field to 0xFF.

7.1.5 R5 Register (Offset = 0x5) [Reset = 0x36F6]

R5 is shown in [表 7-7](#).

Return to the [Summary Table](#).

表 7-7. R5 Register Field Descriptions

Bit	Field	Type	Reset	Description
15	UNDISCLOSED	R	0x0	Program this field to 0x0.
14-6	UNDISCLOSED	R/W	0xDB	Program this field to 0xDB.
5-3	CLKOUT3_PWR	R/W	0x6	Sets the output power of CLKOUT3. Larger values correspond to higher output power.
2-0	CLKOUT2_PWR	R/W	0x6	Sets the output power of CLKOUT2. Larger values correspond to higher output power.

7.1.6 R7 Register (Offset = 0x7) [Reset = 0x543F]

R7 is shown in [表 7-8](#).

Return to the [Summary Table](#).

表 7-8. R7 Register Field Descriptions

Bit	Field	Type	Reset	Description
15	UNDISCLOSED	R	0x0	Program this field to 0x0.
14-13	UNDISCLOSED	R/W	0x2	Program this field to 0x2.
12-11	AUXCLKOUT_VCM	R/W	0x2	In LVDS mode, sets the output common mode of the auxiliary clock output. Other output formats ignore this field. 0x0 = 1.2V 0x1 = 1.1V 0x2 = 1.0V 0x3 = 0.9V
10-9	UNDISCLOSED	R/W	0x2	Program this field to 0x2.
8-7	AUXCLK_DIV_PWR_PRE	R/W	0x0	Sets the output power of the AUXCLK pre-driver. Larger values correspond to higher output power.
6-4	UNDISCLOSED	R/W	0x3	Program this field to 0x3.
3-1	AUXCLKOUT_PWR	R/W	0x7	Sets the output power of AYXCLKOUT for CML format only (other output formats ignore this field). Larger values correspond to higher output power.
0	UNDISCLOSED	R/W	0x1	Program this field to 0x1.

7.1.7 R8 Register (Offset = 0x8) [Reset = 0x0120]

R8 is shown in [表 7-9](#).

Return to the [Summary Table](#).

表 7-9. R8 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-9	UNDISCLOSED	R	0x0	Program this field to 0x0.

表 7-9. R8 Register Field Descriptions (続き)

Bit	Field	Type	Reset	Description
8-6	AUXCLK_DIV_PRE	R/W	0x4	Sets pre-divider value. Output of the pre-divider must be less than or equal to 3.2 GHz. When AUXCLK_DIV_PRE=1, register R79 is also required to be programmed to a value of 0x0005 and R90 to 0x0060 (AUXCLK_DIV_BYP2=1, AUXCLK_DIV_BYP3=1). Values for AUXCLK_DIV_PRE other than those listed below are reserved. 0x1 = /1 0x2 = /2 0x4 = /4
5	AUXCLKOUT_EN	R/W	0x1	Enables AUXCLK subsystem.
4-2	UNDISCLOSED	R/W	0x0	Program this field to 0x0.
1-0	AUXCLKOUT_FMT	R/W	0x0	Selects the output driver format of the AUXCLKOUT output. 0x0 = LVDS 0x1 = Reserved 0x2 = CML 0x3 = Reserved

7.1.8 R9 Register (Offset = 0x9) [Reset = 0x0020]

R9 is shown in 表 7-10.

Return to the [Summary Table](#).

表 7-10. R9 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-14	SYNC_VCM	R/W	0x0	Sets the internal DC Bias for the SYNC pins. Bias must be enabled for AC-coupled inputs; but can be enabled and overdriven, or disabled, for DC-coupled inputs. SYNC DC pin voltage must be in the range of 0.7 V to VCC, including minimum and maximum signal swing. 0x0 = 1.3V 0x1 = 1.1V 0x2 = 1.5V 0x3 = Disabled
13	SYNC_EN	R/W	0x0	Enables synchronization path for the dividers and allows the clock position capture circuitry to be enabled. Used for multi-device synchronization.
12	UNDISCLOSED	R/W	0x0	Program this field to 0x0.
11	AUXCLK_DIV_BYP	R/W	0x0	Bypasses the AUXCLK_DIV divider to derive the AUXCLK output directly from the AUXCLK_DIV_PRE divider. Use only when AUXCLK_DIV_PRE=1 as one of the steps to achieve a total divide of 1 for the AUXCLK. To achieve a divide by 1, the following steps are required. 1. Set AUXCLK_DIV_PRE=1 2. Verify that register R79 is programmed to a value of 0x0005 3. Program R90 to 0x0060 (AUXCLK_DIV_BYP2=1, AUXCLK_DIV_BYP3=1) 4. Set AUXCLK_DIV_BYP=1 If a total divide of 1 for the AUXCLK is undesired, set this bit to 0.
10	UNDISCLOSED	R/W	0x0	Program this field to 0x0.
9-0	AUXCLK_DIV	R/W	0x20	Sets AUXCLK divider value. Maximum input frequency from AUXCLK_DIV_PRE must be \leq 3200 MHz. The maximum AUXCLKOUT frequency must be \leq 800 MHz to avoid amplitude degradation. 0x0 = Reserved 0x1 = Reserved 0x2 = /2 0x3 = /3 0x3FF = /1023

7.1.9 R11 Register (Offset = 0xB) [Reset = 0x0000]

R11 is shown in [表 7-11](#).

Return to the [Summary Table](#).

表 7-11. R11 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-0	rb_CLKPOS	R	0x0	Stores a snapshot of the CLKIN signal rising edge positions relative to a SYNC rising edge, with the snapshot starting from the LSB and ending at the MSB. Each bit represents a sample of the CLKIN signal, separated by a delay determined by the SYNC_DLY_STEP field. The first and last bits of rb_CLKPOS are always set, indicating uncertainty at the capture window boundary conditions. CLKIN rising edges are represented by every sequence of two set bits from LSB to MSB, including bits at the boundary conditions. The position of the CLKIN rising edges in the snapshot, along with the CLKIN signal period and the delay step size, can be used to compute the value of SYNC_DLY which maximizes setup and hold times for SYNC signals on the SYNC pins.

7.1.10 R12 Register (Offset = 0xC) [Reset = 0x0000]

R12 is shown in [表 7-12](#).

Return to the [Summary Table](#).

表 7-12. R12 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-0	rb_CLKPOS[31:16]	R	0x0	MSB of rb_CLKPOS field.

7.1.11 R13 Register (Offset = 0xD) [Reset = 0x0003]

R13 is shown in [表 7-13](#).

Return to the [Summary Table](#).

表 7-13. R13 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-2	UNDISCLOSED	R	0x0	Program this field to 0x0.
1-0	SYNC_DLY_STEP	R/W	0x3	Sets the step size of the delay element used in the SYSNC path, both for SYNC input delay and for clock position captures. The recommended frequency range for each step size creates the maximum number of usable steps for a given CLKIN frequency. The ranges include some overlap to account for process and temperature variations. If the CLKIN frequency is covered by an overlapping span, larger delay step sizes improve the likelihood of detecting a CLKIN rising edge during a clock position capture. However, since larger values include more delay steps, larger step sizes have greater total delay variation across PVT relative to smaller step sizes. 0x0 = 28 ps (1.4GHz to 2.7GHz) 0x1 = 15 ps (2.4GHz to 4.7GHz) 0x2 = 11 ps (3.1GHz to 5.7GHz) 0x3 = 8 ps (4.5GHz to 12.8GHz)

7.1.12 R14 Register (Offset = 0xE) [Reset = 0x0002]

R14 is shown in [表 7-14](#).

Return to the [Summary Table](#).

表 7-14. R14 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-3	UNDISCLOSED	R/W	0x0	Program this field to 0x0.
2	CLKPOS_CAPTURE_EN	R/W	0x0	Enables the windowing circuit which captures the clock position in the rb_CLKPOS registers with respect to a SYNC edge. The windowing circuit must be cleared by toggling SYNC_CLR high then low before a clock position capture. The first rising edge on the SYNC pins after clearing the windowing circuit triggers the capture. The capture circuitry greatly increases supply current, and does not need to be enabled to delay the SYNC signal in SYNC mode. Once the desired value of SYNC_DLY is determined, set this bit to 0x0 to minimize current consumption. If SYNC_EN = 0, the value of this bit is ignored, and the windowing circuit is disabled.
1	UNDISCLOSED	R/W	0x1	Program this field to 0x1.
0	SYNC_LATCH	R/W	0x0	Latches the internal SYNC state to logic high on the first rising edge of the SYNC pins. This latch can be cleared by setting SYNC_CLR=1.

7.1.13 R15 Register (Offset = 0xF) [Reset = 0x0B01]

R15 is shown in [表 7-15](#).

Return to the [Summary Table](#).

表 7-15. R15 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-12	UNDISCLOSED	R	0x0	Program this field to 0x0.
11-7	UNDISCLOSED	R/W	0x16	Program this field to 0x16.
6-1	SYNC_DLY	R/W	0x0	Sets the delay line step for the external SYNC signal. Each delay line step delays the SYNC signal by an amount equal to SYNC_DLY_STEP x SYNC_DLY_STEP. In SYNC mode, the value for this field can be determined based on the rb_CLKPOS value to satisfy the internal setup and hold time of the SYNC signal with respect to the CLKIN signal. In SYSREF Repeater Mode, the value for this field can be used as a coarse global delay. Values greater than 0x3F are invalid. Since larger values include more delay steps, larger values have greater total step size variation across PVT relative to smaller values. Refer to the data sheet or the device TICS Pro profile for detailed description of the delay step computation procedure.
0	SYNC_CLR	R/W	0x1	Clears SYNC_LATCH and resets synchronization path timing for SYNC signal. Holding this bit high keeps internal SYNC signal low. This bit must be set and cleared once before the SYNC or clock position capture operations are performed.

7.1.14 R23 Register (Offset = 0x17) [Reset = 0x4000]

R23 is shown in [表 7-16](#).

Return to the [Summary Table](#).

表 7-16. R23 Register Field Descriptions

Bit	Field	Type	Reset	Description
15	TS_EN	R/W	0x0	Enables the on-die temperature sensor. Temperature sensor counter (TS_CNT_EN) must also be enabled for readback.
14	UNDISCLOSED	R/W	0x1	Program this field to 0x1.

表 7-16. R23 Register Field Descriptions (続き)

Bit	Field	Type	Reset	Description
13	MUXOUT_EN	R/W	0x0	Enables or tri-states the MUXOUT pin driver. 0x0 = Tri-States 0x1 = Push-Pull
12-0	UNDISCLOSED	R/W	0x0	Program this field to 0x0.

7.1.15 R24 Register (Offset = 0x18) [Reset = 0x0000]

R24 is shown in [表 7-17](#).

Return to the [Summary Table](#).

表 7-17. R24 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-14	UNDISCLOSED	R	0x0	Program this field to 0x0.
13-12	UNDISCLOSED	R/W	0x0	Program this field to 0x0.
11-1	rb_TS	R	0x0	Readback value of on-die temperature sensor.
0	TS_CNT_EN	R/W	0x0	Enables temperature sensor counter. Temperature sensor (TS_EN) must be enabled for accurate data.

7.1.16 R25 Register (Offset = 0x19) [Reset = 0x0211]

R25 is shown in [表 7-18](#).

Return to the [Summary Table](#).

表 7-18. R25 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-7	UNDISCLOSED	R/W	0x4	Program this field to 0x4.
6	CLK_DIV_RST	R/W	0x0	Resets the main clock divider. If the clock divider value is changed during operation, set this bit high then low after setting the new divider value. Synchronizing the device with the SYNC pins with SYNC_EN = 0x1 also resets the main clock divider. This bit has no effect when outside of Divider Mode.
5-3	CLK_DIV	R/W	0x2	Sets the clock divider value when CLK_MUX=2 (Divider Mode). The clock divider value is CLK_DIV+1. Valid value for CLK_DIV is 1 to 7. Setting this to 0 disables the main clock divider and reverts to buffer mode.
2-0	CLK_MUX	R/W	0x1	Selects the function for the main clock outputs 0x0 = Reserved 0x1 = Buffer 0x2 = Divider 0x3 = Reserved

7.1.17 R75 Register (Offset = 0x4B) [Reset = 0x0006]

R75 is shown in [表 7-19](#).

Return to the [Summary Table](#).

表 7-19. R75 Register Field Descriptions

Bit	Field	Type	Reset	Description
15	rb_CLKOUT2_EN	R	0x0	Readback Pin Status
14	rb_CLKOUT1_EN	R	0x0	Readback Pin Status

表 7-19. R75 Register Field Descriptions (続き)

Bit	Field	Type	Reset	Description
13	rb_CLKOUT0_EN	R	0x0	Readback Pin Status
12	rb_MUXSEL1	R	0x0	Readback Pin Status
11-7	UNDISCLOSED	R	0x0	Program this field to 0x0.
6	rb_DIVSEL1	R	0x0	Readback Pin Status
5	rb_DIVSEL0	R	0x0	Readback Pin Status
4	rb_CE	R	0x0	Readback Pin Status
3-0	UNDISCLOSED	R/W	0x6	Program this field to 0x6.

7.1.18 R79 Register (Offset = 0x4F) [Reset = 0x0205]

R79 is shown in [表 7-20](#).

Return to the [Summary Table](#).

表 7-20. R79 Register Field Descriptions

Bit	Field	Type	Reset	Description
15	UNDISCLOSED	R	0x0	Program this field to 0x0.
14-0	UNDISCLOSED	R/W	0x205	Program this field to 0x5. Note that this is different than the reset value.

7.1.19 R86 Register (Offset = 0x56) [Reset = 0x0000]

R86 is shown in [表 7-21](#).

Return to the [Summary Table](#).

表 7-21. R86 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-3	UNDISCLOSED	R/W	0x0	Program this field to 0x0.
2	MUXOUT_EN_OVRD	R/W	0x0	This bit must be set to 1 to enable MUXOUT_EN to tri-state the MUXOUT pin.
1-0	UNDISCLOSED	R/W	0x0	Program this field to 0x0.

7.1.20 R90 Register (Offset = 0x5A) [Reset = 0x0000]

R90 is shown in [表 7-22](#).

Return to the [Summary Table](#).

表 7-22. R90 Register Field Descriptions

Bit	Field	Type	Reset	Description
15-8	UNDISCLOSED	R	0x0	Program this field to 0x0.
7	UNDISCLOSED	R/W	0x0	Program this field to 0x0.
6	AUXCLK_DIV_BYP3	R/W	0x0	Set this bit to 1 if AUXCLK_BYP=1, set to 0 otherwise.
5	AUXCLK_DIV_BYP2	R/W	0x0	Set this bit to 1 if AUXCLK_BYP=1, set to 0 otherwise.
4-0	UNDISCLOSED	R/W	0x0	Program this field to 0x0.

8 Application and Implementation

注

以下のアプリケーション情報は、TI の製品仕様に含まれるものではなく、TI ではその正確性または完全性を保証いたしません。個々の目的に対する製品の適合性については、お客様の責任で判断していただくこととなります。お客様は自身の設計実装を検証しテストすることで、システムの機能を確認する必要があります。

8.1 Application Information

8.1.1 SYNC Input Configuration

The SYNC pins support single-ended or differential input in AC or DC coupling mode. The SYNC pins have an internal 50-Ω termination with capacitive ground, which acts as 100-Ω differential.

図 8-1 shows the generic SYNC input circuit recommendation to support all AC or DC, single-ended or differential inputs. Some of the discrete components in fig are just placeholder for individual input signal (single-ended or differential input) and AC or DC coupled input.

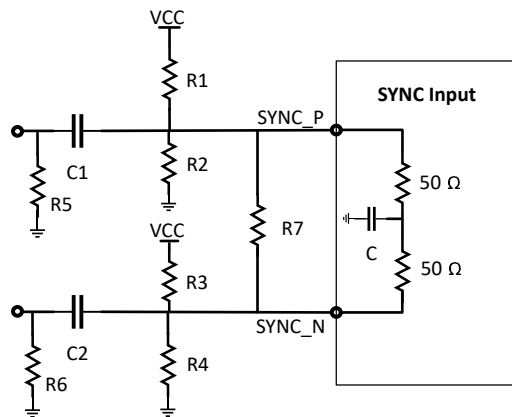


図 8-1. SYNC Input Circuit Recommendations

The following figures show the individual circuit diagram for each configurations:

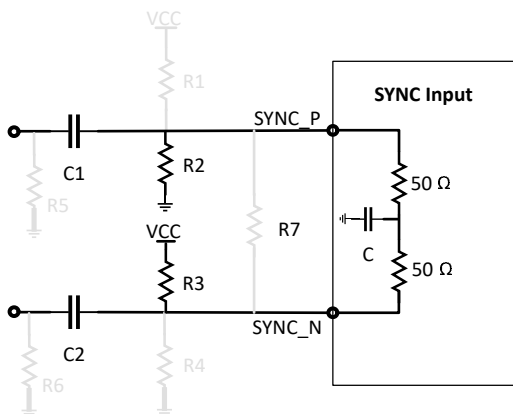


図 8-2. AC-Coupled Differential Input

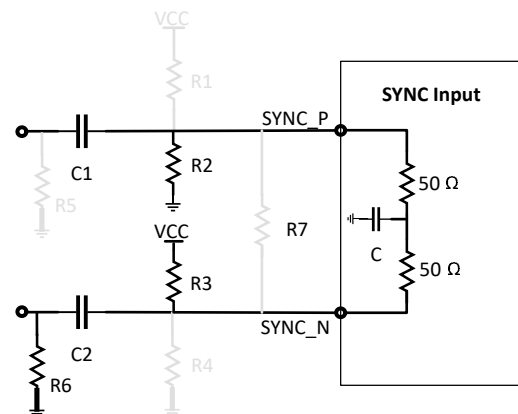


図 8-3. AC-Coupled, Single-Ended Input

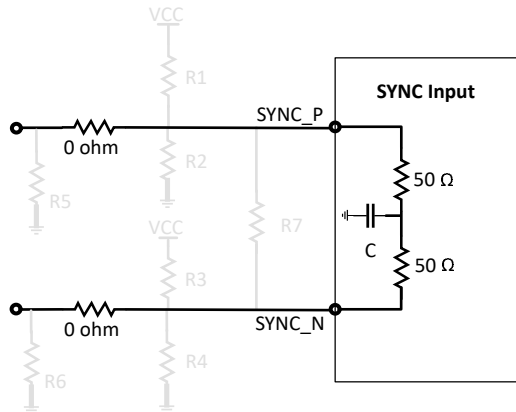


图 8-4. DC-Coupled Differential Input

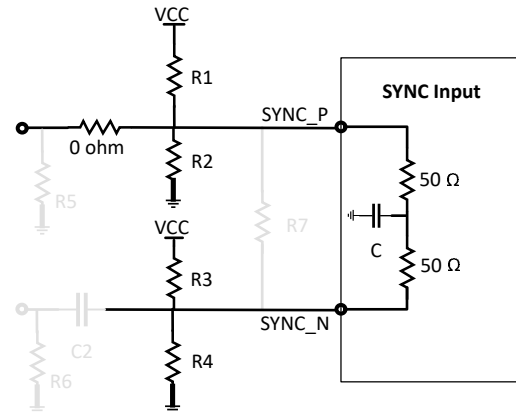


图 8-5. DC-Coupled, Single-Ended Input

- AC coupled differential and single-ended input configurations required the resistor terminations (R2 and R3) to create the VCM at each pin and resistor values must select to maintain greater than 150-mV potential difference between pin P and pin N.
 - As an example, to create the VCM of 1.5 V at pin P and 1.65 V at pin N, with the 2.5 V VCC, set R3 = 550 Ω and R2 = 1 k Ω .
 - For single-ended input configuration, place R6 = 50 Ω to avoid any reflection at complementary input pin.
- DC coupled differential and single-ended input configuration required to have the source common-mode voltage matched with the device input common mode specifications.
 - For single-ended input configuration, keep the R1, R2, R3 and R4 resistors. This method creates the same common-mode voltage at both pins, and the resistive dividers create 75 Ω at pin P and 50- Ω Thevenin's equivalent at pin N.
 - As an example, to have the common-mode voltage of 1.35 V at each pin, set the resistive divider components values to R1 = 130 Ω , R2 = 165 Ω , R3 = 86.6 Ω and R4 = 110 Ω with the 2.5-V VCC.

8.1.2 Treatment of Unused Pins

In many cases, not all pins are used. 表 8-1 lists the recommendations on how to handle these unused pins.

表 8-1. Treatment of Unused or Partially Used Pins

PINS	TREATMENT
All VCC Pins	These pins must always be connected to the power supply. If the block that powers these VCC pins (as implied by the pin name) is not used, then the bypassing can be minimized or eliminated.
SYNC_P / SYNC_N	<ol style="list-style-type: none"> If driving single-ended input, the complementary input must terminate based on セクション 8.1.1 If the SYNC pins are unused, tie the pins to the VCC with 1-kΩ resistor.
CLKIN_P / CLKIN_N	If driving single-ended input, the complementary input must have an AC-coupling capacitor and 50 Ω to ground.
CLKOUTx_P / CLKOUTx_N	If not used, connect a AC-coupling capacitor and 50 Ω to ground.
AUXCLKOUT_P / AUXCLKOUT_N	If these pins are unused, connect the pins to ground with a 1-k Ω resistor.
CLKx_EN	<ol style="list-style-type: none"> If operating device in SPI controlled mode, these pins must tie to VCC with 1-kΩ resistor. If the respective output channels are not using, the pins must connect to ground with 1-kΩ resistor.
MUXSEL, DIVSELx	If these pins are unused, connect these pins to ground with a 1-k Ω resistor.

8.1.3 Current Consumption

The current consumption varies as a function of the setup condition. By adding up all the block currents shown in 表 8-2 users can obtain a reasonable estimate of the current for any setup condition.

表 8-2. Current Consumption per Block

BLOCK		CONDITIONS	CURRENT (mA)	
Device Core		CLK_MUX = Buffer Mode	294	
		CLK_MUX = Divide Mode	260	
SYNC Windowing	Windowing Circuitry	Windowing Circuitry (CLKPOS_CAPTURE_EN=1)	113	
CLKOUT (Per active clock channel)	Core		25	
	Output Buffer	CLKOUTx_EN=1	4+6*CLKOUTx_PWR	
AUXCLKOUT	Core	AUXCLKOUT_EN=1	49	
	Output Buffer		CML(R _p =50Ω)	16+1*AUXCLKOUT_PWR
			LVDS	12

8.2 Typical Application

This application shows the low additive noise impact of LMX1214, when the source LMX2820 exports the 6000-MHz clock output to LMX1214 for local oscillator distribution. This particular setup used a single-ended clock to drive the LMX1214 for the sake of simplicity of connecting two EVMs together, but driving the device differentially is generally recommended.

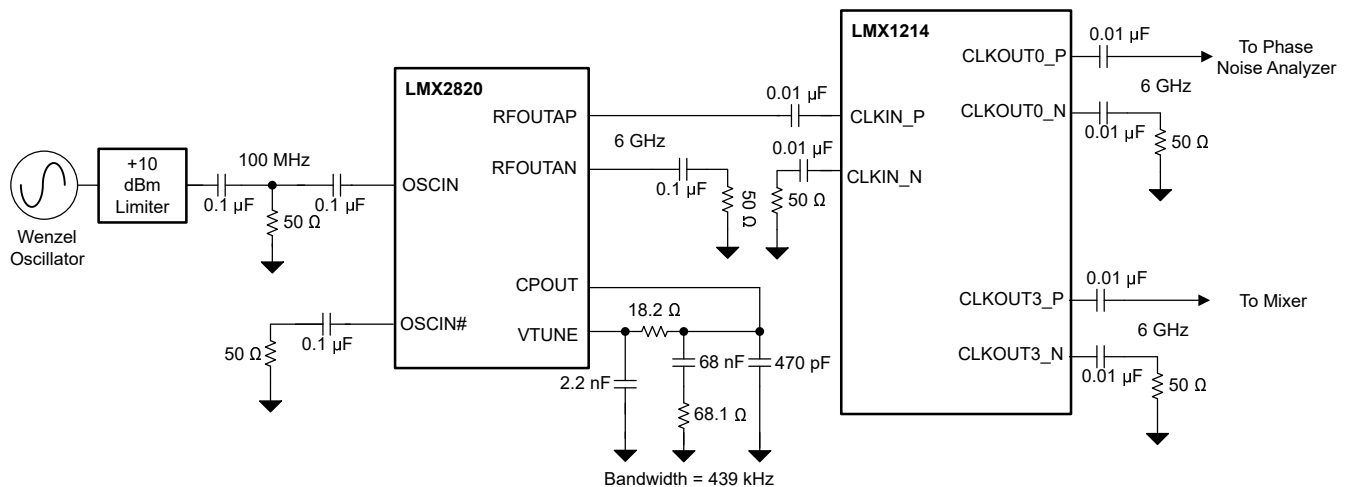


図 8-6. Typical Application Schematic

LMX1214

JAJSSL4A – DECEMBER 2023 – REVISED SEPTEMBER 2024

8.2.1 Design Requirements

Design Parameters shows the design parameters for this example.

If not all outputs are used, TI recommends to compress the layout to minimize trace lengths, especially that of the input trace.

表 8-3. Design Parameters

PARAMETER	VALUE
LMX2820 Input Frequency	100-MHz
LMX2820 Output Frequency	6000-MHz
LMX1214 Input Clock Frequency	6000-MHz
LMX1214 Output Clock Frequency	6000-MHz
LMX1214	Buffer Mode

8.2.2 Detailed Design Procedure

In this example, a 6000-MHz input clock is bypassed to a 6000-MHz output clock. The external components do not change that much based on internal configuration. The TICS Pro software is very useful in calculating the necessary register values and configuring the device.

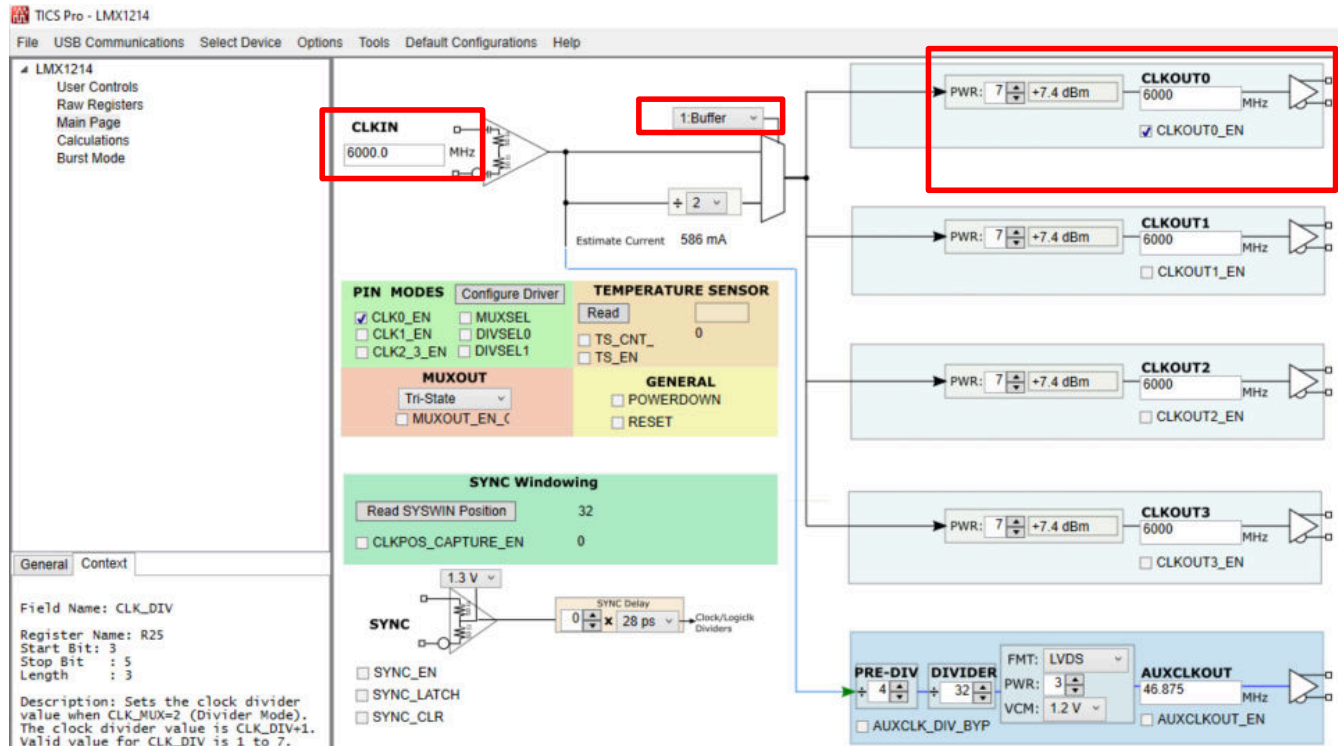
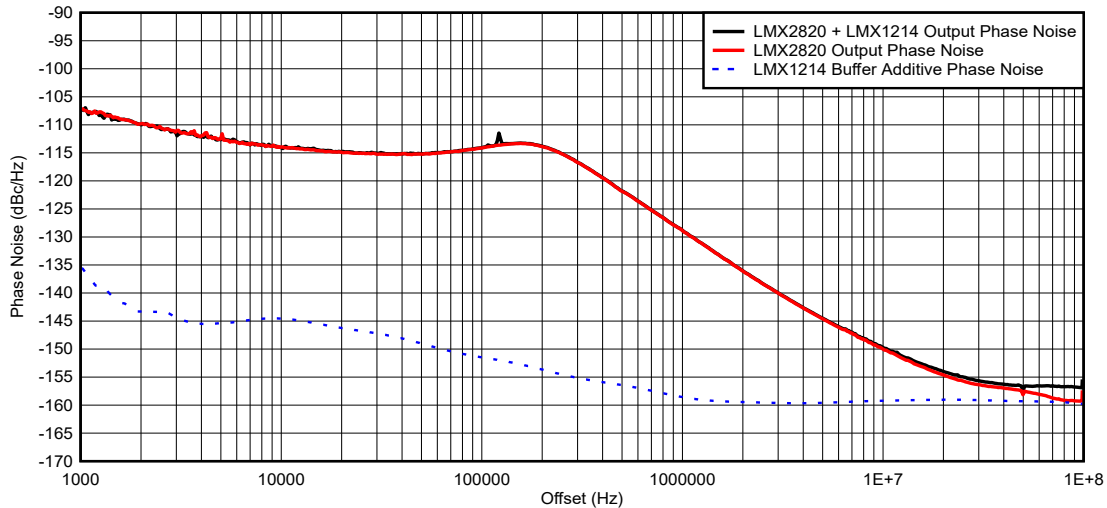


図 8-7. LMX1214 TICS Pro Setup

8.2.3 Application Plots

8-8 the total plot is the sum of the noise of the LMX1214 buffer mode noise and the LMX2820 6000-MHz output. Note that the LMX1214 follow the same noise curve of LMX2820.



8-8. LMX1214 Buffer Output

8.3 Power Supply Recommendations

This device uses a 2.5-V supply for the whole device. A direct connection to a switching power supply likely results in unwanted spurs at the output. Bypassing can be done individually at all the power pins. TI recommends placing smaller capacitors with higher frequency of minimum impedance on the same layer as the device, as close to the pins as possible. The frequencies of nearly all signals in the device are 100 MHz or greater, therefore larger value bypass capacitors with low frequency of minimum impedance are only used for internal LDO stability, and the distance to the device (and the loop inductance of the bypass path) can be larger. Isolate the supply pins for the clocks and the AUXCLKOUT with a small resistor or ferrite bead if both are being used simultaneously. See the *Pin Configuration and Functions* section for additional recommendations for each pin.

注

This device has minimal PSRR due to the low operating voltage and internal filtering by LDOs. Verifying that this device is connected to a low noise supply that does not have excessive spurious noise is important.

8.4 Layout

8.4.1 Layout Guidelines

- If using a single-ended output, terminate the complementary side with AC coupled to 50 Ω so that the impedance for the signal output is same as complementary pin side.
- GND pins on the outer perimeter of the package can be routed on the package back to the DAP.
- Minimize the length of the CLKIN trace for optimal phase noise. Poor matching can degrade the noise floor.
- Verify that the DAP on the device is well-grounded with many vias.
- Use a low loss dielectric material, such as Rogers 4350B, for optimal output power.
- Be aware that if all the outputs are enabled, the current consumption can be high enough to exceed the recommended internal junction temperature of 125°C; a heat sink can be necessary.

8.4.2 Layout Example

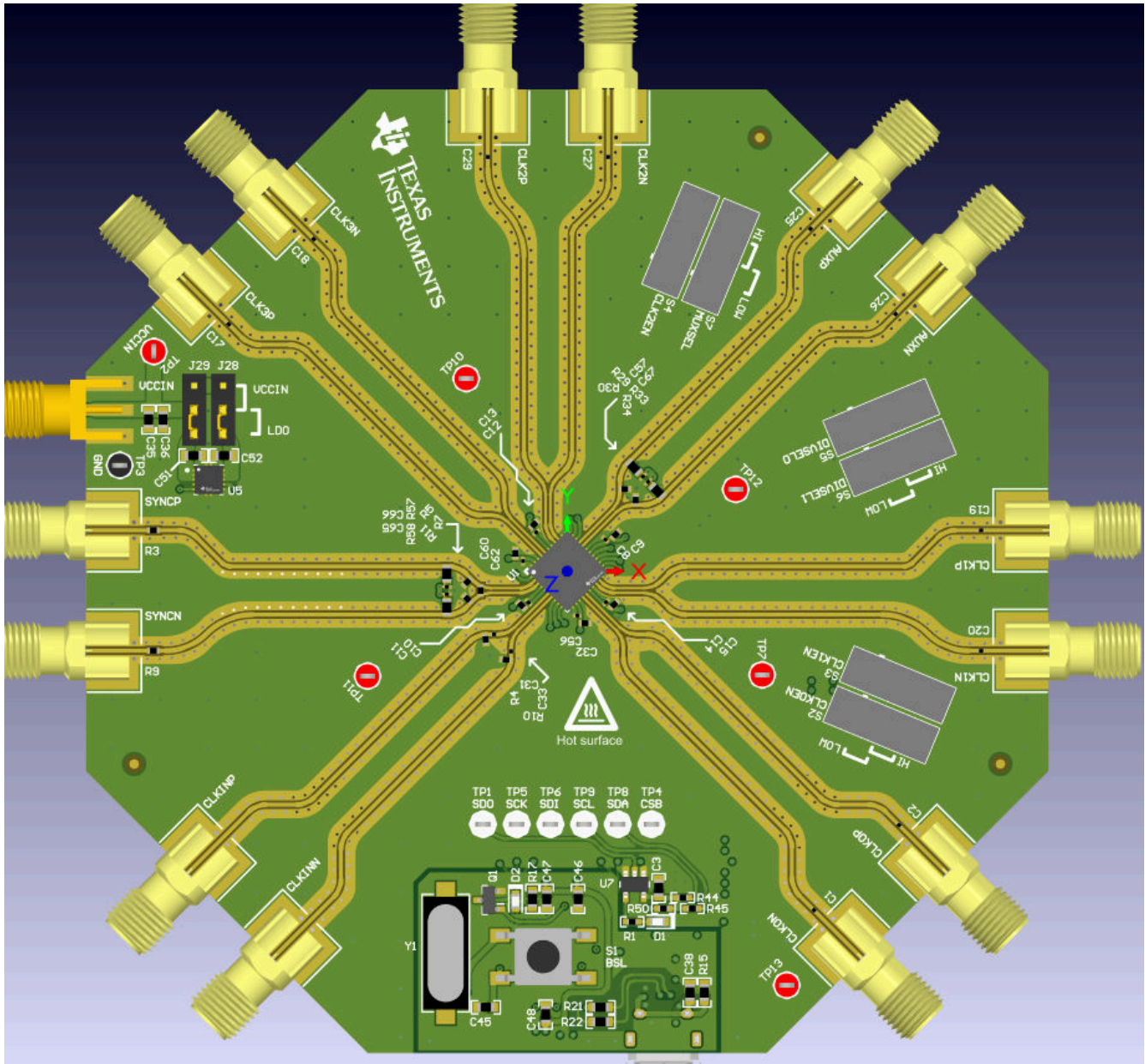


図 8-9. Layout Example

9 Device and Documentation Support

9.1 Device Support

TI offers an extensive line of development tools and software to simulate the device performance and program the device.

表 9-1. Development Tools and Software

TOOL	TYPE	DESCRIPTION
PLLatinum™ Sim	Software	Simulates phase noise in all modes

表 9-1. Development Tools and Software (続き)

TOOL	TYPE	DESCRIPTION
TICS Pro	Software	Programs the device with a user-friendly GUI with interactive feedback and hex register export.

9.2 Documentation Support

9.2.1 Related Documentation

Texas Instruments, [LMX1214 Evaluation Module](#), EVM user's guide

9.3 ドキュメントの更新通知を受け取る方法

ドキュメントの更新についての通知を受け取るには、www.tij.co.jp のデバイス製品フォルダを開いてください。[通知] をクリックして登録すると、変更されたすべての製品情報に関するダイジェストを毎週受け取ることができます。変更の詳細については、改訂されたドキュメントに含まれている改訂履歴をご覧ください。

9.4 サポート・リソース

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9.7 用語集

[テキサス・インスツルメンツ用語集](#) この用語集には、用語や略語の一覧および定義が記載されています。

10 Revision History

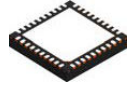
資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

Changes from Revision * (December 2023) to Revision A (August 2024)	Page
• バッファおよびデバイダ モードの出力の最大周波数を、それぞれ 18GHz と 8GHz に変更.....	1
• ドキュメント全体にわたって表、図、相互参照の採番方法を更新.....	1
• Changed max. input frequency spec to 18GHz in buffer mode in <i>Electrical Characteristics clock input</i> section.....	5
• Added output frequency spec in divider-by-2 without sync in <i>Electrical Characteristics Clock Outputs</i> section	5
• Changed max. output frequency spec to 18GHz in buffer mode in <i>Electrical Characteristics Clock Outputs</i> section.....	5
• Changed output power value in <i>Electrical Characteristics Clock Outputs</i> section.....	5
• Test conditions added in <i>Electrical Characteristics leakage power spec</i>	5
• Test conditions added in <i>Electrical Characteristics leakage power spec</i>	5

- Updated the *LMX1214 Flicker Noise at 6 GHz* and *Noise Floor in Buffer Mode* figures.....9
 - added LMX1214 evaluation module user's guide link in Related Documentation.....35
-

11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

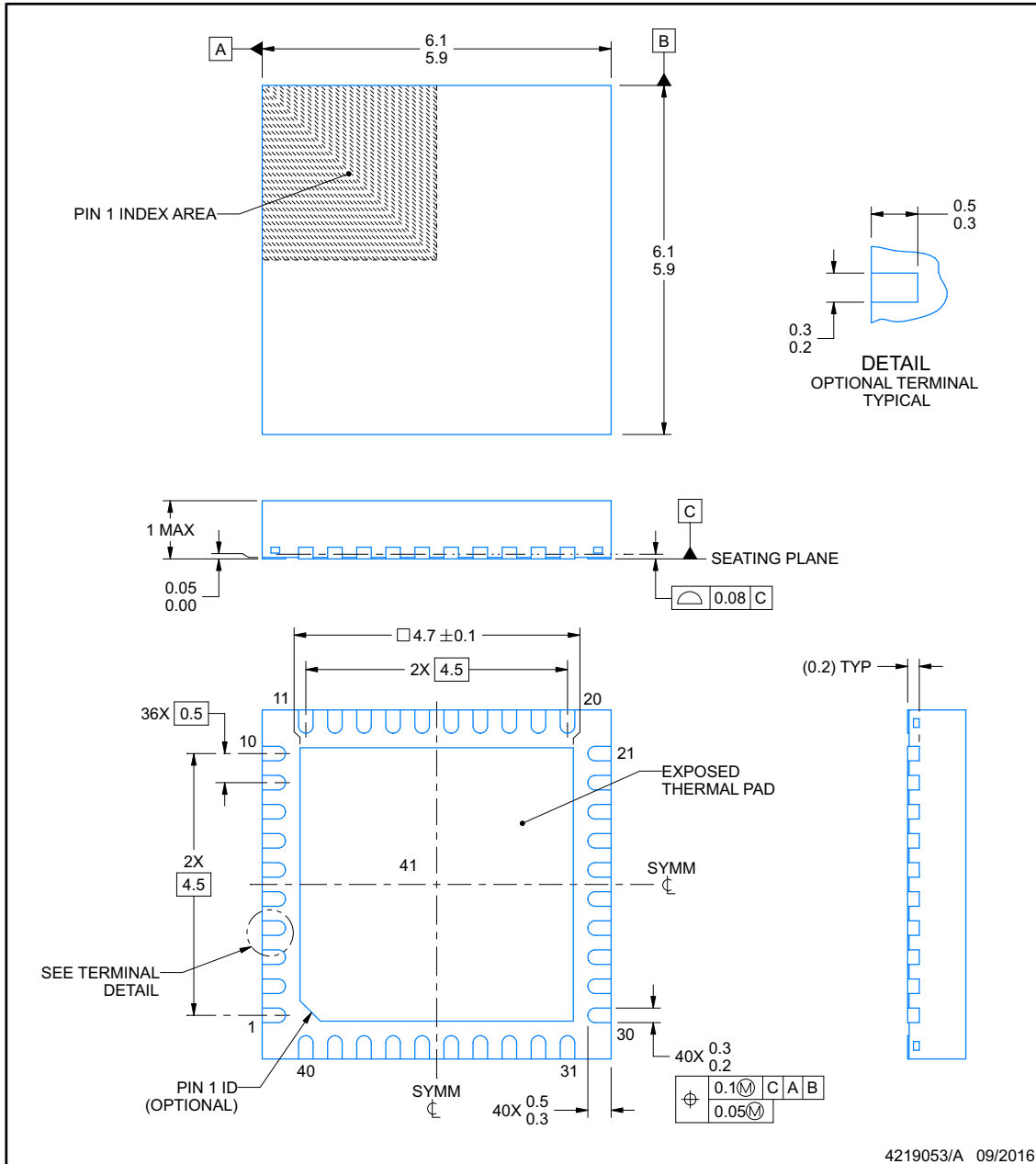


RHA0040C

PACKAGE OUTLINE

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES:

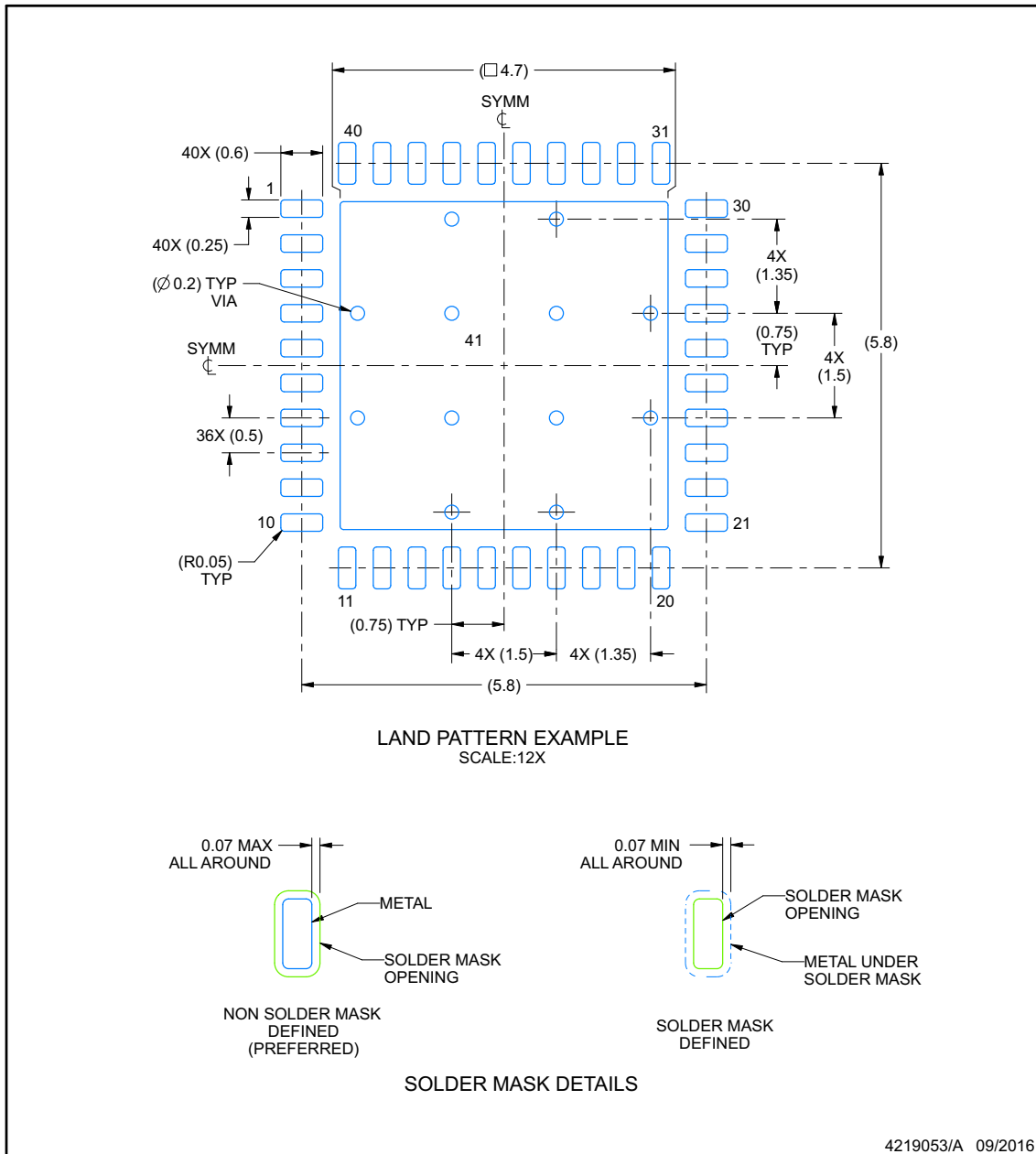
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

EXAMPLE BOARD LAYOUT

RHA0040C

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

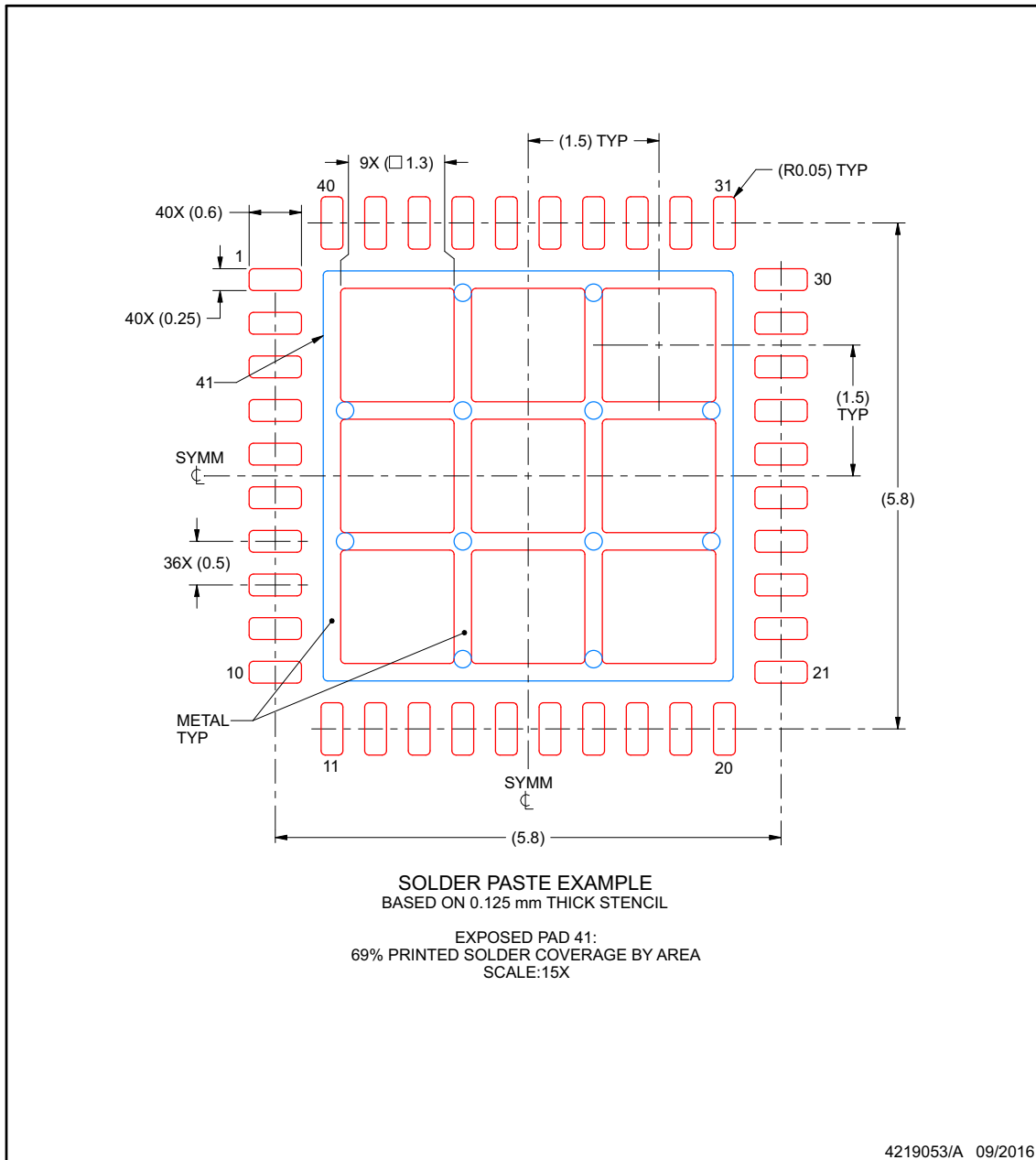
www.ti.com

EXAMPLE STENCIL DESIGN

RHA0040C

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD

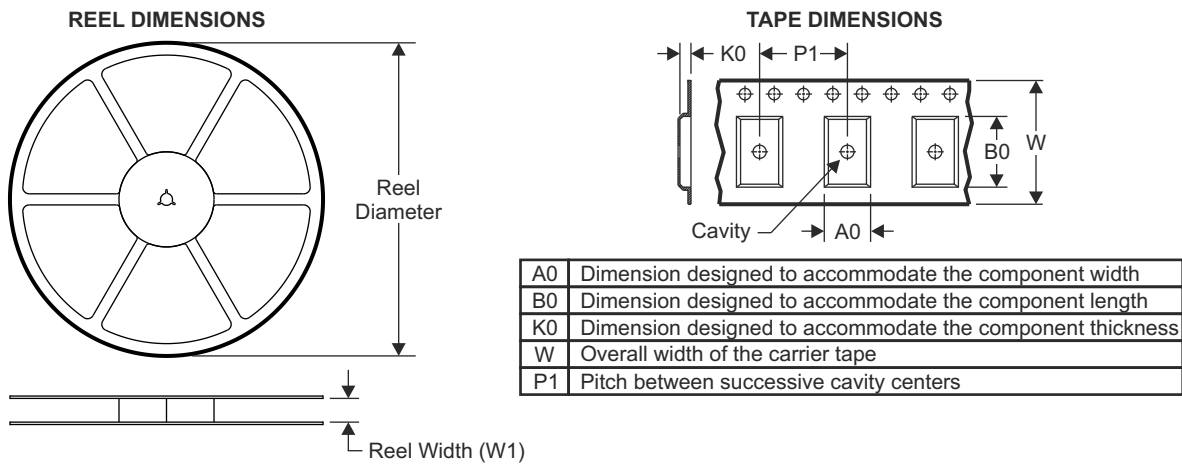


NOTES: (continued)

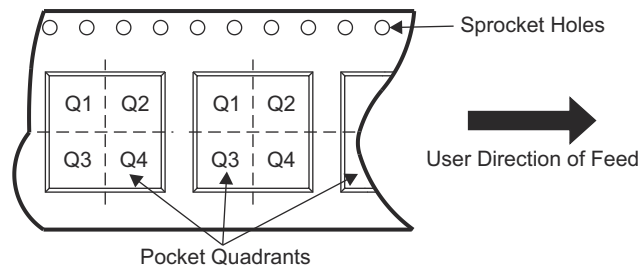
6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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11.1 Tape and Reel Information



QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
MX1214RHAR	VQFN	RHA	40	2500	330	16.4	6.3	6.3	1.1	12	16	Q2
LMX1214RHAT	VQFN	RHA	40	250	180	16.4	6.3	6.3	1.1	12	16	Q2

TAPE AND REEL BOX DIMENSIONS



Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMX1214RHAR	VQFN	RHA	40	2500	367	367	38
LMX1214RHAT	VQFN	RHA	40	250	210	185	35

重要なお知らせと免責事項

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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LMX1214RHAR	ACTIVE	VQFN	RHA	40	2500	RoHS & Green	NIPDAUAG	Level-3-260C-168 HR	-40 to 85	LMX1214	Samples
LMX1214RHAT	ACTIVE	VQFN	RHA	40	250	RoHS & Green	NIPDAUAG	Level-3-260C-168 HR	-40 to 85	LMX1214	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMX1214RHAR	VQFN	RHA	40	2500	330.0	16.4	6.3	6.3	1.1	12.0	16.0	Q2
LMX1214RHAT	VQFN	RHA	40	250	180.0	16.4	6.3	6.3	1.1	12.0	16.0	Q2

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMX1214RHAR	VQFN	RHA	40	2500	367.0	367.0	38.0
LMX1214RHAT	VQFN	RHA	40	250	210.0	185.0	35.0

GENERIC PACKAGE VIEW

RHA 40

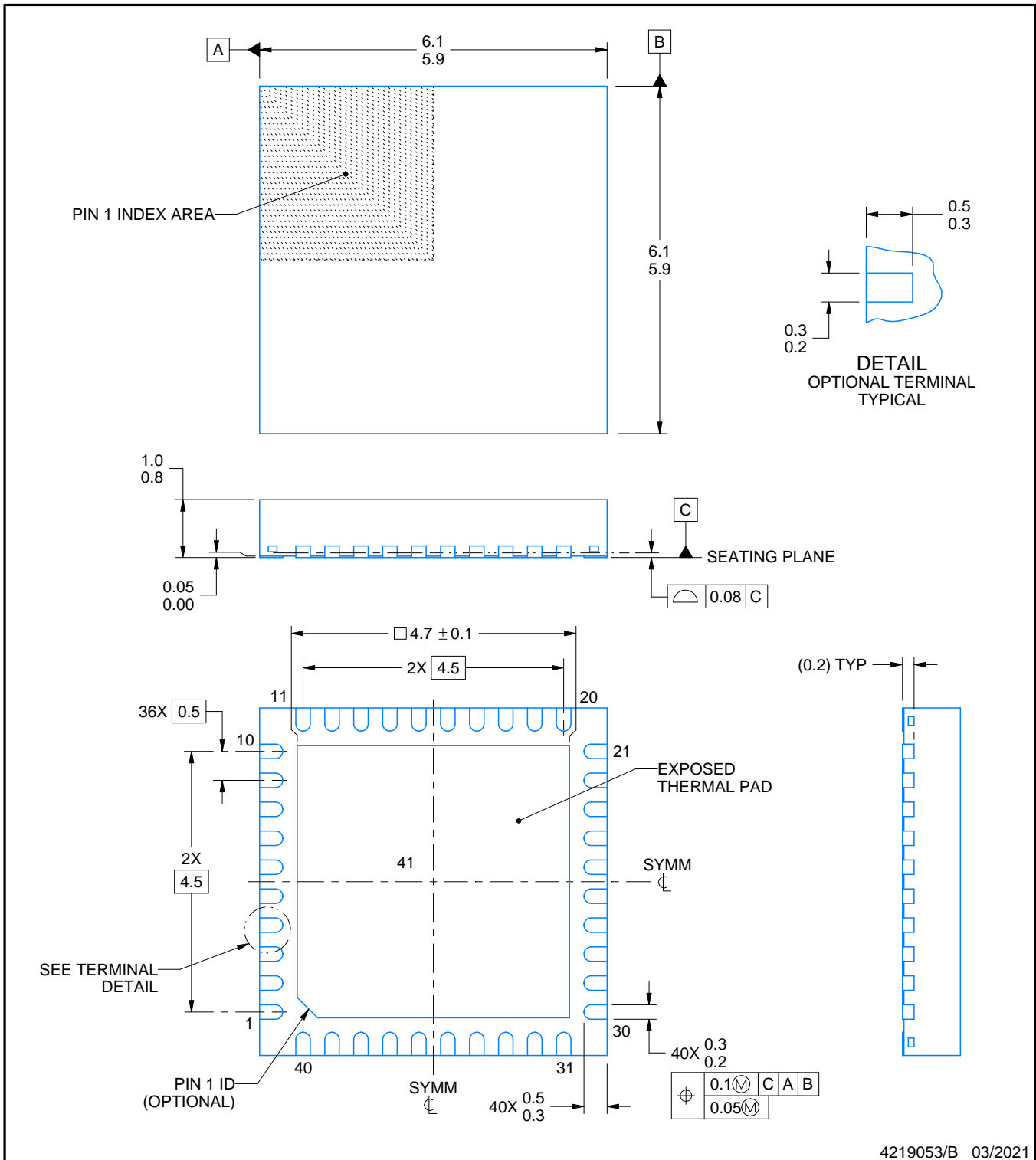
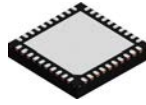
VQFN - 1 mm max height

6 x 6, 0.5 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD

This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.





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NOTES:

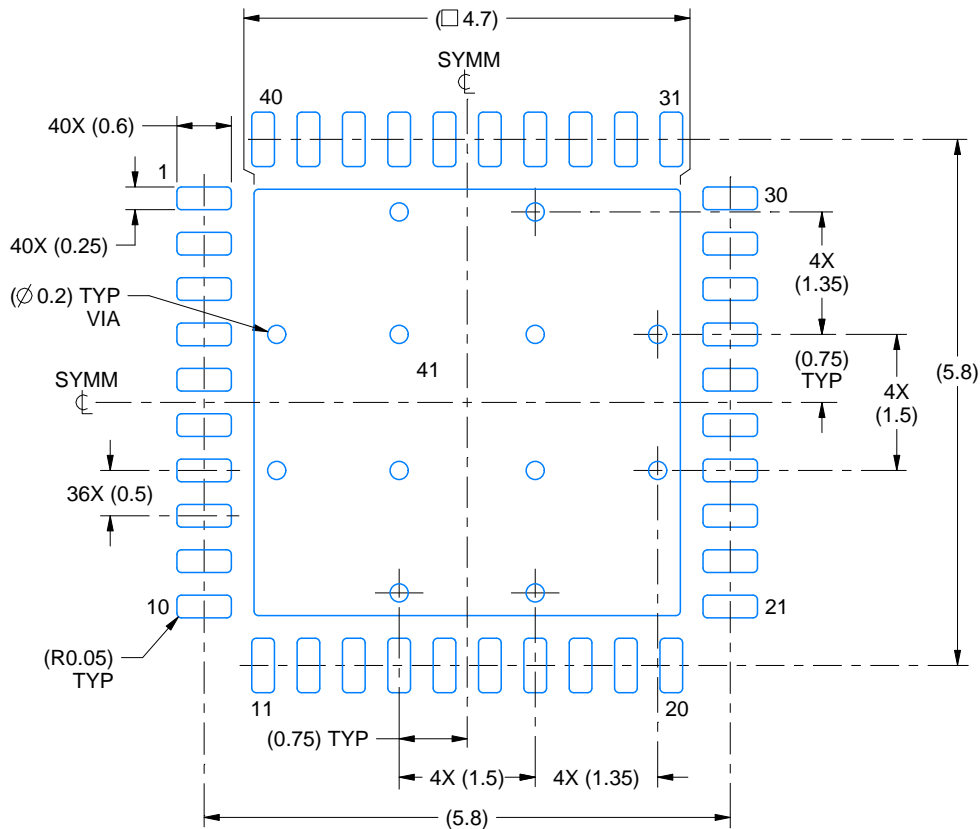
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

EXAMPLE BOARD LAYOUT

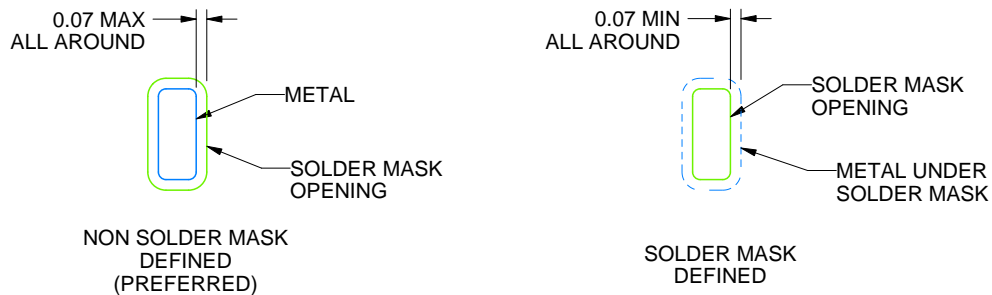
RHA0040C

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE
SCALE:12X



SOLDER MASK DETAILS

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NOTES: (continued)

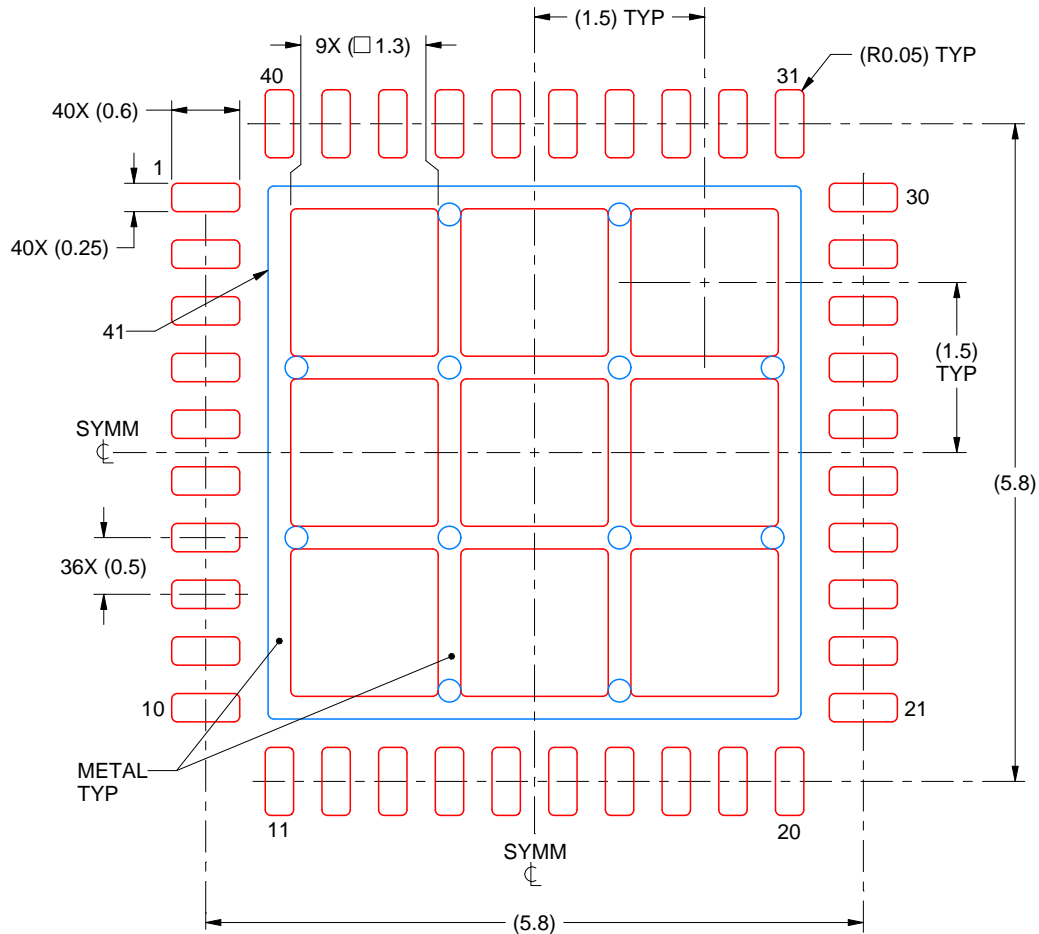
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/sluea271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

EXAMPLE STENCIL DESIGN

RHA0040C

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD 41:
69% PRINTED SOLDER COVERAGE BY AREA
SCALE:15X

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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