

OPA2333-Q1 車載用、1.8V、マイクロパワー、CMOS、ゼロドリフトのオペアンプ

1 特長

- 車載アプリケーション用に AEC 認定済み
 - 温度グレード 1: $-40^{\circ}\text{C} \sim +125^{\circ}\text{C}$ 、 T_A
- 低いオフセット電圧: $23\mu\text{V}$ 以下
- 0.01Hz~10Hzのノイズ: $1.1\mu\text{V}_{PP}$
- 静止電流: $17\mu\text{A}$
- 単一電源動作
- 電源電圧: $1.8\text{V} \sim 5.5\text{V}$
- レール・ツー・レール入出力
- パッケージ: 8ピン SOIC および VSSOP

2 アプリケーション

- ポンプ
- 位置センサ
- 車両占有検出センサ
- ブレーキ・システム
- エアバッグ

3 概要

OPA2333-Q1 CMOS オペアンプは、独自の自動較正技術を採用することで、非常に低いオフセット電圧 ($10\mu\text{V}$ 以下) を実現しながら、時間経過および温度変動に対するドリフトをほぼゼロに抑えています。これらの小型、高精度、低静止電流のオペアンプは、レールを 100mV 上回る同相範囲を持つ高インピーダンス入力と、レールの 50mV 以内までスイングするレール・ツー・レール出力を備えています。最低 $1.8\text{V} (\pm 0.9\text{V})$ 、最高 $5.5\text{V} (\pm 2.75\text{V})$ のシングルまたはデュアル電源が使用できます。このデバイスは、低電圧の単一電源で動作するように最適化されています。

OPA2333-Q1 は、従来の相補入力段に見られるクロスオーバーを持たず、同相除去比 (CMRR) が優れています。この設計により、アナログ/デジタル・コンバータ (ADC) の微分直線性の低下がなく、優れた ADC 駆動性能が得られます。

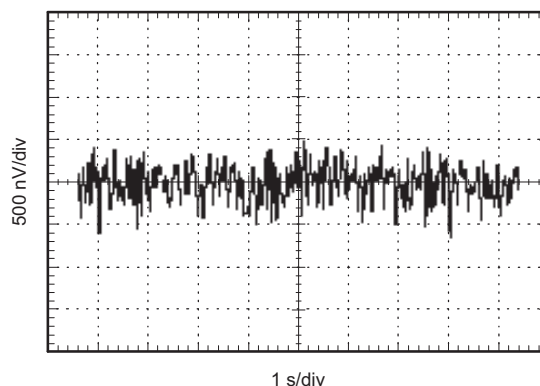
OPA2333-Q1 は、 $-40^{\circ}\text{C} \sim +125^{\circ}\text{C}$ で動作が規定されています。

製品情報⁽¹⁾

型番	パッケージ	本体サイズ(公称)
OPA2333-Q1	SOIC (8)	4.90mm×3.91mm
	VSSOP (8)	3.00mm×3.00mm

(1) 利用可能なすべてのパッケージについては、このデータシートの末尾にあるパッケージ・オプションについての付録を参照してください。

0.1Hz~10Hzのノイズ



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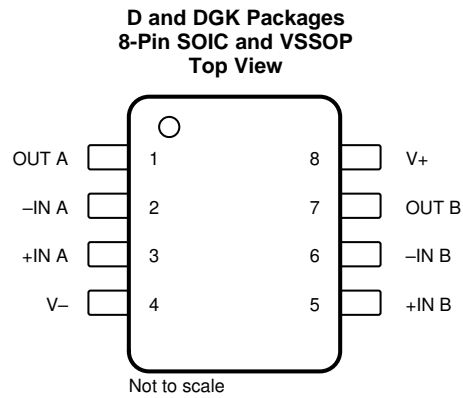
4 改訂履歴

Revision A (June 2010) から Revision B に変更

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• 「製品情報」表、「ピンの機能」表、「ESD 定格」表、「推奨動作条件」表、「熱に関する情報」表、「機能説明」セクション、「デバイスの機能モード」セクション、「アプリケーションと実装」セクション、「電源に関する推奨事項」セクション、「レイアウト」セクション、「デバイスおよびドキュメントのサポート」セクション、「メカニカル、パッケージ、および注文情報」セクションを追加	1
• Changed input offset voltage (over full temp range) from 22 μV to 15 μV in <i>Electrical Characteristics</i> table	5
• Added maximum value of 0.05 $\mu\text{V}/^\circ\text{C}$ to the V_{OS} drift parameter in the <i>Electrical Characteristics</i> table	5
• Deleted <i>Thermal resistance</i> parameter from <i>Electrical Characteristics</i> table	5

5 Pin Configuration and Functions



Pin Functions

PIN		I/O	DESCRIPTION
NO.	NAME		
1	OUT A	O	Channel A output
2	-IN A	I	Channel A inverting input
3	+IN A	I	Channel A noninverting input
4	V-	—	Negative (lowest) supply voltage
5	+IN B	I	Channel B noninverting input
6	-IN B	I	Channel B inverting input
7	OUT B	O	Channel B output
8	V+	—	Positive (highest) supply voltage

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
V _{CC}	Supply voltage		7	V
V _I	Input voltage, signal input pins ⁽²⁾	–0.3	(V+) + 0.3	V
	Output short-circuit ⁽³⁾	Continuous		
T _A	Operating free-air temperature	–40	125	°C
T _J	Operating virtual-junction temperature		150	°C
T _{stg}	Storage temperature	–65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) Input terminals are diode clamped to the power-supply rails. Input signals that can swing more than 0.3 V beyond the supply rails should be current limited to 10 mA or less.
- (3) Short circuit to ground, one amplifier per package

6.2 ESD Ratings

		VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 ⁽¹⁾ HBM ESD classification level 2	±2000
		Charged-device model (CDM), per AEC Q100-011 CDM ESD Classification Level C6	±1000

- (1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V _S	Specified supply voltage	1.8		5.5	V
T _A	Specified free-air temperature	–40		125	°C

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		OPA2333-Q1		UNIT
		D (SOIC)	DGK (VSSOP)	
		8 PINS	8 PINS	
R _{θJA}	Junction-to-ambient thermal resistance	124.0	180.3	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	73.7	48.1	°C/W
R _{θJB}	Junction-to-board thermal resistance	64.4	100.9	°C/W
Ψ _{JT}	Junction-to-top characterization parameter	18.0	2.4	°C/W
Ψ _{JB}	Junction-to-board characterization parameter	63.9	99.3	°C/W
R _{θJC(bot)}	Junction-to-case (bottom) thermal resistance	N/A	N/A	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

6.5 Electrical Characteristics: $V_S = 1.8\text{ V to }5.5\text{ V}$

At $T_A = 25^\circ\text{C}$, $R_L = 10\text{ k}\Omega$ connected to $V_S/2$, $V_{CM} = V_S/2$, and $V_{OUT} = V_S/2$ (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
OFFSET VOLTAGE						
V_{OS}	Input offset voltage	$V_S = 5\text{ V}$		2	10	μV
		$V_S = 5\text{ V}$, $T_A = -40^\circ\text{C to }+125^\circ\text{C}$			15	μV
dV_{OS}/dT	V_{OS} drift	$V_S = 5\text{ V}$, $T_A = -40^\circ\text{C to }125^\circ\text{C}$		0.02	0.05	$\mu\text{V}/^\circ\text{C}$
PSRR	Power-supply rejection ratio	$V_S = 1.8\text{ V to }5.5\text{ V}$, $T_A = -40^\circ\text{C to }+125^\circ\text{C}$		1	6	$\mu\text{V/V}$
	Long-term stability ⁽¹⁾			1 ⁽¹⁾		μV
	Channel separation, dc			0.1		$\mu\text{V/V}$
INPUT BIAS CURRENT						
I_B	Input bias current			± 70	± 200	μA
		$T_A = -40^\circ\text{C to }+125^\circ\text{C}$		± 150		μA
I_{OS}	Input offset current			± 140	± 400	μA
NOISE						
	Input voltage noise	$f = 0.01\text{ Hz to }1\text{ Hz}$		0.3		μV_{PP}
		$f = 0.1\text{ Hz to }10\text{ Hz}$		1.1		μV_{PP}
i_n	Input current noise	$f = 10\text{ Hz}$		100		$\text{fA}/\sqrt{\text{Hz}}$
INPUT VOLTAGE						
V_{CM}	Common-mode supply voltage		$(V-) - 0.1$		$(V+) + 0.1$	V
CMRR	Common-mode rejection ratio	$(V-) - 0.1\text{ V} < V_{CM} < (V+) + 0.1\text{ V}$, $T_A = -40^\circ\text{C to }+125^\circ\text{C}$	102	130		dB
INPUT CAPACITANCE						
	Differential			2		pF
	Common-mode			4		pF
OPEN-LOOP GAIN						
A_{OL}	Open-loop voltage gain	$(V-) + 100\text{ mV} < V_O < (V+) - 100\text{ mV}$, $R_L = 10\text{ k}\Omega$, $T_A = -40^\circ\text{C to }+125^\circ\text{C}$	104	130		dB
FREQUENCY RESPONSE						
GBW	Gain-bandwidth product	$C_L = 100\text{ pF}$		350		kHz
SR	Slew rate	$G = 1$		0.16		$\text{V}/\mu\text{s}$
OUTPUT						
	Voltage output swing from rail	$R_L = 10\text{ k}\Omega$		30	50	mV
		$R_L = 10\text{ k}\Omega$, $T_A = -40^\circ\text{C to }+125^\circ\text{C}$			85	mV
I_{SC}	Short-circuit current			± 5		mA
C_L	Capacitive load drive					
	⁽²⁾ Open-loop output impedance	$f = 350\text{ kHz}$, $I_O = 0\text{ A}$		2		k Ω
POWER SUPPLY						
I_Q	Quiescent current per amplifier	$I_O = 0\text{ A}$		17	25	μA
		$I_O = 0\text{ A}$, $T_A = -40^\circ\text{C to }+125^\circ\text{C}$			30	μA
	Turn-on time	$V_S = 5\text{ V}$		100		μs

(1) 300-hour life test at 150°C demonstrated randomly distributed variation of approximately $1\ \mu\text{V}$.

(2) See the [Typical Characteristics](#) section.

6.6 Typical Characteristics

At $T_A = 25^\circ\text{C}$, $V_S = 5\text{ V}$, and $C_L = 0\text{ pF}$ (unless otherwise noted)

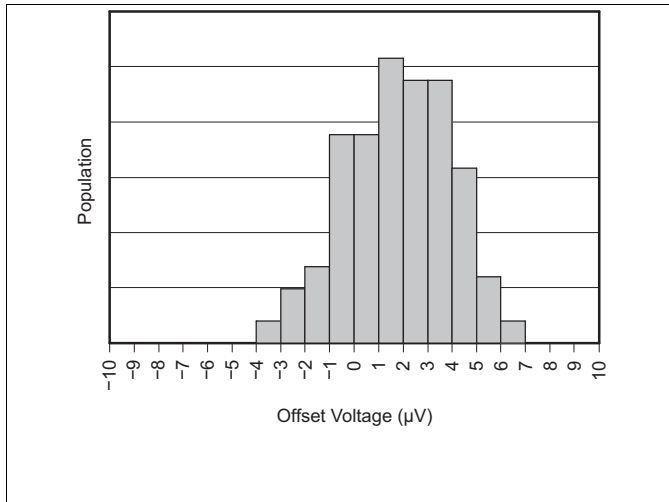


Figure 1. Offset Voltage Production Distribution

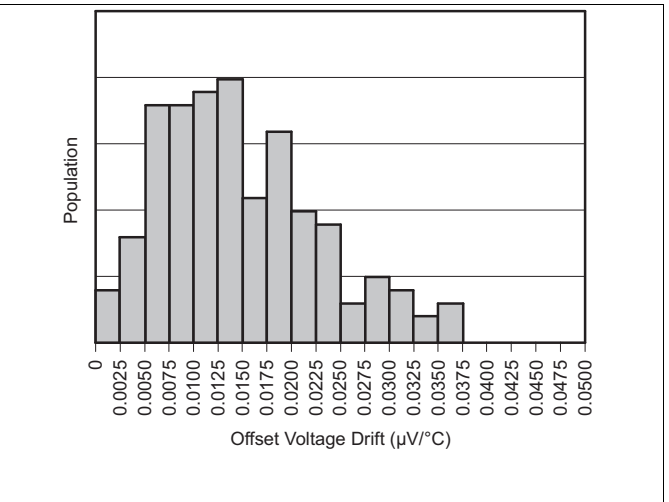


Figure 2. Offset Voltage Drift Production Distribution

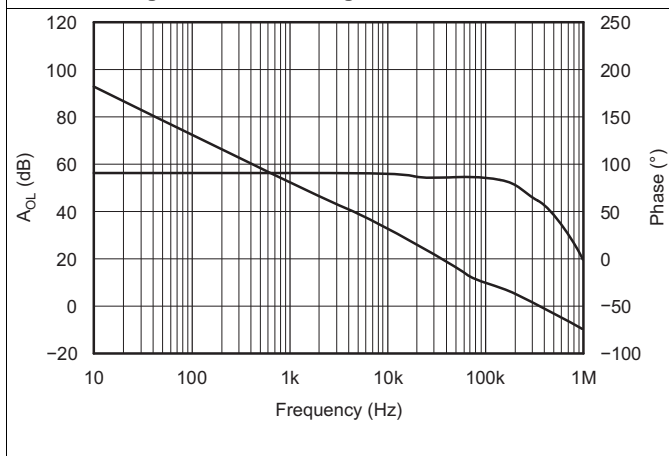


Figure 3. Open-Loop Gain vs Frequency

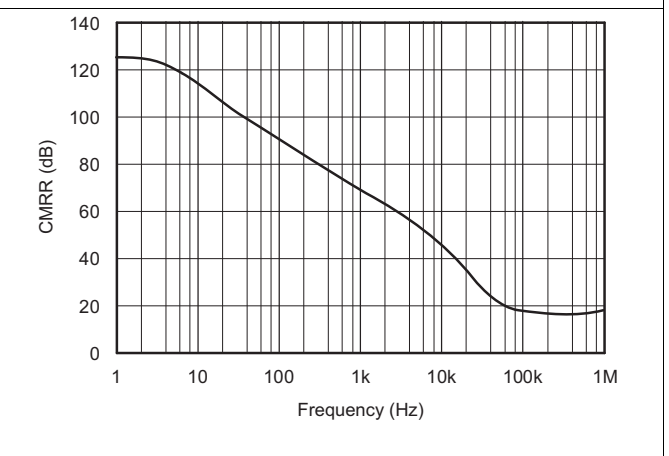


Figure 4. Common-Mode Rejection Ratio vs Frequency

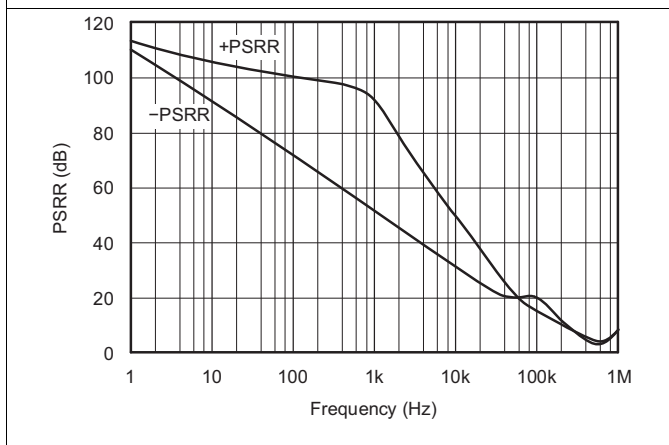


Figure 5. Power-Supply Rejection Range vs Frequency

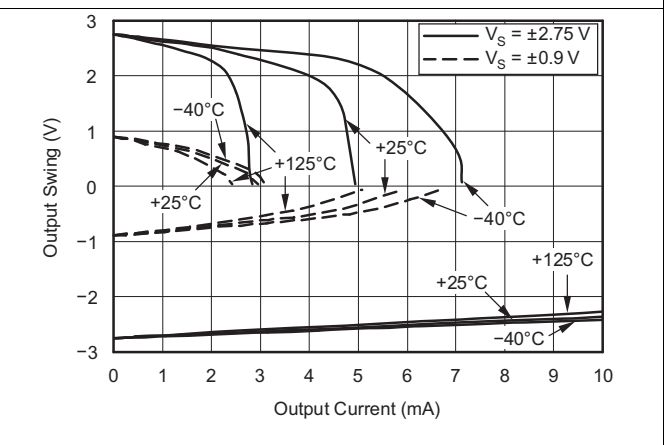


Figure 6. Output Voltage Swing vs Output Current

Typical Characteristics (continued)

At $T_A = 25^\circ\text{C}$, $V_S = 5\text{ V}$, and $C_L = 0\text{ pF}$ (unless otherwise noted)

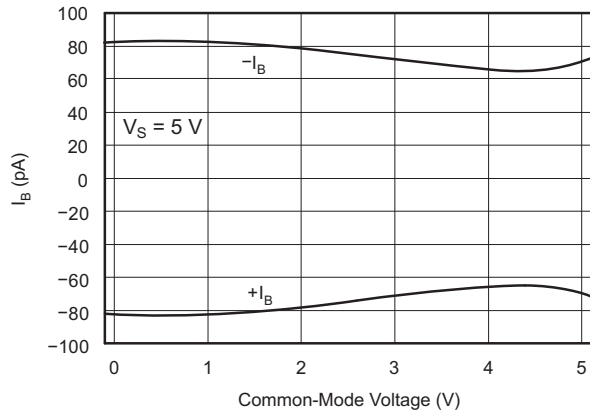


Figure 7. Input Bias Current vs Common-Mode Voltage

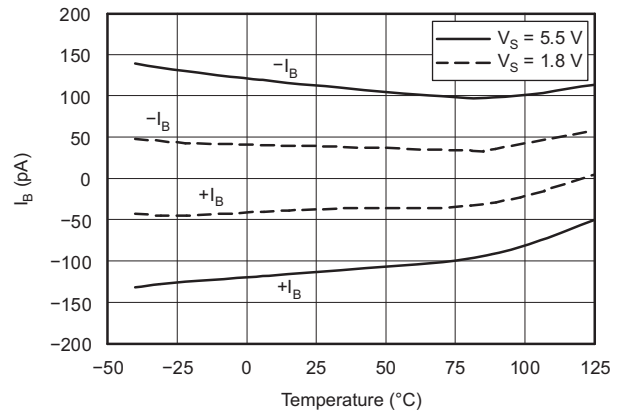


Figure 8. Input Bias Current vs Temperature

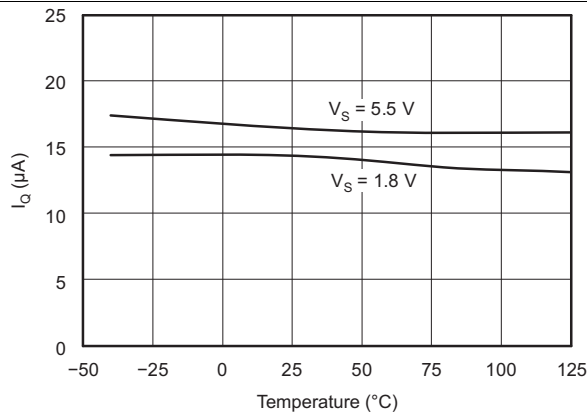


Figure 9. Quiescent Current vs Temperature

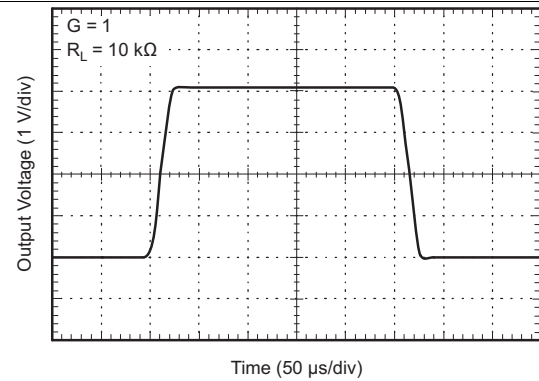


Figure 10. Large-Signal Step Response

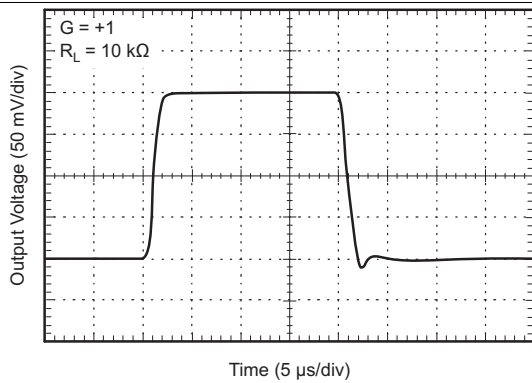


Figure 11. Small-Signal Step Response

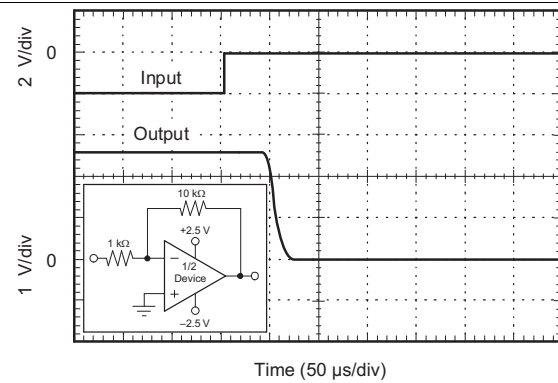


Figure 12. Positive Over-Voltage Recovery

Typical Characteristics (continued)

At $T_A = 25^\circ\text{C}$, $V_S = 5\text{ V}$, and $C_L = 0\text{ pF}$ (unless otherwise noted)

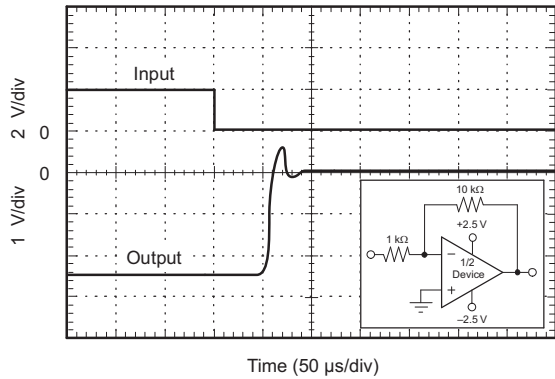


Figure 13. Negative Over-Voltage Recovery

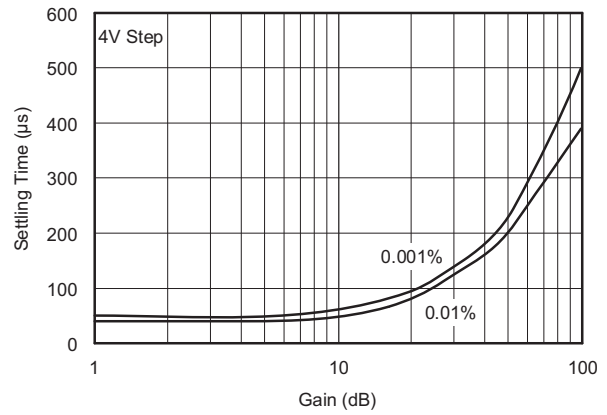


Figure 14. Settling Time vs Closed-Loop Gain

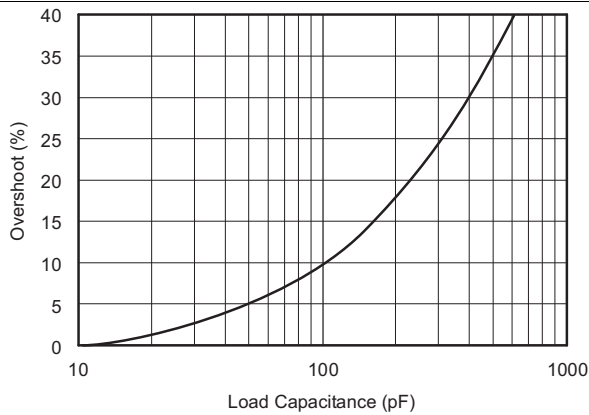


Figure 15. Small-Signal Overshoot vs Load Capacitance

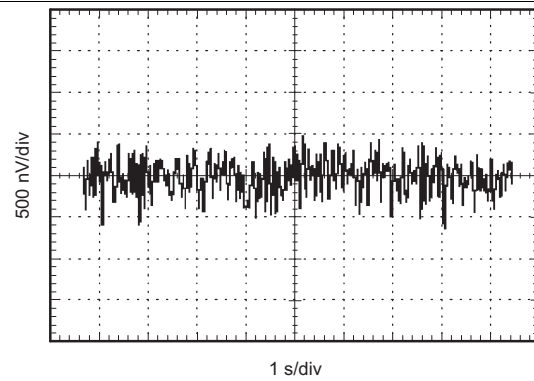


Figure 16. 0.1-Hz to 10-Hz Noise

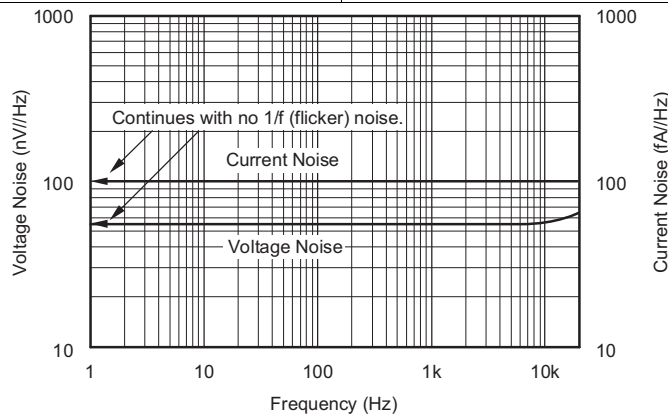


Figure 17. Current and Voltage Spectral Density vs Frequency

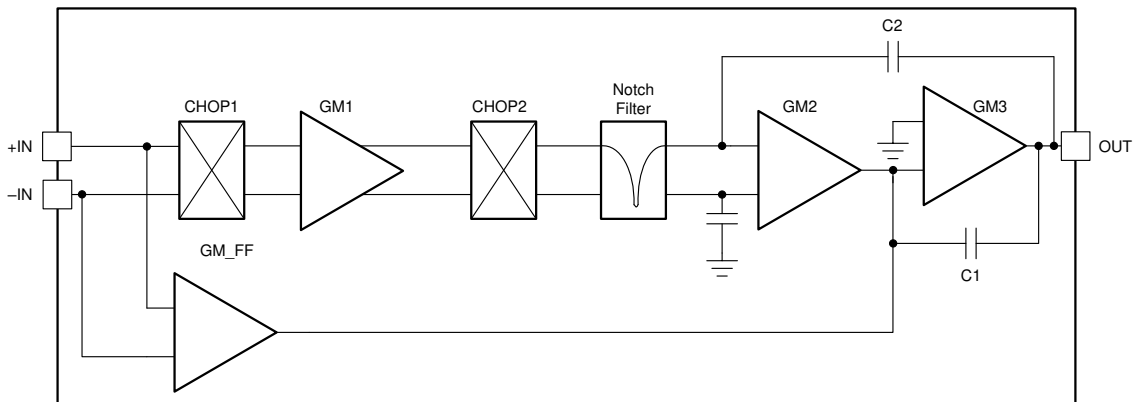
7 Detailed Description

7.1 Overview

The OPA2333-Q1 device is a zero-drift, low-power, rail-to-rail input and output operational amplifier. The device operates from 1.8 V to 5.5 V, is unity-gain stable, and is designed for a wide range of general-purpose applications. The zero-drift architecture provides ultra-low offset voltage and near-zero offset voltage drift.

The OPA2333-Q1 is unity-gain stable and free from unexpected output phase reversal. The device uses a proprietary auto-calibration technique to provide low offset voltage and very low drift over time and temperature.

7.2 Functional Block Diagram

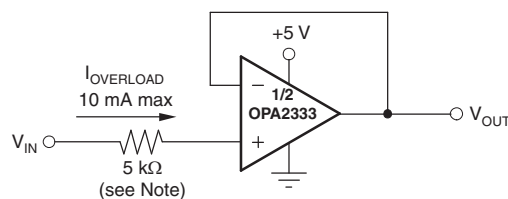


7.3 Feature Description

7.3.1 Rail-to-Rail Input Voltage

The OPA2333-Q1 input common-mode voltage range extends 0.1 V beyond the supply rails. The device is designed to cover the full range without the troublesome transition region found in some other rail-to-rail amplifiers.

Normally, input bias current is approximately 70 pA; however, input voltages exceeding the power supplies can cause excessive current to flow into or out of the input pins. Momentary voltages greater than the power supply can be tolerated if the input current is limited to 10 mA. This limitation is easily accomplished with an input resistor (see Figure 18).



NOTE: A current-limiting resistor required if the input voltage exceeds the supply rails by ≥ 0.5 V.

Figure 18. Input Current Protection

7.3.2 Internal Offset Correction

The OPA2333-Q1 op amps use an auto-calibration technique with a time-continuous 350-kHz op amp in the signal path. This amplifier is zero corrected every 8 μ s using a proprietary technique. At power up, the amplifier requires approximately 100 μ s to achieve the specified V_{OS} accuracy. This design has no aliasing or flicker noise.

7.4 Device Functional Modes

The OPA2333-Q1 has a single functional mode. The device is powered on as long as the power-supply voltage is between 1.8 V (± 0.9 V) and 5.5 V (± 2.75 V).

8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The OPA2333-Q1 is a unity-gain stable, precision operational amplifier with very low offset voltage drift. The device is also free from output phase reversal. Applications with noisy or high-impedance power supplies require decoupling capacitors close to the device power-supply pins. In most cases, 0.1- μ F capacitors are adequate.

8.1.1 Achieving Output Swing to the Op Amp Negative Rail

Some applications require output voltage swings from 0 V to a positive full-scale voltage (such as 2.5 V) with excellent accuracy. With most single-supply op amps, problems arise when the output signal approaches 0 V, near the lower output swing limit of a single-supply op amp. A good single-supply op amp may swing close to single-supply ground, but does not reach ground. The output of the OPA2333-Q1 can be made to swing to ground or slightly below on a single-supply power source. To do so requires the use of another resistor and an additional, more negative, power supply than the op amp negative supply. A pulldown resistor may be connected between the output and the additional negative supply to pull the output down below the value that the output would otherwise achieve (see [Figure 19](#)).

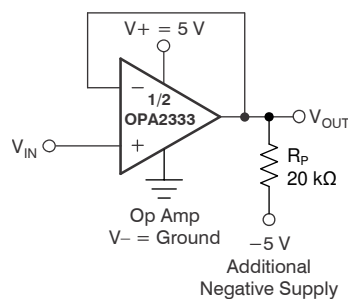


Figure 19. V_{OUT} Range to Ground

The OPA2333-Q1 has an output stage that allows the output voltage to be pulled to its negative supply rail, or slightly below, using the technique previously described. This technique only works with some types of output stages. The OPA2333-Q1 has been characterized to perform with this technique; however, the recommended resistor value is approximately 20 k Ω .

NOTE

This configuration increases the current consumption by several hundreds of microamps.

Accuracy is excellent down to 0 V and as low as -2 mV. Limiting and nonlinearity occurs below -2 mV, but excellent accuracy returns as the output is again driven above -2 mV. Lowering the resistance of the pulldown resistor allows the op amp to swing even further below the negative rail. Resistances as low as 10 k Ω can be used to achieve excellent accuracy down to -10 mV.

8.2 Typical Application

8.2.1 High-Side Voltage-to-Current (V-I) Converter

The circuit shown in Figure 20 is a high-side voltage-to-current (V-I) converter. It translates an input voltage of 0 V to 2 V to an output current of 0 mA to 100 mA. Figure 21 shows the measured transfer function for this circuit. The low offset voltage and offset drift of the OPA333-Q1 device facilitate excellent dc accuracy for the circuit.

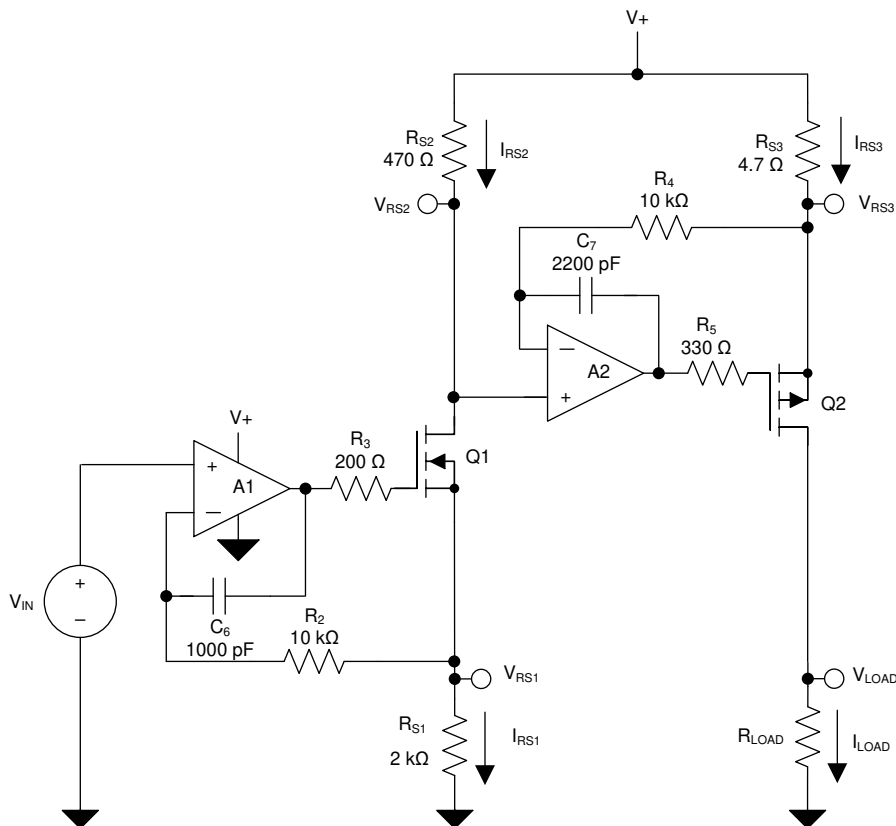


Figure 20. High-Side Voltage-to-Current (V-I) Converter

Typical Application (continued)

8.2.1.1 Design Requirements

The design requirements are as follows:

- Supply voltage: 5 V dc
- Input: 0 V to 2 V dc
- Output: 0 mA to 100 mA dc

8.2.1.2 Detailed Design Procedure

The V-I transfer function of the circuit is based on the relationship between the input voltage, V_{IN} , and the three current sensing resistors, R_{S1} , R_{S2} , and R_{S3} . The relationship between V_{IN} and R_{S1} determines the current that flows through the first stage of the design. The current gain from the first stage to the second stage is based on the relationship between R_{S2} and R_{S3} .

For a successful design, pay close attention to the dc characteristics of the operational amplifier chosen for the application. To meet the performance goals, this application benefits from an operational amplifier with low offset voltage, low temperature drift, and rail-to-rail output. The OPA2333-Q1 CMOS operational amplifier is a high-precision, 5- μ V offset, 0.05- μ V/ $^{\circ}$ C drift amplifier optimized for low-voltage, single-supply operation with an output swing to within 50 mV of the positive rail. The OPA2333-Q1 uses chopping techniques to provide low initial offset voltage and near-zero drift over time and temperature. Low offset voltage and low drift reduce the offset error in the system, making these devices appropriate for precise dc control. The rail-to-rail output stage of the OPA2333-Q1 makes sure that the output swing of the operational amplifier is able to fully control the gate of the MOSFET devices within the supply rails.

A detailed error analysis, design procedure, and additional measured results are given in the [High-Side V-I Converter reference design](#).

8.2.1.3 Application Curve

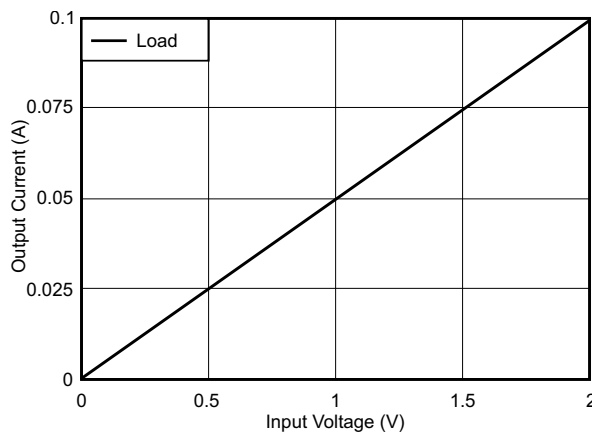


Figure 21. Measured Transfer Function for High-Side V-I Converter

Typical Application (continued)

8.2.1.4 Single Op Amp Bridge Amplifier

Figure 22 shows the basic configuration for a bridge amplifier.

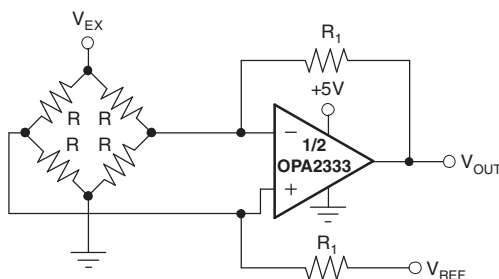
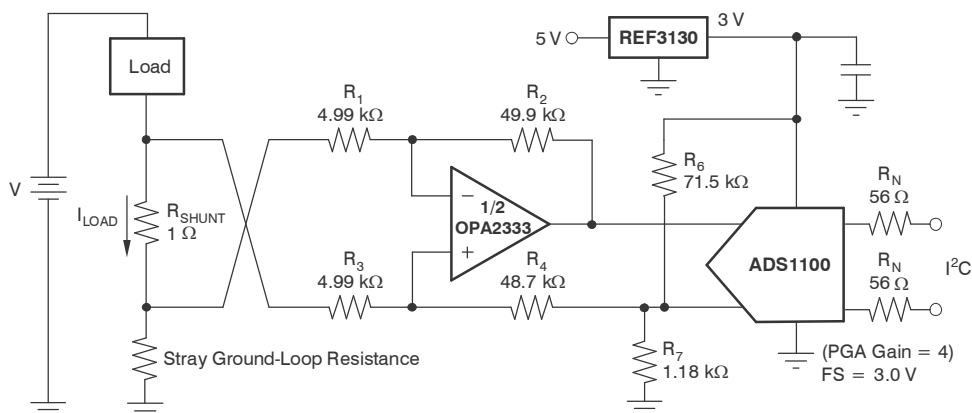


Figure 22. Single Op-Amp Bridge Amplifier

8.2.1.5 Low-Side Current Monitor

A low-side current shunt monitor is shown in Figure 23. R_N are operational resistors used to isolate the ADS1100 from the noise of the digital I²C bus. Because the ADS1100 is a 16-bit converter, a precise reference is essential for maximum accuracy. If absolute accuracy is not required, and the 5-V power supply is sufficiently stable, the REF3130 may be omitted.



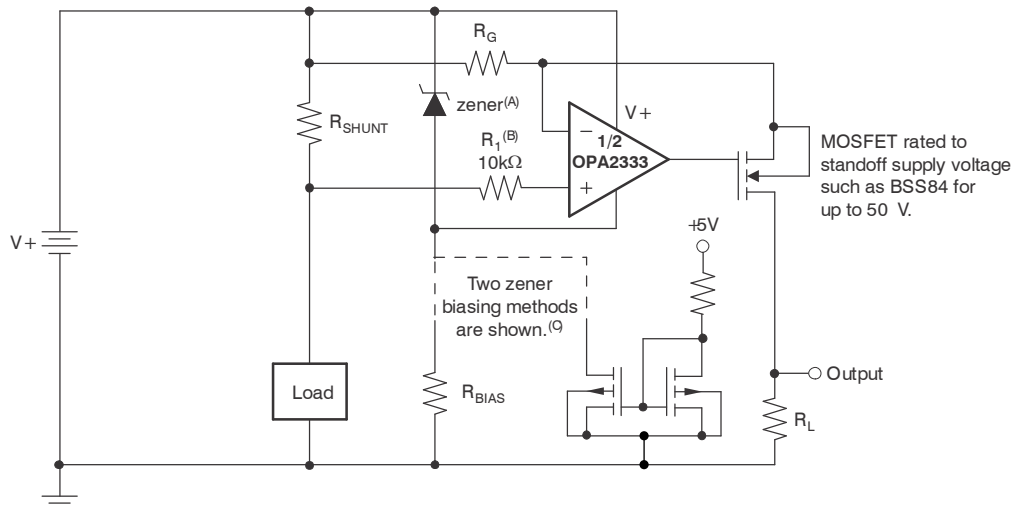
NOTE: 1% resistors provide adequate common-mode rejection at small ground-loop errors.

Figure 23. Low-Side Current Monitor

Typical Application (continued)

8.2.1.6 High-Side Current Monitor

Figure 24 shows the use case for a precision single-supply amplifier for a high-side current sensing circuit.



- A. Zener rated for op amp supply capability (that is, 5.1 V for the OPA2333).
- B. Current-limiting resistor.
- C. Choose a Zener biasing resistor or dual NMOSFETs (FDG6301N, NTJD4001N, or Si1034).

Figure 24. High-Side Current Monitor

8.2.1.7 Precision Instrumentation Amplifier

Figure 25 shows a three op amp implementation for a high-CMRR instrumentation amplifier..

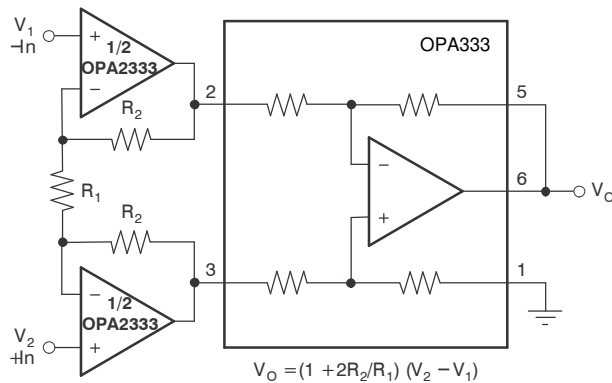


Figure 25. Precision Instrumentation Amplifier

9 Power Supply Recommendations

The OPA2333-Q1 is specified for operation from 1.8 V to 5.5 V (± 0.9 V to ± 2.75 V); many specifications apply from -40°C to $+125^{\circ}\text{C}$. The *Typical Characteristics* section presents parameters that can exhibit significant variance with regard to operating voltage or temperature.

CAUTION

Supply voltages greater than 7 V can permanently damage the device (see the *Absolute Maximum Ratings* table).

Place 0.1- μF bypass capacitors near the power-supply pins to reduce coupling errors from noisy or high-impedance power supplies. For more details on bypass capacitor placement, see the *Layout* section.

10 Layout

10.1 Layout Guidelines

Pay attention to good layout practices. Keep traces short and when possible, use a printed-circuit-board (PCB) ground plane with surface-mount components placed as close to the device pins as possible. Place a 0.1- μF capacitor closely across the supply pins. Apply these guidelines throughout the analog circuit to improve performance and provide benefits, such as reducing the electromagnetic interference (EMI) susceptibility.

Operational amplifiers vary in susceptibility to radio frequency interference (RFI). RFI can generally be identified as a variation in offset voltage or dc signal levels with changes in the interfering RF signal. The OPA2333-Q1 is specifically designed to minimize susceptibility to RFI and demonstrates remarkably low sensitivity compared to previous generation devices. Strong RF fields may still cause varying offset levels.

For lowest offset voltage and precision performance, optimize circuit layout and mechanical conditions. Avoid temperature gradients that create thermoelectric (Seebeck) effects in the thermocouple junctions formed from connecting dissimilar conductors. Cancel these thermally-generated potentials by assuring they are equal on both input terminals. Other layout and design considerations include:

- Use low-thermoelectric-coefficient connections (avoid dissimilar metals).
- Thermally isolate components from power supplies or other heat sources.
- Shield operational amplifier and input circuitry from air currents, such as cooling fans.

Following these guidelines reduces the likelihood of junctions being at different temperatures, which can cause thermoelectric voltages of 0.1 $\mu\text{V}/^{\circ}\text{C}$ or higher, depending on materials used.

10.2 Layout Example

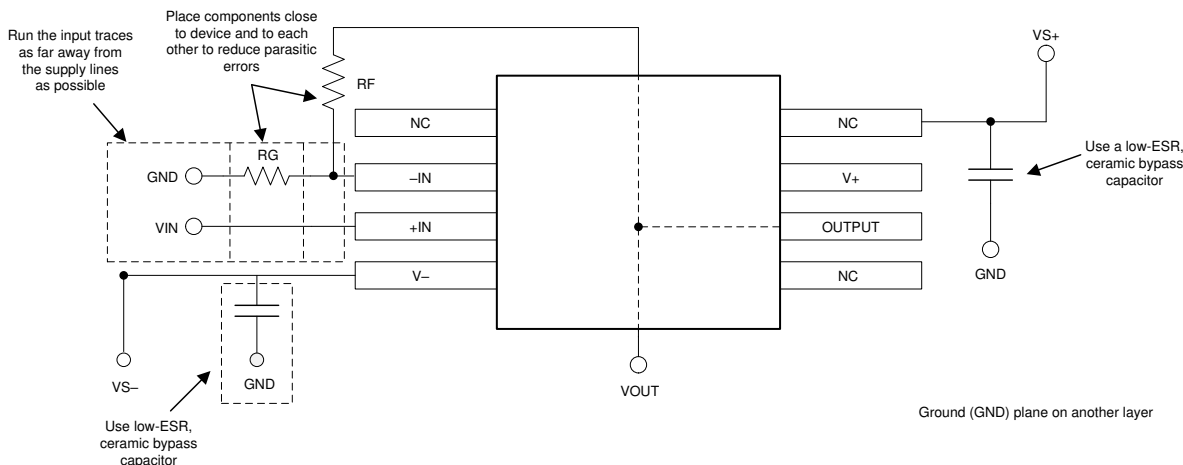


Figure 26. Layout Example

11 デバイスおよびドキュメントのサポート

11.1 ドキュメントのサポート

11.1.1 関連資料

関連資料については、以下を参照してください。

- テキサス・インスツルメンツ、『[ADS1100 Self-Calibrating, 16-Bit ANALOG-TO-DIGITAL CONVERTER](#)』データシート (英語)
- テキサス・インスツルメンツ、『[REF31xx 15ppm/°C Maximum, 100-μA, SOT-23 Series Voltage Reference](#)』データシート (英語)
- テキサス・インスツルメンツ、『[INAx321 microPower, Single-Supply, CMOS Instrumentation Amplifier](#)』データシート (英語)
- テキサス・インスツルメンツ、『[INA32x Precision, Rail-to-Rail I/O INSTRUMENTATION AMPLIFIER](#)』データシート (英語)

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11.3 サポート・リソース

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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11.6 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

12 メカニカル、パッケージ、および注文情報

以降のページには、メカニカル、パッケージ、および注文に関する情報が記載されています。この情報は、そのデバイスについて利用可能な最新のデータです。このデータは予告なく変更されることがあり、ドキュメントが改訂される場合もあります。本データシートのブラウザ版を使用されている場合は、画面左側の説明をご覧ください。

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
OPA2333AQDGKRQ1	ACTIVE	VSSOP	DGK	8	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR		OCOQ	Samples
OPA2333AQDRQ1	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	02333Q	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBsolete: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
OPA2333AQDGKRQ1	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.3	1.3	8.0	12.0	Q1
OPA2333AQDRQ1	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
OPA2333AQDGKRQ1	VSSOP	DGK	8	2500	367.0	367.0	38.0
OPA2333AQDRQ1	SOIC	D	8	2500	356.0	356.0	35.0

DGK0008A



PACKAGE OUTLINE

VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



NOTES:

PowerPAD is a trademark of Texas Instruments.

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-187.

EXAMPLE BOARD LAYOUT

DGK0008A

™ VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 15X



4214862/A 04/2023

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
8. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
9. Size of metal pad may vary due to creepage requirement.

EXAMPLE STENCIL DESIGN

DGK0008A

TM VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE
SCALE: 15X

4214862/A 04/2023

NOTES: (continued)

11. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
12. Board assembly site may have different recommendations for stencil design.



D0008A

PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4214825/C 02/2019

NOTES:

1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed $.006$ [0.15] per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MS-012, variation AA.

EXAMPLE BOARD LAYOUT

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON .005 INCH [0.125 MM] THICK STENCIL
SCALE:8X

4214825/C 02/2019

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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