



# **NexFET™ Smart Synchronous Rectifier**

#### **FEATURES**

- Typical R<sub>on</sub> of 0.55 mΩ at 4.5 V<sub>DD</sub>
- Integrated FET Driver
- Max Rated Current 80A
- High Density SON 5-mm × 6-mm Footprint
- Ultra Low Inductance Package
- System Optimized PCB Footprint
- TTL IN signal Compatible
- Halogen Free
- RoHS Compliant Lead Free Terminal Plating Halogen Free

#### **APPLICATIONS**

 Secondary Synchronous Rectification for DC/DC Converters

## **DESCRIPTION**

The CSD43301Q5M NexFET<sup>TM</sup> Smart Synchronous Rectifier is a highly optimized design for secondary synchronous rectification in a high power high density DC/DC converter. This product integrates the driver IC and an ultra low  $R_{\text{on}}$  Power MOSFET to complete the synchronous rectification function. In addition, the PCB footprint has been optimized to help reduce design time and simplify the completion of the overall system design.

#### **ORDERING INFORMATION**

Device	Package	Media	Qty	Ship
CSD43301Q5M	SON 5-mm × 6-mm Plastic Package	13-Inch Reel	2500	Tape and Reel

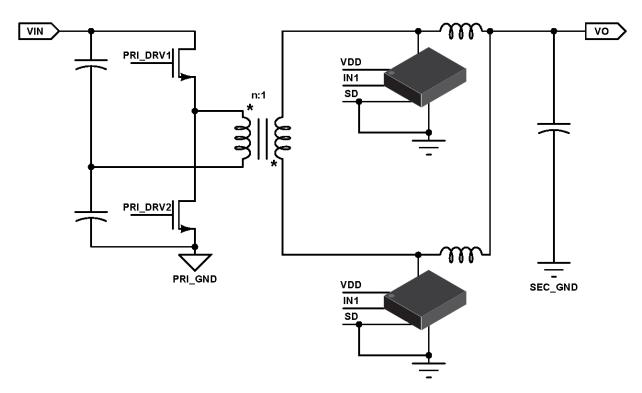


Figure 1. Application Diagram

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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

# **ABSOLUTE MAXIMUM RATINGS**(1)

 $T_A = 25$ °C (unless otherwise noted)

		VAI	UNIT			
		MIN	MAX			
DRAIN to P <sub>GN</sub>	ID	-0.3	12	V		
DRAIN to P <sub>GN</sub>	<sub>ID</sub> (10ns)	-7	14	V		
V <sub>DD</sub> to P <sub>GND</sub>		-0.3	8	V		
IN, SD to P <sub>GN</sub>	D <sup>(2)</sup>	-0.3	V <sub>DD</sub> + 0.3	V		
CCD Datin	Human Body Model (HBM)		2000	V		
ESD Rating	Charged Device Model (CDM)		500	V		
Power Dissipation (P <sub>D</sub> )			12	W		
Operating Temperature Range, (T <sub>J</sub> )		-40	150	°C		
Storage Temp	perature Range, (T <sub>STG</sub> )	-65	150	°C		

<sup>(1)</sup> Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under "Recommended Operating Conditions" is not implied. Exposure to Absolute Maximum rated conditions for extended periods may affect device reliability.

#### RECOMMENDED OPERATING CONDITIONS

 $T_A = 25^{\circ}$  (unless otherwise noted)

Parameter	Conditions	MIN	MAX	UNIT
Bias Voltage (V <sub>DD</sub> )		4.5	6	V
Input Supply Voltage (V <sub>IN</sub> )			9.6	V
Continuous Output Current (I <sub>OUT</sub> )			80	Α
Peak Output Current, (I <sub>OUT-PK</sub> ) <sup>(1)</sup>			120	Α
Switching Frequency, (f <sub>SW</sub> )			1500	kHz
Minimum IN Pulse Width		48		ns
Operating Temperature		-40	125	°C

<sup>(1)</sup> Peak Output Current is applied for  $t_p = 50\mu s$ .

#### THERMAL INFORMATION

 $T_A = 25$ °C (unless otherwise noted)

	PARAMETER	MIN	TYP	MAX	UNIT
$R_{\theta JC}$	Thermal Resistance, Junction-to-Case (Top of package)			20	°C/W
$R_{\theta JB}$	Thermal Resistance, Junction-to-Board <sup>(1)</sup>			2	°C/W

(1)  $R_{\theta JB}$  value based on hottest board temperature within 1mm of the package.

<sup>(2)</sup> Must not exceed 8V



# **ELECTRICAL CHARACTERISTICS**

 $T_A = 25$ °C,  $V_{DD} = 5V$  (unless otherwise noted)

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
Device On Resistance					
D	I <sub>D</sub> = 50A, , T <sub>J</sub> = 25°C		0.55	0.70	mΩ
R <sub>on</sub>	I <sub>D</sub> = 50A, T <sub>J</sub> = 125°C		0.70	0.88	mΩ
$V_{DD}$		,			
Standby Supply Current ( I <sub>DD</sub> )	$SD = V_{DD} = 5V$		153	300	μΑ
Operating Supply Current (I <sub>DD</sub> )	SD = 0V, IN = 50% Duty Cycle, f <sub>SW</sub> = 300kHz		29.5		mA
POWER-ON RESET AND UNDER VOLTA	GE LOCKOUT				
Davids on David ()/ Divine)	T <sub>A</sub> = 25°C	3.9	4.2	4.5	V
Power on Reset (V <sub>DD</sub> Rising)	$T_A = -40$ °C to 140°C	3.7	4.2	4.65	V
UVLO (V <sub>DD</sub> Falling)		3.45	3.9	4.35	V
Hysteresis		200	300	500	mV
IN				·	
IN Logic Level High (V <sub>INH</sub> )		2.0			V
IN Logic Level Low (V <sub>INL</sub> )				0.8	V
IN Input Hysteresis			0.8		V
IN to DRAIN Propagation Delay (t <sub>PDLH</sub> )			32		ns
IN to DRAIN Propagation Delay (t <sub>PDHL</sub> )	$V_{DD} = 5V$ , SD = 0, $I_D = 25A$ (See Figure 4)		80		ns
Minimum Pulse Width Changes Output			36	48	ns
SD					
SD Logic Level High Threshold (V <sub>IH</sub> )		2.0			V
SD Logic Level Low Threshold (V <sub>IL</sub> )				0.8	V
Hysteresis			0.8		V
SD to DRAIN Propagation Delay (t <sub>PDLH</sub> )	V 5V IV V 1 054 (0 5 5)		80		ns
SD to DRAIN Propagation Delay (t <sub>PDHL</sub> )	$V_{DD} = 5V$ , $IN = V_{DD}$ , $I_D = 25A$ (See Figure 5)		32		ns
Dynamic Characteristics					
Output Capacitance (C <sub>O</sub> )	V 6V		10	13	nF
Output Charge (Q <sub>O</sub> )	$V_{DRAIN} = 6V$		54		nC
Body Diode		·			
Forward Voltage (V <sub>F</sub> )	I <sub>D</sub> = 40A		0.75	0.85	V
Reverse Recovery Charge (Q <sub>RR</sub> )	1 404 1/ 01/1/1/ 4504/		161		nC
Reverse Recovery Time Delay (t <sub>RR</sub> )	$I_D = 40A$ , $V_{DRAIN} = 6V$ , $di/dt = 150A/\mu s$		72		ns



# **PIN CONFIGURATION**

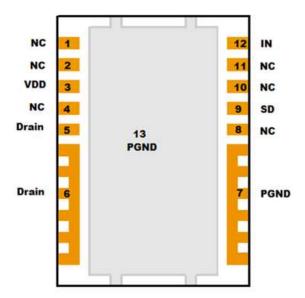


Figure 2. Pin Configuration

## **PIN DESCRIPTION**

Р	IN	DECCRIPTION					
NO.	NAME	DESCRIPTION					
1,2,4, 8, NC No connect. These should not be used for any electrical connection. These pins should not be connected other. Connect to dead copper only.							
3	$V_{DD}$	Supply Voltage for IC					
5,6 DRAIN Drain terminal of internal MOSFET		Drain terminal of internal MOSFET					
7 P <sub>GND</sub>		Power Ground and source terminal of the internal MOSFET. Needs to be connected to Pin 13 on PCB					
9 SD		Shut Down Pin: Logic High disables the Device					
12 IN		Input for Gate Driver					
13 P <sub>GND</sub>		Power Ground and source terminal of the internal MOSFET. Needs to be connected to Pin 7 on PCB					

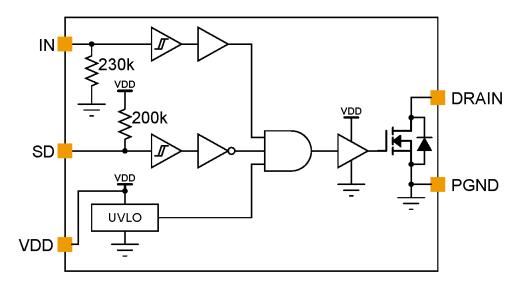


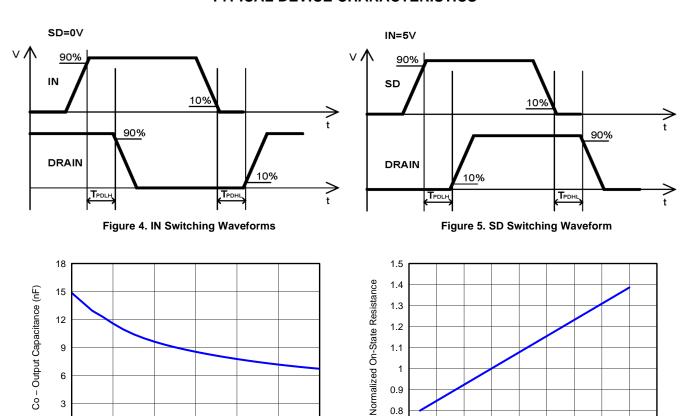
Figure 3. Functional Block Diagram



3

0

## TYPICAL DEVICE CHARACTERISTICS



12

G000

0.9

8.0

0.7

-50

Drain Voltage (V) Figure 6. Output Capacitance

6

Temperature (°C) Figure 7. Normalized On Resistance Ron

75

100

125

175

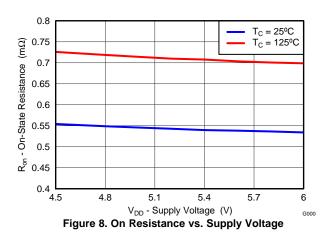
G000

50

25



#### TYPICAL DEVICE CHARACTERISTICS CONTINUED



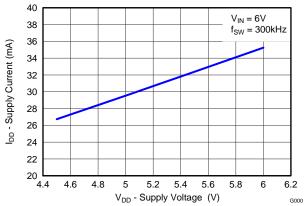


Figure 9. Supply Current vs. Supply Voltage

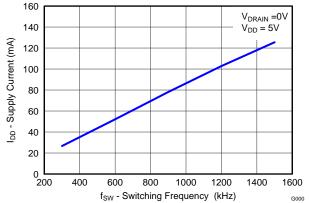


Figure 10. Supply Current vs. Switching Frequency



#### Application Information

#### **V<sub>DD</sub>** and Under-Voltage Lockout (UVLO)

The driver IC in the CSD43301Q5M has an internal UVLO protection feature on the  $V_{DD}$  pin. Whenever the driver is in the UVLO condition (i.e. when  $V_{DD}$  voltage is less than  $V_{ON}$  during power up and when  $V_{DD}$  voltage is less than  $V_{OFF}$  during power down), this circuit holds the gate of the integrated MOSFET LOW, regardless of the status of IN and SD. The UVLO is typically 4.2V with 300-mV typical hysteresis. This hysteresis helps prevent chatter when low  $V_{DD}$  supply voltages have noise from the power supply and also when there are droops in the  $V_{DD}$  bias voltage when the system commences switching and there is a sudden increase in  $V_{DD}$ . This provides the capability to operate at low voltage levels (below 5V), along with best-in-class switching characteristics. For example, at power up, the MOSFET remains OFF until the  $V_{DD}$  voltages reaches the UVLO threshold. This prevents operating the MOSFET in the linear region and conducting a large load current at the same time, which often results in device overheating and can potentially damage the device.

Since the driver draws current from the  $V_{DD}$  pin to bias all internal circuits, for the best high-speed circuit performance, Multi-Layer Ceramic Capacitor (MLCC) bypass capacitors are recommended to prevent noise problems. A 1  $\mu$ F MLCC type capacitor should be located as close as possible to the  $V_{DD}$  to GND pins of the gate driver.

#### **Operating Supply Current**

The driver IC in the CSD43301Q5M has a low quiescent current in normal operation.  $I_{DDQ}$  is less than 0.2 mA when the device is disabled (SD = 0). The operating current vs. supply voltage is shown in Figure 9, and the operating current vs. frequency is shown in Figure 10.

#### **Input Stage**

The input pins (IN and SD) of the CSD43301Q5M are based on a TTL/CMOS compatible input threshold logic that is independent of the  $V_{DD}$  supply voltage. With a typical high threshold of 2.2 V and a typical low threshold of 1.2 V, the logic level thresholds can be conveniently driven with PWM control signals derived from 3.3-V or 5-V digital power controllers. Wider hysteresis (typical of 0.8 V) offers enhanced noise immunity compared to traditional TTL logic implementations, where the hysteresis is typically less than 0.5 V. These devices also feature tight control of the input pin threshold voltage levels which eases system design considerations and ensures stable operation across temperature. The very low input capacitance on these pins reduces loading and increases switching speed. The device features an important safety function wherein, whenever any of the input pins are in a floating condition, the output of the respective channel is held in the low state. This is achieved using a  $V_{DD}$  pull-up resistor on the SD input or a GND pull-down resistor on the IN input. This can be seen in the block diagram in Figure 3.

#### **Power Dissipation**

Power Dissipation of the CSD43301Q5M used in secondary rectification is given by the following:

$$P_{LOSS} = P_{DRV} + P_{COND} + P_{SW} \tag{1}$$

where driver loss is given by

$$P_{DRV} = V_{DD} \times I_{DD} \tag{2}$$

and conduction loss is given by

$$P_{COND} = I_{D RMS}^2 \times R_{ON} \tag{3}$$

Switching losses consist of body diode conduction losses during dead time, body diode reverse recovery losses, and output charge losses, given by the following:

$$P_{SW} = I_D \times V_F \times (DT_R + DT_F) \times F_{SW} + Q_{RR} \times V_{DRAIN} \times F_{SW} + \frac{1}{2} Q_{OSS} \times V_{DRAIN} \times F_{SW}$$
(4)



#### **Recommended PCB Design Overview**

The CSD43301Q5M features extremely low nominal R<sub>ON</sub>. In order to maximize the performance of this device, some simple layout guidelines should be followed.

- The DRAIN pins of the CSD43301Q5M should be placed as close as possible to the inductor and connected with a short wide trace. This reduces PCB conduction losses and reduce switching noise level. (1)
- The GND pin (pin 7) must be connected into the thermal pad (pin 13) on the bottom of the device via a copper pour (without thermal spokes) for maximum performance.
- The CSD43301Q5M has the ability to use the GND planes as the primary thermal path. As such, the use of thermal vias is an effective way to pull away heat from the device and into the system board. Concerns of solder voids and manufacturability problems can be addressed by the use of three basic tactics to minimize the amount of solder attach that will wick down via the barrel:
  - Intentionally space out the vias from each other to avoid a cluster of holes in a given area.
  - Use the smallest drill size allowed in your design. The example in Figure 11 uses vias with a 10 mil drill hole and a 16 mil capture pad.
  - Tent the opposite side of the via with a solder mask.

In the end, the number and size of the thermal vias should align with the end user's PCB design rules and manufacturing types.

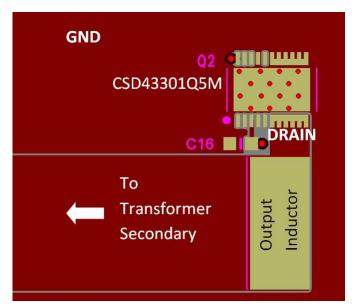
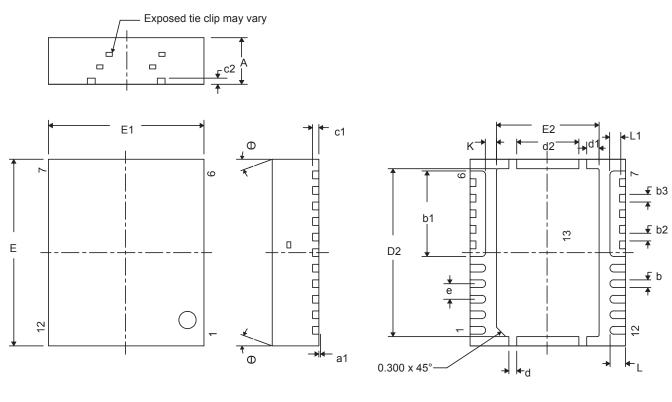


Figure 11. Recommended PCB Layout (Top Down View)

(1) Keong W. Kam, David Pommerenke, "EMI Analysis Methods for Synchronous Buck Converter EMI Root Cause Analysis", University of Missouri – Rolla



## **MECHANICAL DATA**

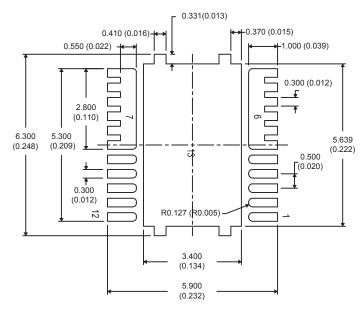


TOD \ ((E)A)	0.10 = 1.41 = 1.44	DOTTOMAN (IEVA)
TOP VIEW	SIDE VIEW	BOTTOM VIEW

5114		MILLIMETERS		INCHES			
DIM	Min	Nom	Max	Min	Nom	Max	
Α	1.400	1.450	1.500	0.055	0.057	0.059	
A1	0.000	0.000	0.050	0.000	0.000	0.002	
b	0.200	0.250	0.350	0.008	0.010	0.013	
b1		2.750 TYP			0.108 TYP		
b2	0.200	0.250	0.320	0.008	0.010	0.013	
b3		0.250 TYP			0.010 TYP		
c1	0.150	0.200	0.200 0.250 0.006		0.008	0.010	
c2	0.150	0.200	0.250	0.006	0.008	0.010	
D2	5.300	5.400	5.500	0.209	0.213	0.217	
d	0.200	0.250	0.300	0.008	0.010	0.012	
d1	0.350	0.400	0.450	0.014	0.016	0.018	
d2	1.900	2.000	2.100	0.075	0.079	0.083	
Е	5.900	6.000	6.100	0.232	0.236	0.240	
E1	4.900	5.000	5.100	0.193	0.197	0.201	
E2	3.200	3.300	3.400	0.126	0.130	0.134	
е		0.500 TYP			0.020 TYP		
K		0.350 TYP			0.014 TYP		
L	0.400	0.500	0.600	0.016	0.020	0.024	
L1	0.210	0.310	0.410	0.008	0.012	0.016	
θ	0.00			0.00	_	_	

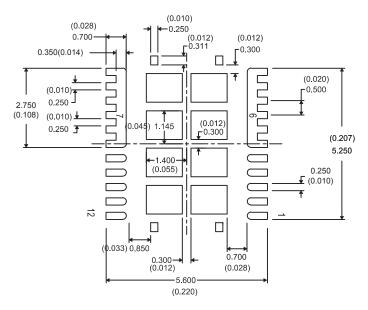


#### **Recommended PCB Pattern**



NOTE: Dimensions are in mm (inches).

## **Recommended Stencil**



NOTE: Dimensions are in mm (inches).



# **REVISION HISTORY**

CI	Changes from Original (December 2012) to Revision A						
•	Changed Figure 3						
CI	hanges from Revision A (December 2012) to Revision B	Page					
•	Changed the MECHANICAL DATA image and corresponding table	9					
•	Changed the Recommended PCB Pattern - lead width From: 0.300(0.012) To: 0.350(0.014)	10					
•	Changed the Recommended Stencil image	10					

11-Nov-2025

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#### PACKAGING INFORMATION

Orderable part number	Status	Material type	Package   Pins	Package qty   Carrier	<b>RoHS</b> (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
CSD43301Q5M	Active	Production	LSON-CLIP (DQP)   12	2500   LARGE T&R	ROHS Exempt	NIPDAU	Level-2-260C-1 YEAR	-55 to 125	43301M
CSD43301Q5M.B	Active	Production	LSON-CLIP (DQP)   12	2500   LARGE T&R	-	Call TI	Call TI	-55 to 125	

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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<sup>(5)</sup> MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

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Last updated 10/2025