

High-Voltage System Power Management Device with PMBus™

Check for Samples: [LM5056](http://www.ti.com/product/lm5056#samples), [LM5056A](http://www.ti.com/product/lm5056a#samples)

- **²• Input Voltage Range: 10 V to 80 V • Server Backplane Systems**
- **and VAUX with 12-bit resolution, 1-kHz • Industrial Telemetry Applications sampling rate**
- **• True input Power Measurement using DESCRIPTION**
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- -
	-
- **Current Measurement Accuracy** interface.
	- **– LM5056A: ±1.25%**
	-
-
-
- **with SMBA Notification**
- **• Black-Box Capture of Telemetry** APPLICATION DIAGRAM **Measurements and Device Status Triggered by WARN or FAULT Condition**
- **• I ²C/SMBus Interface and PMBus Compliant Command Structure**

¹FEATURES APPLICATIONS

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- **• Real-Time Monitoring of VIN, IIN, PIN, VOUT, • Base Station Power Distribution Systems**
	-

simultaneous sampling of Vin and Iin The LM5056/LM5056A combines high-performance • **Remote Temperature Sensing with** analog and digital technology with a PMBus™ **programmable warning thresholds the compliant SMBus™ and I²C interface to accurately programmable warning thresholds** measure the electrical operating conditions of **• Power Measurement Accuracy** systems connected to ^a backplane power bus. The **– LM5056A: ±1.75%** LM5056/LM5056A continuously supplies real-time **– LM5056: ±2.25%** power, voltage, current, temperature and fault data to the system management host via the SMBus

The LM5056/LM5056A monitoring block computes **– LM5056: ±1.5%** both the real time and average values of subsystem **• Voltage Measurement Accuracy: ±1.0%** operating parameters (VIN, IIN, PIN, VOUT) as well **• Averaging of VIN, IIN, PIN, and VOUT with** as the peak power. Accurate power averaging is **Programmable Interval Ranging from 0.001 s** accomplished by averaging the product of the input voltage and current. A black-box (telemetry and fault **to 4 s** snapshot) function captures and stores telemetry data **• Programmable WARN and FAULT Thresholds** and device status in the event of a warning or a fault.

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TYPICAL APPLICATION DIAGRAM r_s
W LOAD $+48V$ o-V_{IN} V_{OUT} Ŧ VIN VIN_K SENSE **OUT** VAU) DIODE NC ADR2 NC ADR1 MMBT3904 **LM5056/LM5056A** VREF NC ADR0 CVREF **SMBA** SMBus SDAI Interface CL SDAO VDD \bullet 5.0V **SCL** DGND AGND CVDD ╕ $\overline{\dagger}$ ╧

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating ratings indicate conditions for which the device is intended to be functional, but do not ensure specific performance limits. For ensured specifications and conditions see Electrical Characteristics Table.

(2) The human body model is a 100-pF capacitor discharged through a 1.5-kΩ resistor into each pin.

RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating ratings indicate conditions for which the device is intended to be functional, but do not ensure specific performance limits. For ensured specifications and conditions see the Electrical Characteristics Table.

THERMAL INFORMATION

(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, [SPRA953](http://www.ti.com/lit/pdf/spra953).

(2) The junction-to-ambient thermal resistance under natural convection is obtained in a simulation on a JEDEC-standard, high-K board, as specified in JESD51-7, in an environment described in JESD51-2a.

(3) The junction-to-case (top) thermal resistance is obtained by simulating a cold plate test on the package top. No specific JEDECstandard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

(4) The junction-to-board thermal resistance is obtained by simulating in an environment with a ring cold plate fixture to control the PCB temperature, as described in JESD51-8.

(5) The junction-to-top characterization parameter, ψ_{JT} , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining θ_{JA} , using a procedure described in JESD51-2a (sections 6 and 7).

(6) The junction-to-board characterization parameter, ψ_{JB} , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining θ_{JA} , using a procedure described in JESD51-2a (sections 6 and 7).

(7) The junction-to-case (bottom) thermal resistance is obtained by simulating a cold plate test on the exposed (power) pad. No specific JEDEC standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

EXAS STRUMENTS

ELECTRICAL CHARACTERISTICS

Limits in standard type are for T_J = 25°C only; limits in boldface type apply over the junction temperature (T_J) range of -40°C to 125°C unless otherwise stated. Minimum and maximum limits are ensured through test, design, or statistical correlation. Typical values represent the most likely parametric norm at $T_J = 25^\circ \text{C}$, and are provided for reference purposes only. Unless otherwise stated the following conditions apply: VIN = 48 V and VDD = 5.0 V. See $^{(1)}$ and $^{(2)}$.

(1) Current out of a pin is indicated as a negative value.
(2) All limits are ensured. All electrical characteristics ha All limits are ensured. All electrical characteristics having room temperature limits are tested during production at $T_A = 25$ °C. All hot and cold limits are ensured by correlating the electrical characteristics to process and temperature variations and applying statistical process control.

ELECTRICAL CHARACTERISTICS (continued)

Limits in standard type are for T_J = 25°C only; limits in boldface type apply over the junction temperature (T_J) range of -40°C to 125°C unless otherwise stated. Minimum and maximum limits are ensured through test, design, or statistical correlation. Typical values represent the most likely parametric norm at $T_J = 25^\circ \text{C}$, and are provided for reference purposes only. Unless otherwise stated the following conditions apply: $VIN = 48$ V and $VDD = 5.0$ V. See ^{[\(1\)](#page-4-0)} and^{[\(2\)](#page-4-0)}.

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VDD = 5.5V

TYPICAL CHARACTERISTICS

Unless otherwise specified the following conditions apply: T_J = 25°C, VIN, VIN_K, SENSE, and OUT = 48 V and VDD = 5.0 V.

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TYPICAL CHARACTERISTICS (continued)

Unless otherwise specified the following conditions apply: $T_J = 25^{\circ}$ C, VIN, VIN_K, SENSE, and OUT = 48 V and VDD = 5.0 V. All graphs show junction temperature.

Junction Temperature (°C) **Figure 11. VOUT Measurement Error Figure 12. VAUX Measurment Error (VAUX = 2.80 V)**

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DEVICE INFORMATION

TERMINAL FUNCTIONS

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TERMINAL FUNCTIONS (continued)

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FUNCTIONAL DESCRIPTION

The LM5056/LM5056A provides intelligent monitoring of the input voltage, output voltage, input current, input power, temperature, and an auxiliary input. The LM5056/LM5056A also provides a peak capture of the input power and programmable hardware averaging of the input voltage, current, power, temperature, and output voltage. Warning thresholds which trigger the SMBA pin may be programmed for input voltage, current, power, output voltage, and temperature via the PMBus interface.

Powering The LM5056/LM5056A

The LM5056/LM5056A is enabled by increasing the input voltage on VIN above the POR_{FN} threshold voltage, typically 8.7 V. There exists a VDD power on reset (VDD_{POR}) threshold on VDD of 3.8 V. The VDD_{POR} threshold must be surpassed to ensure proper telemetry readings. VDD must be powered externally by a 5 V power supply with an allowable tolerance of ±10%. The SMBus address of the LM5056/LM5056A is captured based on the states of the ADR0, ADR1, and ADR2 pins (GND, NC, VDD) during turn on and is latched into a volatile register once VDD has exceeded its POR threshold of 3.8 V. Reassigning or postponing the address capture is accomplished by holding the VREF pin to AGND. Pulling the VREF pin low also resets the logic and erases the volatile memory of the LM5056/LM5056A. Once released, the VREF pin charges up to its final value and the address is latched into a volatile register when the voltage at the VREF exceeds 2.55 V.

VDD

As mentioned in the previous paragraph, the LM5056/LM5056A VDD pin must be externally powered by a 5.0 V, ±10% supply. The required current is typically 6.1 mA. The pull-up voltage for the CL, ADR2, ADR1 and ADR0 pins should be the same as the voltage applied to VDD if they are to be tied high. It may also be used as the pull-up supply for the SMBus signals (SDAI/O, SCL, SMBA). It is recommended to connect a ceramic bypass capacitance having a value of 1 μ F or greater as close to the VDD pin as the PCB layout allows.

Remote Temperature Sensing

The LM5056/LM5056A is designed to measure temperature remotely using an MMBT3904 NPN transistor. The base and collector of the MMBT3904 should be connected to the DIODE pin and the emitter to the LM5056/LM5056A AGND. Place the MMBT3904 near the device that requires temperature sensing. The temperature is measured by means of a change in the diode voltage in response to a step in current supplied by the DIODE pin. The DIODE pin sources a constant 9.4 µA but pulses 250 µA once every millisecond in order to measure the diode temperature. Care must be taken in the PCB layout to keep the parasitic resistance between the DIODE pin and the MMBT3904 low as to not degrade the measurement. Additionally, a small 1 nF bypass capacitor should be placed in parallel with the MMBT3904 to reduce the effects of noise. The temperature can be read using the READ_TEMPERATURE_1 PMBus command (8Dh). The default limits of the LM5056/LM5056A causes SMBA pin to be pulled low if the measured temperature exceeds 150°C. These thresholds can be reprogrammed via the PMBus interface using the OT_WARN_LIMIT (51h) and OT_FAULT_LIMIT (4Fh) commands. If the temperature measurement and protection capability of the LM5056/LM5056A are not used, the DIODE pin should be connected to the ground plane.

Erroneous temperature measurements may result when the device input voltage is below the minimum operating voltage (10 V), due to VREF dropping out below the nominal voltage (2.97 V).

APPLICATION INFORMATION

DESIGN-IN PROCEDURE

Refer to [Figure](#page-11-0) 13 for the Typical Application Circuit diagram. The following is the step-by-step procedure for hardware design of the LM5056/LM5056A. This procedure refers to section numbers that provide detailed information on the following design steps. The recommended design-in procedure is as follows:

Figure 13. Typical Application Circuit

CURRENT Range (Selecting RS)

The LM5056/LM5056A monitors the input current by measuring the voltage across the sense resistor (R_S) , connected from VIN_K to SENSE. The required resistor value is calculated from:

$$
R_{\rm S} = \frac{V_{\rm S}}{I_{\rm FS}}
$$

where

 I_{FS} is the expected full scale current and V_S is the current sense voltage range based on the current select range setting (CL). (1)

If the voltage across R_s reaches V_s , the current measurement reaches the full-scale measurement. As mentioned before, it is important to limit the current to the full-scale reading. While there is internal circuitry intended to maintain the integrity of the other readings in the telemetry, the ADC and MUX are shared so overranging an input may compromise the integrity of the other readings.

V_S can be set to either 27.0 mV or 54.4 mV through software commands or the CL pin. This setting defaults to the sense voltage set at the CL pin during start-up. The value can be set via the PMBus with the DEVICE_SETUP (D9h) command, which defaults to the 27.0 mV setting. Once the full scale current, I_{FS} is known and the V_S range is chosen, the sense resistor can be calculated. The maximum load current in normal operation can be used to determine the required power rating for the sense resistor $R_{\rm S}$.

Connections from R_S to the LM5056/LM5056A should be made using Kelvin techniques. In the suggested layout of [Figure](#page-12-0) 14 the small pads at the lower corners of the sense resistor connect only to the sense resistor terminals, and not to the traces carrying the high current. With this technique, only the voltage across the sense resistor is applied to VIN_K and SENSE, eliminating the voltage drop across the high-current solder connections.

Figure 14. Sense Resistor Connections, Edge Sensed

If the PCB layout and resistor pads allow for it, the connection shown in [Figure](#page-12-1) 15 gives optimal kelvin sensing performance.

Figure 15. Sense Resistor Connections, Centered Sensed

 C_{VIN} , C_D , C_{VREF} , C_S , and C_{VDD}

Using ceramic bypass capacitors can improve performance in noise heavy environments. Not every pin of the LM5056/LM5056A is the same when it comes to placing bypass capacitors.

- **CVIN**: This capacitor is not required but can improve VIN telemetry performance in noisy situations. Typical values for the VIN bypass capacitor can range from 1 nF to 100 nF to effectivly reduce input noise. The voltage on C_{VIN} is high, so a 100 V or higher voltage capacitor will work.
- **C_D**:The C_D capacitor is recommended if the diode is placed far from the LM5056/LM5056A DIODE pin. Too large of a capacitance will corrupt the voltage waveform across the diode used to measure the absolute temperature. A typical value of capacitance for C_D is 1 nF. The voltage on C_D is low, so a 6.3 V or higher voltage capacitor will work.
- **CVREF**: CVREF is required since it is placed on the output of the internal votlage reference. This capacitor should be a 1 μ F ceramic. The voltage on C_{VRFF} is low, so a 6.3 V or higher voltage capacitor will work.
- **CS**: The current sense amplifier is designed to amplifiy small voltages. Using a bypass capacitor across the current sense amplifier input pins (VIN_K and SENSE) will facilitate accurate current telemetry. Functional values of C_S can range from 10 nF to 1 µF. The voltage on C_S is low, so a 6.3 V or higher voltage capacitor will work.
- C_{VDD} : C_{VDD} is required because it provides bypassing from the 5.0 V rail for the internal digital circuitry. This capacitor should be a 1-µF ceramic. The voltage on C_{VDD} is low, so a 6.3 V or higher voltage capacitor will work.

PC Board Guidelines

The following guidelines should be followed when designing the PC board for the LM5056/LM5056A:

- Place a 1-μF ceramic capacitor as close as possible to VREF pin and AGND.
- Place a 1-μF ceramic capacitor as close as possible to VDD pin and AGND.
- Minimize the inductance between the VIN and VIN K pins. There are anti-parallel diodes between these pins so any voltage greater than 0.3 V in either polarity causes significant current flow through the diodes, which can result in device failure. Do not place any resistors between these two nodes.
- Minimize the voltage between the VIN_K and SENSE pins. There are anti-parallel diodes between these pins so any voltage greater than 3.0 V in either polarity causes significant current flow through the diodes. Internal series resistors limit the current in these pins and provide a limited level of protection in the event of a voltage transient.
- The sense resistor (R_S) should be placed close to the LM5056/LM5056A. Connect R_S using the Kelvin techniques shown in [Figure](#page-12-0) 14 or [Figure](#page-12-1) 15.
- The high-current path from the board's input to the load and the return path, should be parallel and close to each other to minimize loop inductance.
- The AGND and DGND connections should be connected at the pins of the device. The return connections for the various components around the LM5056/LM5056A should be connected directly to each other, and to the LM5056/LM5056A's DGND and AGND pin connections, and then connected to the system ground at one point. Do not connect the various component return pads to each other through the high current ground line.

Applications Circuit

Power, voltage, current, fault, temperature, and LED luminosity telemetry for LED street lamps.

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Applications Circuit

Telemetry of input rail and 5-V bias rail output current and voltage.

PMBus™ Command Support

The device features an SMBus interface that allows the use of PMBus commands to set warn levels, error masks, and get telemetry on VIN, VOUT, IIN, VAUX, and PIN. The supported PMBus commands are shown in Table 1.

Table 1. PMBus™ Command Support

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Table 1. PMBus™ Command Support (continued)

Standard PMBus Commands

CLEAR_FAULTS (03h)

The CLEAR_FAULTS command is a standard PMBus command that resets all stored warning and fault flags and the SMBA signal. If a fault or warning condition still exists when the CLEAR_FAULTS command is issued, the SMBA signal may not clear or will reassert almost immediately. This command uses the PMBus send byte protocol.

CAPABILITY (19h)

The CAPABILITY command is a standard PMBus command that returns information about the PMBus functions supported by the LM5056/LM5056A/LM5056/LM5056AA. This command is read with the PMBus read byte protocol.

VOUT_UV_WARN_LIMIT (58h)

The VOUT_UV_WARN_LIMIT command is a standard PMBus command that allows configuring or reading the threshold for the VOUT under-voltage warning detection. Reading and writing to this register should use the coefficients shown in the [Table](#page-35-0) 39. Accesses to this command should use the PMBus read or write word protocol. If the measured value of VOUT falls below the value in this register, VOUT under-voltage warn flags are set and the **SMBA** signal is asserted.

Table 3. VOUT_UV_WARN_LIMIT Register

OT_FAULT_LIMIT (4Fh)

The OT_FAULT_LIMIT command is a standard PMBus command that allows configuring or reading the threshold for the over-temperature fault detection. Reading and writing to this register should use the coefficients shown in the [Table](#page-35-0) 39. Accesses to this command should use the PMBus read or write word protocol. If the measured temperature exceeds this value, an over-temperature fault is triggered and the SMBA signal is asserted. After the measured temperature falls below the value in this register, the CLEAR_FAULTS command (03h) should be sent to de-assert the SMBA signal. A single temperature measurement is an average of 16 round-robin cycles; therefore, the minimum temperature fault detection time is 16 ms.

Table 4. OT_FAULT_LIMIT Register

OT_WARN_LIMIT (51h)

The OT_WARN_LIMIT command is a standard PMBus command that allows configuring or reading the threshold for the over-temperature warning detection. Reading and writing to this register should use the coefficients shown in the [Table](#page-35-0) 39. Accesses to this command should use the PMBus read or write word protocol. If the measured temperature exceeds this value, an over-temperature warning is triggered and the over-temperature warn flags set in the respective registers and the SMBA signal asserted. A single temperature measurement is an average of 16 round-robin cycles; therefore, the minimum temperature warn detection time is 16 ms.

Table 5. OT_WARN_LIMIT Register

VIN_OV_WARN_LIMIT (57h)

The VIN_OV_WARN_LIMIT command is a standard PMBus command that allows configuring or reading the threshold for the VIN over-voltage warning detection. Reading and writing to this register should use the coefficients shown in the [Table](#page-35-0) 39 Table. Accesses to this command should use the PMBus read or write word protocol. If the measured value of VIN rises above the value in this register, VIN over-voltage warn flags are set in the respective registers and the SMBA signal is asserted.

Table 6. VIN_OV_WARN_LIMIT Register

VIN_UV_WARN_LIMIT (58h)

The VIN_UV_WARN_LIMIT command is a standard PMBus command that allows configuring or reading the threshold for the VIN under-voltage warning detection. Reading and writing to this register should use the coefficients shown in the [Table](#page-35-0) 39. Accesses to this command should use the PMBus read or write word protocol. If the measured value of VIN falls below the value in this register, VIN under-voltage warn flags are set in the respective register, and the SMBA signal is asserted.

Table 7. VIN_UV_WARN_LIMIT Register

STATUS_BYTE (78h)

The STATUS BYTE is a standard PMBus command that returns the value of a number of flags indicating the state of the LM5056/LM5056A. Accesses to this command should use the PMBus read byte protocol. To clear bits in this register, the underlying fault should be removed on the system and a CLEAR_FAULTS command issued.

Table 8. STATUS_BYTE Definitions

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Table 8. STATUS_BYTE Definitions (continued)

STATUS_WORD (79h)

The STATUS_WORD command is a standard PMBus command that returns the value of a number of flags indicating the state of the LM5056/LM5056A. Accesses to this command should use the PMBus read word protocol. To clear bits in this register, the underlying fault should be removed and a CLEAR _FAULTS command issued.

Table 9. STATUS_WORD Definitions

STATUS_VOUT (7Ah)

The STATUS_VOUT command is a standard PMBus command that returns the value of the VOUT under-voltage warn flag. Accesses to this command should use the PMBus read byte protocol. To clear bits in this register, the underlying fault should be cleared and a CLEAR_FAULTS command issued.

Table 10. STATUS_VOUT Definitions

STATUS_INPUT (7Ch)

The STATUS_INPUT command is a standard PMBus command that returns the value of a number of flags related to input voltage, current, and power. Accesses to this command should use the PMBus read byte protocol. To clear bits in this register, the underlying fault should be cleared and a CLEAR_FAULTS command issued.

Table 11. STATUS_INPUT Definitions

STATUS_TEMPERATURE (7dh)

The STATUS_TEMPERATURE is a standard PMBus command that returns the value of the of a number of flags related to the temperature telemetry value. Accesses to this command should use the PMBus read byte protocol. To clear bits in this register, the underlying fault should be cleared and a CLEAR_FAULTS command issued.

Table 12. STATUS_TEMPERATURE Definitions

STATUS_CML (7Eh)

The STATUS CML is a standard PMBus command that returns the value of a number of flags related to communication faults. Accesses to this command should use the PMBus read byte protocol. To clear bits in this register, a CLEAR_FAULTS command should be issued.

Table 13. STATUS_CML Definitions

STATUS_MFR_SPECIFIC (80h)

The STATUS_MFR_SPECIFIC command is a standard PMBus command that contains manufacturer specific status information. Accesses to this command should use the PMBus read byte protocol. To clear bits in this register, the underlying fault should be removed and a CLEAR_FAULTS command should be issued.

Table 14. STATUS_MFR_SPECIFIC Definitions

READ_VIN (88h)

The READ_VIN command is a standard PMBus command that returns the 12-bit measured value of the input voltage (VIN to AGND). Reading this register should use the coefficients shown in the [Table](#page-35-0) 39. Accesses to this command should use the PMBus read word protocol. This value is also used internally for the VIN over and under-voltage warning detection.

Table 15. READ_VIN Register

READ_VOUT (8Bh)

The READ_VOUT command is a standard PMBus command that returns the 12-bit measured value of the output voltage. Reading this register should use the coefficients shown in the [Table](#page-35-0) 39 Table. Accesses to this command should use the PMBus read word protocol. This value is also used internally for the VOUT under_voltage warning detection.

Table 16. READ_VOUT Register

READ_TEMPERATURE_1 (8Dh)

The READ_TEMPERATURE_1 command is a standard PMBus command that returns the signed value of the temperature measured by the external temperature sense diode. Reading this register should use the coefficients shown in the [Table](#page-35-0) 39. Accesses to this command should use the PMBus read word protocol. This value is also used internally for the over-temperature fault and warning detection. This data has a range of -256°C to 255°C after the coefficients are applied.

Table 17. READ_TEMPERATURE_1 Register

MFR_ID (99h)

The MFR_ID command is a standard PMBus command that returns the identification of the manufacturer. To read the MFR_ID, use the PMBus block read protocol.

Table 18. MFR_ID Register

MFR_MODEL (9Ah)

The MFR_MODEL command is a standard PMBus command that returns the part number of the chip. To read the MFR_MODEL, use the PMBus block read protocol.

Table 19. MFR_MODEL Register

MFR_REVISION (9Bh)

The MFR_REVISION command is a standard PMBus command that returns the revision level of the part. To read the MFR_REVISION, use the PMBus block read protocol.

Table 20. MFR_REVISION Register

Manufacturer Specific PMBus™ Commands

MFR_SPECIFIC_00: MFR_READ_VAUX (D0h)

The MFR, READ, VAUX command will report the 12-bit ADC measured auxiliary voltage. Voltages greater than or equal to 2.97 V to ground are reported at plus full scale (0FFFh). Voltages less than or equal to 0 V referenced to ground are reported as 0 (0000h). To read data from the MFR_READ_VAUX command, use the PMBus Read Word protocol.

Table 21. MFR_READ_VAUX Register

MFR_SPECIFIC_01: MFR_READ_IIN (D1h)

The MFR, READ, IIN command reports the 12-bit ADC measured current sense voltage. To read data from the MFR READ IIN command, use the PMBus Read Word protocol. Reading this register should use the coefficients shown in the [Table](#page-35-0) 39. Please see the section on coefficient calculations to calculate the values to use.

Table 22. MFR_READ_IIN Register

MFR_SPECIFIC_02: MFR_READ_PIN (D2h)

The MFR_READ_PIN command reports the upper 12 bits of the VIN x IIN product as measured by the 12-bit ADC. To read data from the MFR READ PIN command, use the PMBus Read Word protocol. Reading this register should use the coefficients shown in the [Table](#page-35-0) 39. Please see the section on coefficient calculations to calculate the values to use.

Table 23. MFR_READ_PIN Register

MFR_SPECIFIC_03: MFR_IN_OC_WARN_LIMIT (D3h)

The MFR_IIN_OC_WARN_LIMIT PMBus command sets the input over-current warning threshold. In the event that the input current rises above the value set in this register, the IIN over-current flags are set in the respective registers and the SMBA is asserted. To access the MFR_IIN_OC_WARN_LIMIT register, use the PMBus Read/Write Word protocol. Reading/writing to this register should use the coefficients shown in the [Table](#page-35-0) 39.

Table 24. MFR_IIN_OC_WARN_LIMIT Register

MFR_SPECIFIC_04: MFR_PIN_OP_WARN_LIMIT (D4h)

The MFR_PIN_OP_WARN_LIMIT PMBus command sets the input over-power warning threshold. In the event that the input power rises above the value set in this register, the PIN over-power flags are set in the respective registers and the SMBA is asserted. To access the MFR_PIN_OP_WARN_LIMIT register, use the PMBus Read/Write Word protocol. Reading/writing to this register should use the coefficients shown in the [Table](#page-35-0) 39.

Table 25. MFR_PIN_OPWARN_LIMIT Register

MFR_SPECIFIC_05: MFR_READ_PIN_PEAK (D5h)

The MFR_READ_PIN_PEAK command reports the maximum input power measured since a power-on reset or the last MFR_CLEAR_PIN_PEAK command. To access the MFR_READ_PIN_PEAK command, use the PMBus Read Word protocol. Use the [Table](#page-35-0) 39.

Table 26. MFR_READ_PIN_PEAK Register

MFR_SPECIFIC_06: MFR_CLEAR_PIN_PEAK (D6h)

The MFR_CLEAR_PIN_PEAK command clears the MFR_PIN_PEAK register. This command uses the PMBus Send Byte protocol.

MFR_SPECIFIC_08: MFR_ALERT_MASK (D8h)

The MFR_ALERT_MASK command is used to mask the **SMBA** when a specific fault or warning has occurred. Each bit corresponds to one of the 8 different analog and digital faults or warnings that would normally result in an SMBA being asserted. When the corresponding bit is high, that condition will not cause the SMBA to be asserted. If that condition occurs, the registers where that condition is captured will still be updated (STATUS registers, DIAGNOSTIC_WORD) . This register is accessed with the PMBus Read and Write Word protocol.

Table 27. MFR_ALERT_MASK Definitions

MFR_SPECIFIC_09: MFR_DEVICE_SETUP (D9h)

The MFR_DEVICE_SETUP command may be used to override pin settings to define operation of the LM5056/LM5056A under host control. This command is accessed with the PMBus read and write byte protocol.

Table 28. MFR_DEVICE_SETUP Byte Format

In order to configure the current sense gain via this register, it is necessary to set the Current Sense Gain Select Configuration bit (2) to 1 to enable the register to control the current sense gain and the Current Sense Gain bit (4) to select the desired setting. If the Current Sense Gain Select Configuration bit is not set, the pin setting is used.

NOTE

If the Current Sense Gain Select Configuration is changed, the samples for the telemetry averaging function will not be reset. It is recommended to allow a full averaging update period with the new Current Sense Gain before processing the averaged data.

The Current Sense Gain Select Configuration affects the coefficients used for the current and power measurements and warning registers.

MFR_SPECIFIC_10: MFR_BLOCK_READ (DAh)

The MFR_BLOCK_READ command concatenates the MFR_DIAGNOSTIC_WORD with input and output telemetry information (IIN, VOUT, VIN, PIN) as well as TEMPERATURE to capture all of the operating information of the LM5056/LM5056A in a single SMBus transaction. The block is 12 bytes long with telemetry information being sent out in the same manner as if an individual READ_XXX command had been issued (shown below). The contents of the MFR_BLOCK_READ register are updated every clock cycle (85 ns) as long as the SMBus interface is idle. MFR_BLOCK_READ also ensures that the VIN, VOUT, IIN and PIN measurements are all time-aligned. If separate commands are used, individual samples may not be time-aligned, because of the delay necessary for the communication protocol. The MFR_BLOCK_READ command is read via the PMBus block read protocol.

Table 29. MFR_BLOCK_READ Register Format

MFR_SPECIFIC_11: MFR_SAMPLES_FOR_AVG (DBh)

The MFR SAMPLES FOR AVG command is a manufacturer specific command for setting the number of samples used in computing the average values for IIN, VIN, VOUT, PIN. The decimal equivalent of the AVGN nibble is the power of two samples, (e.g. $AVGN = 12$ equates to $N = 4096$ samples used in computing the average). The LM5056/LM5056A supports average numbers of 1, 2, 4, 8, 16, 32, 64, 128, 256, 512, 1024, 2048, 4096. The MFR_SAMPLES_FOR_AVG number applies to average values of IIN, VIN, VOUT, PIN simultaneously. The LM5056/LM5056A uses simple averaging. This is accomplished by summing consecutive results up to the number programmed, then dividing by the number of samples. Averaging is calculated according to the following sequence:

$$
Y = \frac{(X(N) + X(N-1) + ... + X(0))}{N}
$$

(2)

When the averaging has reached the end of a sequence (for example, 4096 samples are averaged), then a whole new sequence begins that requires the same number of samples (in this example, 4096) to be taken before the new average is ready.

Table 30. MFR_SAMPLES_FOR_AVG Register

NOTE

A change in the MFR_SAMPLES_FOR_AVG register are not reflected in the average telemetry measurements until the present averaging interval has completed. The default setting for AVGN is 0000, therefore, the average telemetry mirrors the instantaneous telemetry until a value higher than zero is programmed.

The MFR_SAMPLES_FOR_AVG register is accessed via the PMBus read and write byte protocol.

Table 31. MFR_SAMPLES_FOR_AVG Register

MFR_SPECIFIC_12: MFR_READ_AVG_VIN (DCh)

The MFR_READ_AVG_VIN command will report the 12-bit ADC measured input average voltage. If the data is not ready, the returned value is the previous averaged data. However, if there is no previously averaged data, the default value (0000h) is returned. This data is read with the PMBus Read Word protocol. This register should use the coefficients shown in the [Table](#page-35-0) 39.

Table 32. MFR_READ_AVG_VIN Register

MFR_SPECIFIC_13: MFR_READ_AVG_VOUT (DDh)

The MFR_READ_AVG_VOUT command reports the 12-bit ADC measured current sense average voltage. The returned value is the default value (0000h) or previous data when the average data is not ready. This data is read with the PMBus Read Word protocol. This register should use the coefficients shown in the [Table](#page-35-0) 39.

Table 33. MFR_READ_AVG_VOUT Register

MFR_SPECIFIC_14: MFR_READ_AVG_IIN (DEh)

The MFR READ AVG IIN command reports the 12-bit ADC measured current sense average voltage. The returned value is the default value (0000h) or previous data when the average data is not ready. This data is read with the PMBus Read Word protocol. This register should use the coefficients shown in the [Table](#page-35-0) 39.

Table 34. MFR_READ_AVG_IIN Register

MFR_SPECIFIC_15: MFR_READ_AVG_PIN

The MFR, READ, AVG, PIN command reports the upper 12-bits of the average VIN x IIN product as measured by the 12-bit ADC. Read the default value (0000h) or previous data when the average data is not ready. This data is read with the PMBus Read Word protocol. This register should use the coefficients shown in the [Table](#page-35-0) 39.

Table 35. TABLE 35. MFR_READ_AVG_PIN Register

MFR_SPECIFIC_16: MFR_BLACK_BOX_READ (E0h)

The MFR_BLACK_BOX_READ command retrieves the MFR_BLOCK_READ data which was latched in at the first assertion of SMBA by the LM5056/LM5056A. It is re-armed with the CLEAR_FAULTS command. It is the same format as the MFR_BLOCK_READ registers, the only difference being that its contents are updated with the SMBA edge rather than the internal clock edge. This command is read with the PMBus Block Read protocol.

MFR_SPECIFIC_17: MFR_READ_DIAGNOSTIC_WORD (E1h)

The MFR READ DIAGNOSTIC WORD PMBus command reports all of the LM5056/LM5056A faults and warnings in a single read operation. The standard response to the assertion of the SMBA signal of issuing multiple read requests to various status registers can be replaced by a single word read to the MFR_DIAGNOSTIC_WORD register. The MFR_READ_DIAGNOSTIC_WORD command should be read with the PMBus Read Word protocol. The MFR_READ_DIAGNOSTIC_WORD is also returned in the MFR_BLOCK_READ, MFR_BLACK_BOX_READ, and MFR_AVG_BLOCK_READ operations.

Table 36. MFR_DIAGNOSTIC_WORD Format

MFR_SPECIFIC_18: MFR_AVG_BLOCK_READ (E2h)

The MFR_AVG_BLOCK_READ command concatenates the MFR_DIAGNOSTIC_WORD with input and output average telemetry information (IIN, VOUT, VIN, PIN) as well as temperature to capture all of the operating information of the part in a single PMBus transaction. The block is 12 bytes long with telemetry information being sent out in the same manner as if an individual READ_AVG_XXX command had been issued (shown below). MFR_AVG_BLOCK_READ also ensures that the VIN, VOUT, and IIN measurements are all time-aligned whereas there is a chance they may not be if read with individual PMBus commands. To read data from the MFR_AVG_BLOCK_READ command, use the SMBus Block Read protocol.

Table 37. MFR_AVG_BLOCK_READ Register Format

To load

MFR_DIAGNOSTIC_WORD_READ E1h OVER-TEMP WARN
STATUS_TEMPERATURE 7Dh OVER-TEMP FAULT
STATUS_TEMPERATURE7Dh STATUS_TEMPERATURE 7Dh STATUS_TEMPERATURE 7Dh **WARNING SYSTEM** VIN OV WARN
STATUS_INPUT 7 Ch IIN OC WARN
STATUS_INPUT 7 Ch PIN OP WARN
STATUS_INPUT 7Ch STATUS_INPUT 7Ch **VIN UV WARN
STATUS_INPUT 7Ch** STATUS_INPUT 7Ch STATUS_INPUT 7Ch MFR_PIN_OP ALL D4h CMP PIN OP WARN STATUS_INFUT OP WARN STATUS_INFUT 7Ch VOUT UVWARN
STATUS_VOUT 7Ah OVER-TEMP FAULT STATUS_VOUT 7Ah OVER-TEMP WARN VOUT UV WARN П П **NEK** <u>⁄हे</u> م
B e)
⊝ e)
c ا∳
© م¶
⊝ $\bar{\tilde{\varepsilon}}$ $LMITDAh$ MFR_IN_OC_WARN_UMTD3h MFR_IIN_OC_WARN_LIMIT D3h 58 VOUT_UV_WARN_LIMIT 58h VIN_OV_WARN_LIMIT 57h VIN_UV_WARN_LIMIT 58h VIN_UV_WARN_LIMIT 58h OT WARNING LIMIT 51h OT_WARNING_LIMIT 51h VIN OV_WARN_LIMIT 57 OT_FAULT_LIMIT 4Fh OT_FAULT_LIMIT 4FH <u>LIMIT NY WY MENT</u> **WARNING LIMITS** MFR_PIN_OP_WARN **PMBus Interface PMBus Interface** AVG DBI MFR_SAMPLES_FOR_AVG DBh MFR_READ_AVG_VOUT DDh MFR_CLEAR_PIN_PEAK D6h MFR_READ_AVG_VOUT DDY MFR_READ_PIN_PEAK D5h MFR_CLEAR_PIN_PEAKD6 ADC MFR_READ_AVG_VIN DCh MFR_READ_AVG_PIN DFh MUX**K** MFR_READ_AVG_IIN DEh MFR READ PIN PEA $\sum_{i=1}^{n}$ **AVERAGED PEAK-HOLD** le **DATA** MFR SAMPLES ıιx S/H \equiv VOUT $\overline{\mathbf{z}}$

READ_VIN 88h MFR_READ_IIN D1h MFR_READ_PIN D2h READ_VOUT 8Bh READ_TEMPER ATUR E_1 8Dh

MFR_READ_IND READ_VIN 88H

MFR_READ_PN D2 READ_VOUT 8Bh am
B

EMPERATURE

ą ₩

DATA OUTPUT

MFR_READ_VAUX D0h

MFR

VAUX

- +

VIN_K SENSE

SENSE

<u>S</u>

 $\frac{1}{2}$

 $rac{2}{5}$

 $\frac{48}{3}$

DIODE

Reading and Writing Telemetry Data and Warning Thresholds

All measured telemetry data and user programmed warning thresholds are communicated in 12-bit two's compliment binary numbers read and written in 2-byte increments conforming to the direct format as described in section 8.3.3 of the PMBus Power System Management Protocol Specification 1.1 (Part II). The organization of the bits in the telemetry or warning word is shown in [Table](#page-34-0) 38, where Bit_11 is the most significant bit (MSB) and Bit_0 is the least significant bit (LSB). The decimal equivalent of all warning and telemetry words are constrained to be within the range of 0 to 4095, with the exception of temperature. The decimal equivalent value of the temperature word ranges from 0 to 65535.

Conversion from direct format to real-world dimensions of current, voltage, power, and temperature is accomplished by determining appropriate coefficients as described in section 7.2.1 of the PMBus Power System Management Protocol Specification 1.1 (Part II). According to this specification, the host system converts the values received into a reading of volts, amperes, watts, or other units using the following relationship:

$$
X = \frac{1}{m} \left(Y \times 10^{-R} - B\right)
$$

where

- X: the calculated real-world value (volts, amps, watt, etc.)
- m: the slope coefficient
- Y: a two byte two's complement integer received from device
- b: the offset, a two byte, two's complement integer
- R: the exponent, a one byte two's complement integer (3)

R is only necessary in systems where m is required to be an integer (for example, where m may be stored in a register in an integrated circuit). In those cases, R only needs to be large enough to yield the desired accuracy.

Table 39. Telemetry and Warning Conversion Coefficients

(1) The coefficients relating to current/power measurements and warning thresholds shown in [Table](#page-35-0) 39 are normalized to a sense resistor (RS) value of 1mΩ. In general, the current/power coefficients can be calculated using the relationships shown in [Table](#page-35-1) 40.

Care must be taken to adjust the exponent coefficient, R, such that the value of m remains within the range of - 32768 to 32767. For example, if a 5-mΩ sense resistor (RS) is used, the correct coefficients for the READ_IIN command with $CL = VDD$ would be m = 3736, b = 195, R = -1.

Note: The power coefficients given in [Table](#page-35-0) 39 are characterized at a specific operating point of 48-V VIN. If high-power accuracy is desired at voltages other than 48 V, it is recommended to read VIN and IIN using the MFR_BLOCK_READ (DAh) command. After finding the real-world value of VIN and IIN using the coefficients, the power can simply be calculated by the multiplication of the two measurements. This will ensure the user obtains the highest accuracy power measurement. Another method to ensure accurate telemetry is to find new coefficients for your specfic application. This is outlined in the proceeding section.

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Determining Telemetry Coefficients Empirically with Linear Fit

The coefficients for telemetry measurements and warning thresholds presented in [Table](#page-35-0) 39 are adequate for the majority of applications. Current and power coefficients must be calculated per application as they are dependent on the value of the sense resistor, R_s , used. [Table](#page-35-1) 40 provides the equations necessary for calculating the current and power coefficients for the general case. The small signal nature of the current measurement make it and the power measurement more susceptible to PCB parasitics than other telemetry channels. This may cause slight variations in the optimum coefficients (m, b, R) for converting from Direct format digital values to real-world values (e.g., Amps and Watts). The optimum coefficients can be determined empirically for a specific application and PCB layout using two or more measurements of the telemetry channel of interest. The current coefficients can be determined using the following method:

- 1. While the LM5056/LM5056A is in normal operation measure the voltage across the sense resistor using Kelvin test points and a high accuracy DVM while controlling the load current. Record the integer value returned by the MFR_READ_AVG_IIN command (with the MFR_SAMPLES_FOR_AVG set to a value greater than 0) for two or more voltages across the sense resistor. For best results, the individual MFR_READ_AVG_IIN measurements should span nearly the full scale range of the current (For example, voltage across RS of 5 mV and 20 mV).
- 2. Convert the measured voltages to currents by dividing them by the value of R_S. For best accuracy the value of R_S should be measured. [Table](#page-36-0) 41 assumes a sense resistor value of 5 m Ω .

- 3. Using the spreadsheet or math program of your choice determine the slope and the y-intercept of the data returned by the READ_AVG_IIN command versus the measured current. For the data shown in [Table](#page-36-0) 41:
	- READ_AVG_IN value = slope x (Measured Current) + (y-intercept)
	- $-$ slope = 538.9
	- $-$ y-intercept = 29.5
- 4. To determine the 'm' coefficient, simply shift the decimal point of the calculated slope to arrive at at integer with a suitable number of significant digits for accuracy (typically 4) while staying with the range of -32768 to +32767. This shift in the decimal point equates to the 'R' coefficient. For the slope value shown above, the decimal point would be shifted to the right once hence $R = -1$.
- 5. Once the 'R' coefficient has been determined, the 'b' coefficient is found by multiplying the y-intercept by 10- R. In this case the value of $b = 295$. Calculated current coefficients:
	- $m = 5389$
	- $b = 295$
	- $R = -1$

$$
X = \frac{1}{m} \left(Y \times 10^{-R} - b\right)
$$

where

- X: the calculated real-world value (volts, amps, watts, temperature)
- m: the slope coefficient, is the two byte, two's complement integer
- Y: a two byte two's complement integer received from device
- b: the offset, a two byte, two's complement integer
- R: the exponent, a one byte two's complement integer (4)

The above procedure can be repeated to determine the coefficients of any telemetry channel simply by substituting measured current for some other parameter (e.g., power, voltage, etc.). Note that the above procedure can be executed using the PMBus software GUI found in the LM5056/LM5056A online product folder

Writing Telemetry Data

There are several locations that require writing data if their optional usage is desired. Use the same coefficients previously calculated for your application, and apply them using this method as prescribed by the PMBus revision section 7.2.2 Sending a Value.

 $Y = (mX + b) \times 10^R$

where

- X: the calculated real-world value (volts, amps, watts, temperature)
- m: the slope coefficient, is the two byte, two's complement integer
- Y: a two byte two's complement integer received from device
- b: the offset, a two byte, two's complement integer
- R: the exponent, a one byte two's complement integer (5) (5)

PMBus™ Address Lines (ADR0, ADR1, ADR2)

The three address lines are to be set high (connect to VDD), low (connect to GND), or open to select one of 27 addresses for communicating with the LM5056/LM5056A. [Table](#page-37-0) 42 depicts 7-bit addresses (eighth bit is read and write bit):

Table 42. Device Addressing

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Figure 19. SMBus Timing Diagram

Table 43. SMBus Timing Definition

(1) Devices participating in a transfer will timeout when any clock low exceeds the value of TTIMEOUT,MIN of 25 ms. Devices that have detected a timeout condition must reset the communication no later than TTIMEOUT,MAX of 35 ms. The maximum value must be adhered to by both a master and a slave as it incorporates the cumulative stretch limit for both a master (10 ms) and a slave (25 ms).

(2) THIGH MAX provides a simple method for devices to detect bus idle conditions. (3) $T_{LOW:SEXT}$ is the cumulative time a slave device is allowed to extend the clock cycles in one message from the initial start to the stop. If a slave exceeds this time, it is expected to release both its clock and data lines and reset itself.

(4) TLOW:MEXT is the cumulative time a master device is allowed to extend its clock cycles within each byte of a message as defined from start-to-ack, ackto- ack, or ack-to-stop.

(5) Rise and fall time is defined as follows: (a) $T_R = (V_{ILMAX} - 0.15)$ to $(V_{IHMIN} + 0.15)$

(b) $T_F = 0.9 \text{ VDD}$ to $(V_{ILMAX} - 0.15)$

SMBA Response

The SMBA effectively has two masks:

- 1. The Alert Mask Register at D8h, and
- 2. The ARA Automatic Mask.

The ARA Automatic Mask is a mask that is set in response to a successful ARA read. An ARA read operation returns the PMBus address of the lowest addressed part on the bus that has its SMBA asserted. A successful ARA read means that THIS part was the one that returned its address. When a part responds to the ARA read, it releases the SMBA signal. When the last part on the bus that has an SMBA set has successfully reported its address, the SMBA signal de-asserts.

The way that the LM5056/LM5056A releases the SMBA signal is by setting the ARA Automatic mask bit for all fault conditions present at the time of the ARA read. All status registers still show the fault condition, but it is not generated and SMBA on that fault again until the ARA Automatic mask is cleared by the host issuing a CLEAR_FAULTS command to this part. This should be done as a routine part of servicing an SMBA condition on a part, even if the ARA read is not done. [Figure](#page-39-0) 20 depicts a schematic version of this flow.

Figure 20. Typical Flow Schematic for SMBA Fault

PACKAGING INFORMATION

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the \leq =1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

PACKAGE OPTION ADDENDUM

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TEXAS

TAPE AND REEL INFORMATION

ISTRUMENTS

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

PACKAGE MATERIALS INFORMATION

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*All dimensions are nominal

TEXAS INSTRUMENTS

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TUBE

B - Alignment groove width

*All dimensions are nominal

PACKAGE OUTLINE

PWP0028A PowerPAD - 1.1 mm max height TM

PLASTIC SMALL OUTLINE

NOTES:

PowerPAD is a trademark of Texas Instruments.

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
- 5. Reference JEDEC registration MO-153, variation AET.

EXAMPLE BOARD LAYOUT

PWP0028A PowerPAD™ - 1.1 mm max height

PLASTIC SMALL OUTLINE

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
- 8. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature numbers SLMA002 (www.ti.com/lit/slma002) and SLMA004 (www.ti.com/lit/slma004).
- 9. Size of metal pad may vary due to creepage requirement.

EXAMPLE STENCIL DESIGN

PWP0028A PowerPAD™ - 1.1 mm max height

PLASTIC SMALL OUTLINE

NOTES: (continued)

10. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

11. Board assembly site may have different recommendations for stencil design.

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