

LinCMOS™ PRECISION QUAD OPERATIONAL AMPLIFIERS

 Check for Samples: [TLC27M4](#), [TLC27M4A](#), [TLC27M4B](#), [TLC27M4Y](#), [TLC27M9](#)

FEATURES

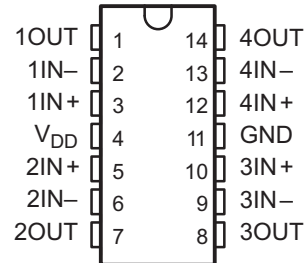
- Trimmed Offset Voltage
 - TLC27M9 . . . 900 μV Max at $T_A = 25^\circ\text{C}$, $V_{DD} = 5\text{ V}$
- Input Offset Voltage Drift . . . Typically 0.1 $\mu\text{V}/\text{Month}$, Including the First 30 Days
- Wide Range of Supply Voltages Over Specified Temperature Range:
 - 0°C to 70°C . . . 3 V to 16 V
 - -40°C to 85°C . . . 4 V to 16 V
 - -55°C to 125°C . . . 4 V to 16 V
- Single-Supply Operation
- Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix Types)
- Low Noise . . . Typically 32 $\text{nV}/\sqrt{\text{Hz}}$ at $f = 1\text{ kHz}$
- Low Power . . . Typically 2.1 mW at $T_A = 25^\circ\text{C}$, $V_{DD} = 5\text{ V}$
- Output Voltage Range Includes Negative Rail
- High Input Impedance . . . $10^{12}\ \Omega$ Typ
- ESD-Protection Circuitry
- Small-Outline Package Option Also Available in Tape and Reel
- Designed-In Latch-Up Immunity

DESCRIPTION

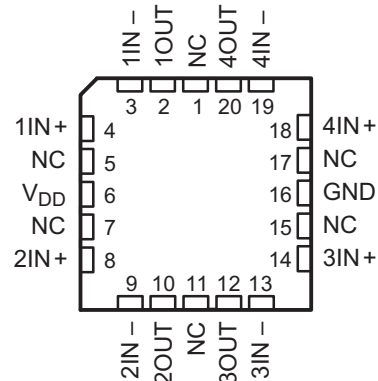
The TLC27M4 and TLC27M9 quad operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds comparable to that of general-purpose bipolar devices. These devices use Texas Instruments silicon-gate LinCMOS™ technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, make these cost-effective devices ideal for applications that have previously been reserved for general-purpose bipolar products, but with only a fraction of the power consumption.

D, J, N, OR PW PACKAGE (TOP VIEW)

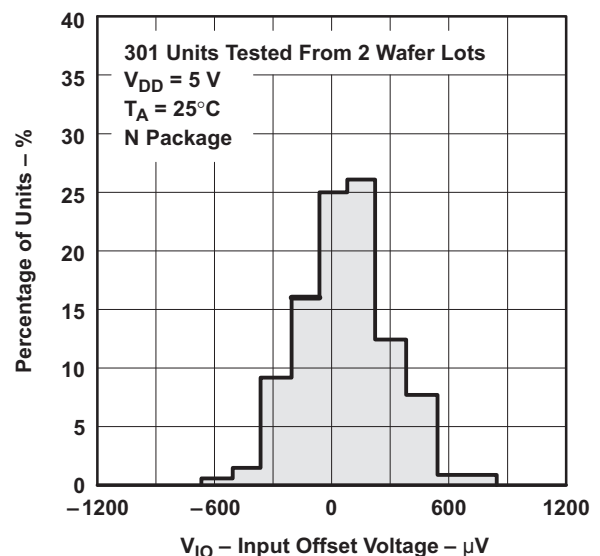


FK PACKAGE (TOP VIEW)



NC – No internal connection

DISTRIBUTION OF TLC27M9 INPUT OFFSET VOLTAGE



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Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC27M4 (10 mV) to the high-precision TLC27M9 (900 μV). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available on LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC27M4 and TLC27M9. The devices also exhibit low voltage single-supply operation, and low power consumption, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip-carrier versions for high-density system applications.

The device inputs and outputs are designed to withstand –100-mA surge currents without sustaining latch-up.

The TLC27M4 and TLC27M9 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015; however, care should be exercised in handling these devices, as exposure to ESD may result in the degradation of the device parametric performance.

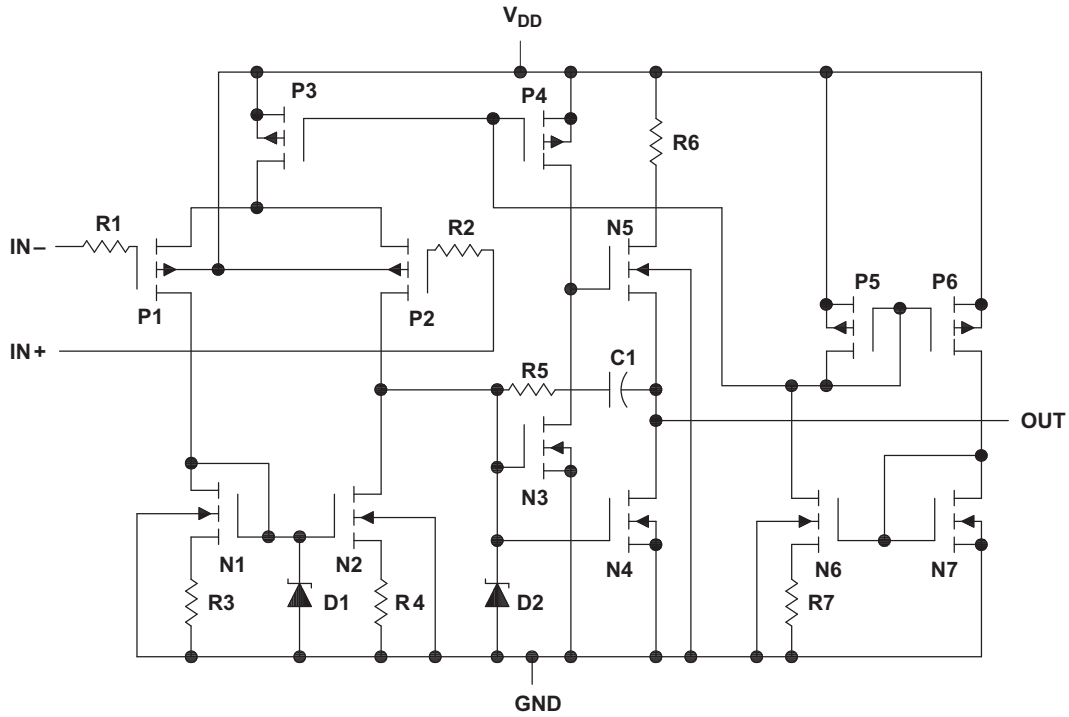
The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from –40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of –55°C to 125°C.

AVAILABLE OPTIONS

T _A	V _{IO} max AT 25°C	PACKAGE					CHIP FORM (Y)
		SMALL OUTLINE (D) ⁽¹⁾	CHIP CARRIER (FK)	CERAMIC DIP (J)	PLASTIC DIP (N)	TSSOP (PW) ⁽¹⁾	
0°C to 70°C	900 μV	TLC27M9CD	—	—	TLC27M9CN	—	—
	2 mV	TLC27M4BCD	—	—	TLC27M4BCN	—	—
	5 mV	TLC27M4ACD	—	—	TLC27M4ACN	—	—
	10 mV	TLC27M4CD	—	—	TLC27M4CN	TLC27M4CPW	TLC27M4Y
–40°C to 85°C	900 μV	TLC27M9ID	—	—	TLC27M9IN	—	—
	2 mV	TLC27M4BID	—	—	TLC27M4BIN	—	—
	5 mV	TLC27M4AID	—	—	TLC27M4AIN	—	—
	10 mV	TLC27M4ID	—	—	TLC27M4IN	TLC27M41PW	—
–55°C to 125°C	900 μV	TLC27M9MD	TLC27M9MFK	TLC27M9MJ	TLC27M9MN	—	—
	10 mV	TLC27M4MD	TLC27M4MFK	TLC27M4MJ	TLC27M4MN	—	—

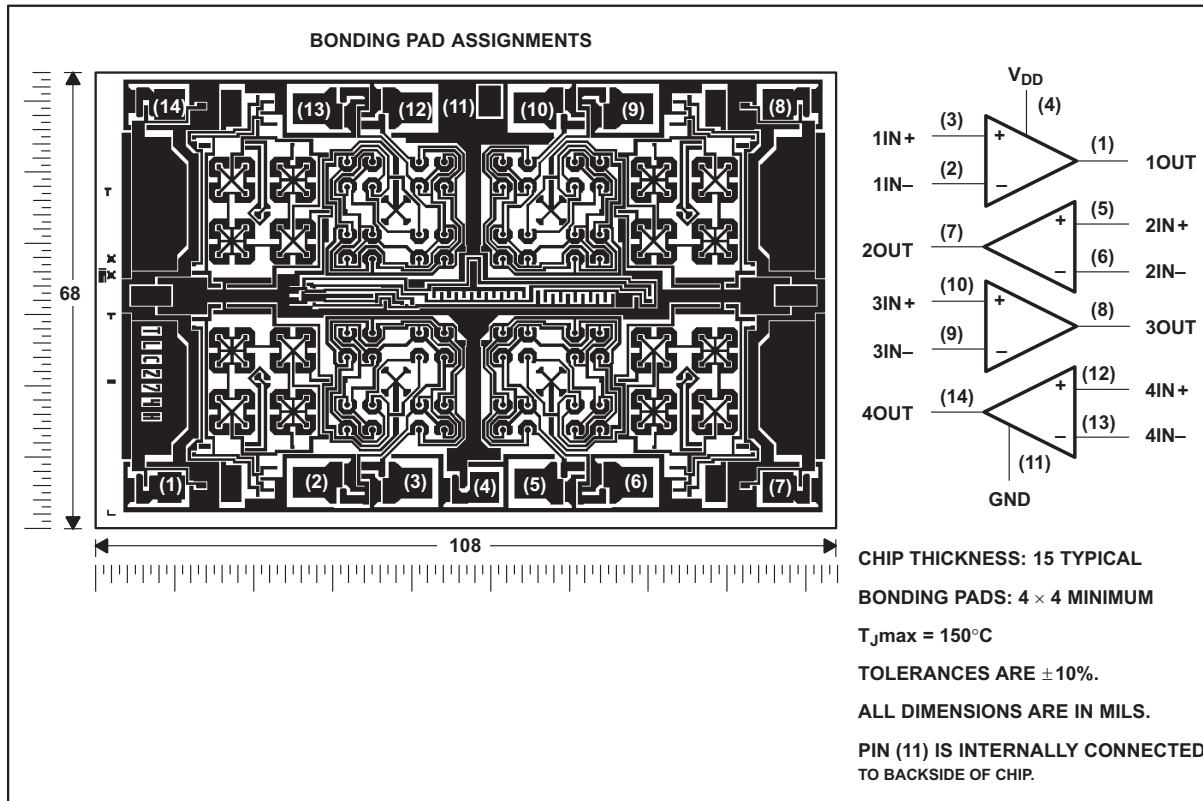
(1) The D and PW package is available taped and reeled. Add R suffix to the device type (e.g., TLC279CDR).

EQUIVALENT SCHEMATIC (EACH AMPLIFIER)



TLC27M4Y chip information

This chip, when properly assembled, displays characteristics similar to the TLC27M4C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



ABSOLUTE MAXIMUM RATINGS

 over operating free-air temperature range (unless otherwise noted)⁽¹⁾

	VALUE	UNIT
Supply voltage, V_{DD} ⁽²⁾	18	V
Differential input voltage, V_{ID} ⁽³⁾	$\pm V_{DD}$	
Input voltage range, V_I (any input)	$-0.3\text{ V to }V_{DD}$	
Input current, I_I	± 5	mA
Output current, I_O (each output)	± 30	mA
Total current into V_{DD}	45	mA
Total current out of GND	45	mA
Duration of short-circuit current at (or below) 25°C ⁽⁴⁾	unlimited	
Continuous total dissipation	See Dissipation Rating Table	
Operating free-air temperature, T_A	C suffix	0 to 70 °C
	I suffix	-40 to 85 °C
	M suffix	-55 to 125 °C
Storage temperature range	-65 to 150	°C
Case temperature for 60 seconds: FK package	260	°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, N, or PW package	260	°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300	°C

- (1) Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values, except differential voltages, are with respect to network ground.
- (3) Differential voltages are at $IN+$ with respect to $IN-$.
- (4) The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

DISSIPATION RATINGS

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW	—
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
J	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
N	1575 mW	12.6 mW/°C	1008 mW	819 mW	—
PW	700 mW	5.6 mW/°C	448 mW	—	—

RECOMMENDED OPERATING CONDITIONS

	MIN	MAX	MIN	MAX	MIN	MAX	UNIT
Supply voltage, V_{DD}	3	16	4	16	4	16	V
Common mode input voltage, V_{IC}	$V_{DD} = 5\text{ V}$	-0.2	3.5	-0.2	3.5	0	3.5
	$V_{DD} = 10\text{ V}$	-0.2	8.5	-0.2	8.5	0	8.5
Operating free-air temperature, T_A	0	70	-40	85	-55	125	°C

ELECTRICAL CHARACTERISTICS

at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A^{(1)}$	TLC27M4C TLC27M4AC TLC27M4BC TLC27M9C			UNIT
					MIN	TYP	MAX	
V_{IO}	Input offset voltage	TLC27M4C	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27M4AC	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	0.9	5	mV
					Full range		6.5	
		TLC274BC	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	250	2000	μV
					Full range		3000	
		TLC279C	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	210	900	μV
					Full range		1500	
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C	1.7		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current ⁽²⁾	$V_O = 2.5\text{ V},$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
I_{IB}	Input bias current ⁽²⁾	$V_O = 2.5\text{ V},$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				70°C	40	600		
V_{ICR}	Common-mode input voltage range ⁽³⁾			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV},$	$R_L = 100\text{ k}\Omega$	25°C	3.2	3.9	V	
				0°C	3	3.9		
				70°C	3	4		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV},$	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V},$	$R_L = 100\text{ k}\Omega$	25°C	25	170	V/mV	
				0°C	15	200		
				70°C	15	140		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	91	dB	
				0°C	60	91		
				70°C	60	92		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V},$	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				0°C	60	92		
				70°C	60	94		
I_{DD}	Supply current (four amplifiers)	$V_O = 2.5\text{ V},$ No load	$V_{IC} = 2.5\text{ V},$	25°C	420	1120	μA	
				0°C	500	1280		
				70°C	340	880		

(1) Full range is 0°C to 70°C.

(2) The typical values of input bias current and input offset current below 5 pA were determined mathematically.

(3) This range also applies to each input individually.

ELECTRICAL CHARACTERISTICS

 at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A^{(1)}$	TLC27M4C TLC27M4AC TLC27M4BC TLC27M9C			UNIT
					MIN	TYP	MAX	
V_{IO}	Input offset voltage	TLC27M4C	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
		TLC27M4AC	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	0.9	5	
					Full range		6.5	
	TLC274BC	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	260	2000	μV	
				Full range		3000		
		TLC279C	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	220		1200
					Full range			1900
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 70°C	2.1		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current ⁽²⁾	$V_O = 5\text{ V},$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				70°C	7	300		
I_{IB}	Input bias current ⁽²⁾	$V_O = 5\text{ V},$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				70°C	50	600		
V_{ICR}	Common-mode input voltage range ⁽³⁾			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV},$	$R_L = 100\text{ k}\Omega$	25°C	8	8.7	V	
				0°C	7.8	8.7		
				70°C	7.8	8.7		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV},$	$I_{OL} = 0$	25°C	0	50	mV	
				0°C	0	50		
				70°C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V},$	$R_L = 100\text{ k}\Omega$	25°C	25	275	V/mV	
				0°C	15	320		
				70°C	15	230		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				0°C	60	94		
				70°C	60	94		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V},$	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				0°C	60	92		
				70°C	60	94		
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V},$ No load	$V_{IC} = 5\text{ V},$	25°C	570	1200	μA	
				0°C	690	1600		
				70°C	440	1120		

(1) Full range is 0°C to 70°C.

(2) The typical values of input bias current and input offset current below 5 pA were determined mathematically.

(3) This range also applies to each input individually.

ELECTRICAL CHARACTERISTICS

at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A^{(1)}$	TLC27M4I TLC27M4AI TLC27M4BI TLC27M9I			UNIT
					MIN	TYP	MAX	
V_{IO}	Input offset voltage	TLC27M4I	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC27M4AI	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	0.9	5	
					Full range		6.5	
	TLC27M4BI	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	250	2000	μV	
				Full range		3000		
		TLC27M9I	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	210		900
					Full range			2000
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C	1.7		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current ⁽²⁾	$V_O = 2.5\text{ V},$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA	
				85°C	24	1000		
I_{IB}	Input bias current ⁽²⁾	$V_O = 2.5\text{ V},$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA	
				85°C	200	2000		
V_{ICR}	Common-mode input voltage range ⁽³⁾			25°C	-0.2 to 4	-0.3 to 4.2	V	
				Full range	-0.2 to 3.5		V	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV},$	$R_L = 100\text{ k}\Omega$	25°C	3.2	3.9	V	
				-40°C	3	3.9		
				85°C	3	4		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV},$	$I_{OL} = 0$	25°C	0	50	mV	
				-40°C	0	50		
				85°C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 0.25\text{ V to }2\text{ V},$	$R_L = 100\text{ k}\Omega$	25°C	25	170	V/mV	
				-40°C	15	270		
				85°C	15	130		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	91	dB	
				-40°C	60	90		
				85°C	60	90		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V},$	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				-40°C	60	91		
				85°C	60	94		
I_{DD}	Supply current (four amplifiers)	$V_O = 2.5\text{ V},$ No load	$V_{IC} = 2.5\text{ V},$	25°C	420	1120	μA	
				-40°C	630	1600		
				85°C	320	800		

(1) Full range is -40°C to 85°C.

(2) The typical values of input bias current and input offset current below 5 pA were determined mathematically.

(3) This range also applies to each input individually.

ELECTRICAL CHARACTERISTICS

 at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A^{(1)}$	TLC27M4I TLC27M4AI TLC27M4BI TLC27M9I			UNIT
					MIN	TYP	MAX	
V_{IO}	Input offset voltage	TLC27M4I	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		13	
		TLC27M4AI	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	0.9	5	
					Full range		7	
	TLC27M4BI	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	260	2000	μV	
				Full range		3500		
		TLC27M9I	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	220		1200
					Full range			2900
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 85°C	2.1		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current ⁽²⁾	$V_O = 5\text{ V},$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA	
				85°C	26	1000		
I_{IB}	Input bias current ⁽²⁾	$V_O = 5\text{ V},$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA	
				85°C	220	2000		
V_{ICR}	Common-mode input voltage range ⁽³⁾			25°C	-0.2 to 9	-0.3 to 9.2	V	
				Full range	-0.2 to 8.5		V	
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV},$	$R_L = 100\text{ k}\Omega$	25°C	8	8.7	V	
				-40°C	7.8	8.7		
				85°C	7.8	8.7		
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV},$	$I_{OL} = 0$	25°C	0	50	mV	
				-40°C	0	50		
				85°C	0	50		
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V},$	$R_L = 100\text{ k}\Omega$	25°C	25	275	V/mV	
				-40°C	15	390		
				85°C	15	220		
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB	
				-40°C	60	93		
				85°C	60	94		
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V},$	$V_O = 1.4\text{ V}$	25°C	70	93	dB	
				-40°C	60	91		
				85°C	60	94		
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V},$ No load	$V_{IC} = 5\text{ V},$	25°C	570	1200	μA	
				-40°C	900	1800		
				85°C	410	1040		

(1) Full range is -40°C to 85°C.

(2) The typical values of input bias current and input offset current below 5 pA were determined mathematically.

(3) This range also applies to each input individually.

ELECTRICAL CHARACTERISTICS

at specified free-air temperature, $V_{DD} = 5\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		$T_A^{(1)}$	TLC27M4M TLC27M9M			UNIT
					MIN	TYP	MAX	
V_{IO}	Input offset voltage	TLC27M4M	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
					Full range		12	
	TLC27M9M	$V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	210	900	μV	
				Full range		3750		
α_{VIO}	Average temperature coefficient of input offset voltage			25°C to 125°C	1.7		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current ⁽²⁾		$V_O = 2.5\text{ V},$	$V_{IC} = 2.5\text{ V}$	25°C	0.1		pA
					125°C	1.4	15	nA
I_{IB}	Input bias current ⁽²⁾		$V_O = 2.5\text{ V},$	$V_{IC} = 2.5\text{ V}$	25°C	0.6		pA
					85°C	9	35	nA
V_{ICR}	Common-mode input voltage range ⁽³⁾				25°C	0 to 4	-0.3 to 4.2	V
					Full range	0 to 3.5		V
V_{OH}	High-level output voltage		$V_{ID} = 100\text{ mV},$	$R_L = 100\text{ k}\Omega$	25°C	3.2	3.9	V
					-55°C	3	3.9	
					125°C	3	4	
V_{OL}	Low-level output voltage		$V_{ID} = -100\text{ mV},$	$I_{OL} = 0$	25°C	0	50	mV
					-55°C	0	50	
					125°C	0	50	
A_{VD}	Large-signal differential voltage amplification		$V_O = 0.25\text{ V to }2\text{ V},$	$R_L = 100\text{ k}\Omega$	25°C	25	170	V/mV
					-55°C	15	270	
					125°C	15	120	
CMRR	Common-mode rejection ratio		$V_{IC} = V_{ICRmin}$		25°C	65	91	dB
					-55°C	60	89	
					125°C	60	91	
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)		$V_{DD} = 5\text{ V to }10\text{ V},$	$V_O = 1.4\text{ V}$	25°C	70	93	dB
					-55°C	60	91	
					125°C	60	94	
I_{DD}	Supply current (four amplifiers)		$V_O = 2.5\text{ V},$ No load	$V_{IC} = 2.5\text{ V},$	25°C	420	1120	μA
					-55°C	680	1760	
					125°C	280	720	

- (1) Full range is -55°C to 125°C.
- (2) The typical values of input bias current and input offset current below 5 pA were determined mathematically.
- (3) This range also applies to each input individually.

ELECTRICAL CHARACTERISTICS

 at specified free-air temperature, $V_{DD} = 10\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	$T_A^{(1)}$	TLC27M4M TLC27M9M			UNIT
				MIN	TYP	MAX	
V_{IO}	Input offset voltage	TLC27M4M $V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	1.1	10	mV
				Full range		12	
	TLC27M9M $V_O = 1.4\text{ V},$ $R_S = 50\ \Omega,$	$V_{IC} = 0,$ $R_L = 100\text{ k}\Omega$	25°C	220	1200	μV	
			Full range		4300		
α_{VIO}	Average temperature coefficient of input offset voltage		25°C to 125°C	2.1		$\mu\text{V}/^\circ\text{C}$	
I_{IO}	Input offset current ⁽²⁾	$V_O = 5\text{ V},$	$V_{IC} = 5\text{ V}$	25°C	0.1		pA
				125°C	1.8	15	nA
I_{IB}	Input bias current ⁽²⁾	$V_O = 5\text{ V},$	$V_{IC} = 5\text{ V}$	25°C	0.7		pA
				125°C	10	35	nA
V_{ICR}	Common-mode input voltage range ⁽³⁾			25°C	0 to 9	-0.3 to 9.2	V
				Full range	-0.2 to 8.5		V
V_{OH}	High-level output voltage	$V_{ID} = 100\text{ mV},$	$R_L = 100\text{ k}\Omega$	25°C	8	8.7	V
				-55°C	7.8	8.6	
				125°C	7.8	8.8	
V_{OL}	Low-level output voltage	$V_{ID} = -100\text{ mV},$	$I_{OL} = 0$	25°C	0	50	mV
				-55°C	0	50	
				125°C	0	50	
A_{VD}	Large-signal differential voltage amplification	$V_O = 1\text{ V to }6\text{ V},$	$R_L = 100\text{ k}\Omega$	25°C	25	275	V/mV
				-55°C	15	420	
				125°C	15	190	
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICRmin}$		25°C	65	94	dB
				-55°C	60	93	
				125°C	60	93	
k_{SVR}	Supply-voltage rejection ratio ($\Delta V_{DD}/\Delta V_{IO}$)	$V_{DD} = 5\text{ V to }10\text{ V},$	$V_O = 1.4\text{ V}$	25°C	70	93	dB
				-55°C	60	91	
				125°C	60	94	
I_{DD}	Supply current (four amplifiers)	$V_O = 5\text{ V},$ No load	$V_{IC} = 5\text{ V},$	25°C	570	1200	μA
				-55°C	980	2000	
				125°C	360	960	

(1) Full range is -55°C to 70°C.

(2) The typical values of input bias current and input offset current below 5 pA were determined mathematically.

(3) This range also applies to each input individually.

ELECTRICAL CHARACTERISTICS

V_{DD} = 5 V, T_A = 25°C (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC27M4Y			UNIT
		MIN	TYP	MAX	
V _{IO}	Input offset voltage V _O = 1.4 V, R _S = 50 Ω, V _{IC} = 0, R _L = 100 kΩ		1.1	10	mV
α _{VIO}	Temperature coefficient of input offset voltage T _A = 25°C to 70°C		1.7		μV/°C
I _{IO}	Input offset current ⁽¹⁾ V _O = 2.5 V, V _{IC} = 2.5 V		0.1		pA
I _{IB}	Input bias current ⁽¹⁾ V _O = 2.5 V, V _{IC} = 2.5 V		0.6		pA
V _{ICR}	Common-mode input voltage range ⁽²⁾	-0.2 to 4	-0.3 to 4.2		V
V _{OH}	High-level output voltage V _{ID} = 100 mV, R _L = 100 kΩ	3.2	3.9		V
V _{OL}	Low-level output voltage V _{ID} = -100 mV, I _{OL} = 0		0	50	mV
A _{VD}	Large-signal differential voltage amplification V _O = 0.25 V to 2 V, R _L = 100 kΩ	25	170		V/mV
CMRR	Common-mode rejection ratio V _{IC} = V _{ICRmin}	65	91		dB
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO}) V _{DD} = 5 V to 10 V, V _O = 1.4 V	70	93		dB
I _{DD}	Supply current (four amplifiers) V _O = 2.5 V, No load, V _{IC} = 2.5 V		420	1120	μA

(1) The typical values of input bias current and input offset current below 5 pA were determined mathematically.

(2) This range also applies to each input individually.

ELECTRICAL CHARACTERISTICS

V_{DD} = 10 V, T_A = 25°C (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TLC27M4Y			UNIT
		MIN	TYP	MAX	
V _{IO}	Input offset voltage V _O = 1.4 V, R _S = 50 Ω, V _{IC} = 0, R _L = 100 kΩ		1.1	10	mV
α _{VIO}	Temperature coefficient of input offset voltage T _A = 25°C to 70°C		1.7		μV/°C
I _{IO}	Input offset current ⁽¹⁾ V _O = 5 V, V _{IC} = 5 V		0.1		pA
I _{IB}	Input bias current ⁽¹⁾ V _O = 5 V, V _{IC} = 5 V		0.6		pA
V _{ICR}	Common-mode input voltage range ⁽²⁾	-0.2 to 9	-0.3 to 9.2		V
V _{OH}	High-level output voltage V _{ID} = 100 mV, R _L = 100 kΩ	8	8.7		V
V _{OL}	Low-level output voltage V _{ID} = -100 mV, I _{OL} = 0		0	50	mV
A _{VD}	Large-signal differential voltage amplification V _O = 1 V to 6 V, R _L = 100 kΩ	25	275		V/mV
CMRR	Common-mode rejection ratio V _{IC} = V _{ICRmin}	65	94		dB
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO}) V _{DD} = 5 V to 10 V, V _O = 1.4 V	70	93		dB
I _{DD}	Supply current (four amplifiers) V _O = 5 V, No load, V _{IC} = 5 V		570	1200	μA

(1) The typical values of input bias current and input offset current below 5 pA were determined mathematically.

(2) This range also applies to each input individually.

OPERATING CHARACTERISTICS

 at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	TLC27M4C TLC27M4AC TLC27M4BC TLC27M9C			UNIT
				MIN	TYP	MAX	
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C	0.43		V/ μ s
				0°C	0.46		
				70°C	0.36		
			$V_{IPP} = 2.5\text{ V}$	25°C	0.40		
				0°C	0.43		
				70°C	0.34		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C	55		kHz
				0°C	60		
				70°C	50		
B_1	Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	25°C	525		kHz
				0°C	610		
				70°C	400		
Φ_m	Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C	40°		
				0°C	41°		
				70°C	39°		

OPERATING CHARACTERISTICS

 at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	TLC27M4C TLC27M4AC TLC27M4BC TLC27M9C			UNIT
				MIN	TYP	MAX	
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C	0.62		V/ μ s
				0°C	0.67		
				70°C	0.51		
			$V_{IPP} = 5.5\text{ V}$	25°C	0.56		
				0°C	0.61		
				70°C	0.46		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C	35		kHz
				0°C	40		
				70°C	30		
B_1	Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	25°C	635		kHz
				0°C	710		
				70°C	510		
Φ_m	Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C	43°		
				0°C	44°		
				70°C	42°		

OPERATING CHARACTERISTICS

at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	TLC27M4I TLC27M4AI TLC27M4BI TLC27M9I			UNIT
				MIN	TYP	MAX	
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C	0.43		V/ μ s
				-40°C	0.51		
				85°C	0.35		
			$V_{IPP} = 2.5\text{ V}$	25°C	0.40		
				-40°C	0.48		
				85°C	0.32		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C	55		kHz
				-40°C	75		
				85°C	45		
B_1	Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	25°C	525		kHz
				-40°C	770		
				85°C	370		
Φ_m	Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C	40°		
				-40°C	43°		
				85°C	38°		

OPERATING CHARACTERISTICS

at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	TLC27M4I TLC27M4AI TLC27M4BI TLC27M9I			UNIT
				MIN	TYP	MAX	
SR	Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C	0.62		V/ μ s
				-40°C	0.77		
				85°C	0.47		
			$V_{IPP} = 5.5\text{ V}$	25°C	0.56		
				-40°C	0.70		
				85°C	0.44		
V_n	Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C	35		kHz
				-40°C	45		
				85°C	25		
B_1	Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	25°C	635		kHz
				-40°C	880		
				85°C	480		
Φ_m	Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C	43°		
				-40°C	46°		
				85°C	41°		

OPERATING CHARACTERISTICS

 at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC27M4M TLC27M9M			UNIT
			MIN	TYP	MAX	
SR	Slew rate at unity gain $R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C	0.43		V/ μ s
			-55°C	0.54		
			125°C	0.29		
		$V_{IPP} = 2.5\text{ V}$	25°C	0.40		
			-55°C	0.50		
			125°C	0.28		
V_n	Equivalent input noise voltage $f = 1\text{ kHz}$, See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output-swing bandwidth $V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C	56		kHz
			-55°C	80		
			125°C	40		
B_1	Unity-gain bandwidth $V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	25°C	525		kHz
			-55°C	850		
			125°C	330		
Φ_m	Phase margin $V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C	40°		
			-55°C	44°		
			125°C	36°		

OPERATING CHARACTERISTICS

 at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC27M4M TLC27M9M			UNIT
			MIN	TYP	MAX	
SR	Slew rate at unity gain $R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	25°C	0.62		V/ μ s
			-55°C	0.81		
			125°C	0.38		
		$V_{IPP} = 5.5\text{ V}$	25°C	0.56		
			-55°C	0.73		
			125°C	0.35		
V_n	Equivalent input noise voltage $f = 1\text{ kHz}$, See Figure 2	$R_S = 20\ \Omega$	25°C	32		nV/ $\sqrt{\text{Hz}}$
B_{OM}	Maximum output-swing bandwidth $V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	25°C	35		kHz
			-55°C	50		
			125°C	20		
B_1	Unity-gain bandwidth $V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	25°C	635		kHz
			-55°C	960		
			125°C	440		
Φ_m	Phase margin $V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	25°C	43°		
			-55°C	47°		
			125°C	39°		

OPERATING CHARACTERISTICS

at specified free-air temperature, $V_{DD} = 5\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC27M4Y			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	0.43		V/ μs
		$V_{IPP} = 2.5\text{ V}$	0.40		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 20\ \Omega$	32		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	55		kHz
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	525		kHz
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	40°		

OPERATING CHARACTERISTICS

at specified free-air temperature, $V_{DD} = 10\text{ V}$, $T_A = 25^\circ\text{C}$

PARAMETER	TEST CONDITIONS	TLC27M4Y			UNIT
		MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 1	$V_{IPP} = 1\text{ V}$	0.62		V/ μs
		$V_{IPP} = 5.5\text{ V}$	0.56		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 2	$R_S = 20\ \Omega$	32		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 1	35		kHz
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 3	$C_L = 20\text{ pF}$	635		kHz
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 3	43°		

PARAMETER MEASUREMENT INFORMATION

Single-Supply versus Split-Supply Test Circuits

Because the TLC27M4 and TLC27M9 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

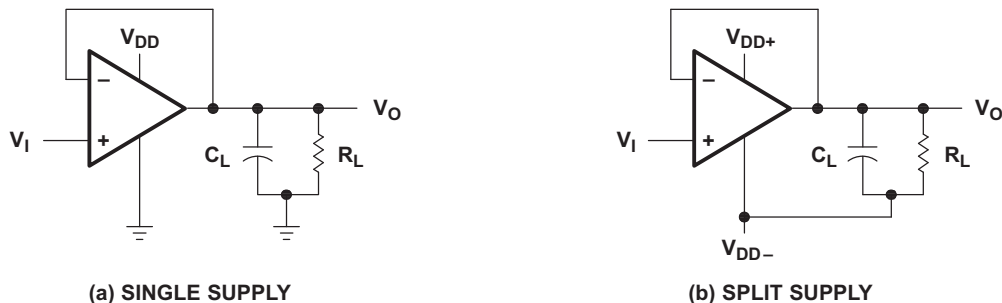


Figure 1. Unity-Gain Amplifier

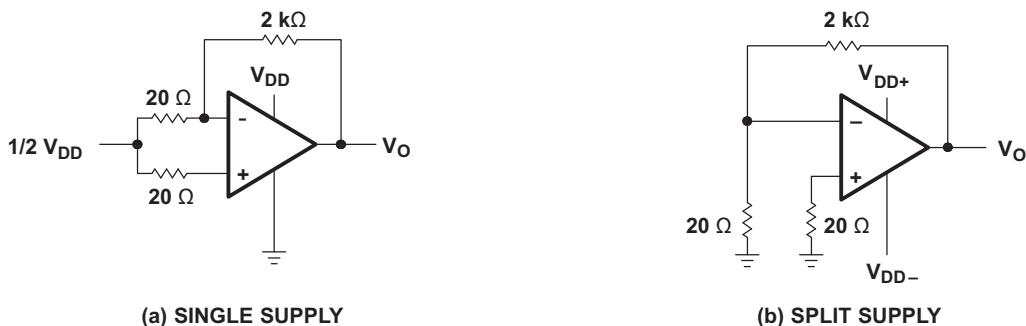


Figure 2. Noise-Test Circuit

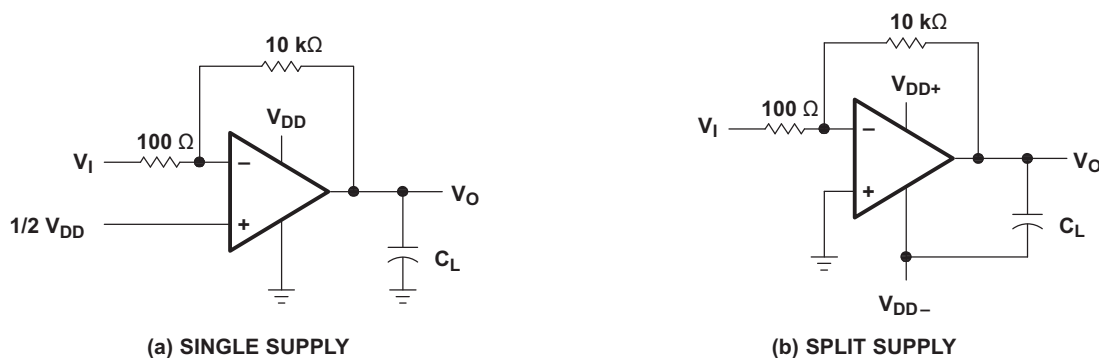


Figure 3. Gain-of-100 Inverting Amplifier

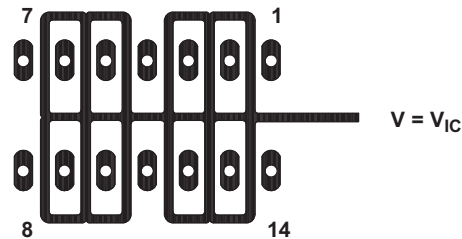
PARAMETER MEASUREMENT INFORMATION (continued)

Input Bias Current

Because of the high input impedance of the TLC27M4 and TLC27M9 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see [Figure 4](#)). Leakages that would otherwise flow to the inputs are shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution—many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current; the voltage drop across the series resistor is measured and the bias current is calculated. This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.



**Figure 4. Isolation Metal Around Device Inputs
(J and N packages)**

Low-Level Output Voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to [Figure 14](#) through [Figure 19](#) in the *Typical Characteristics* of this data sheet.

Input Offset Voltage Temperature Coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance, which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

PARAMETER MEASUREMENT INFORMATION (continued)

Full-Power Response

Full-power response, the frequency above which the operational amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output, while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of [Figure 1](#). The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained ([Figure 5](#)). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

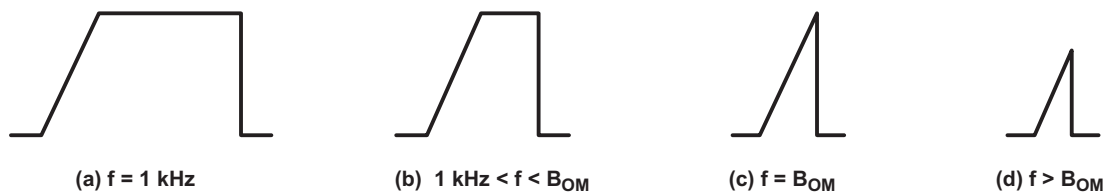


Figure 5. Full-Power-Response Output Signal

Test Time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
V_{IO}	Input offset voltage	Distribution	6, 7
α_{VIO}	Temperature coefficient of input offset voltage	Distribution	8, 9
V_{OH}	High-level output voltage	vs High-level output current vs Supply voltage vs Free-air temperature	10, 11 12 13
V_{OL}	Low-level output voltage	vs Common-mode input voltage vs Differential input voltage vs Free-air temperature vs Low-level output current	14, 15 16 17 18, 19
A_{VD}	Differential voltage amplification	vs Supply voltage vs Free-air temperature Free vs Frequency	20 21 32, 33
I_{IB}	Input bias current	vs Free-air temperature	22
I_{IO}	Input offset current	vs Free-air temperature	22
V_{IC}	Common-mode input voltage	vs Supply voltage	23
I_{DD}	Supply current	vs Supply voltage vs Free-air temperature	24 25
SR	Slew rate	vs Supply voltage vs Free-air temperature	26 27
	Normalized slew rate	vs Free-air temperature	28
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	29
B_1	Unity gain bandwidth	vs Free-air temperature Free vs Supply voltage	30 31
	Phase shift	vs Frequency	32, 33
ϕ_m	Phase margin	vs Supply voltage	34
		vs Free-air temperature Free	35
		vs Load capacitance	36
V_n	Equivalent input noise voltage	vs Frequency	37

TYPICAL CHARACTERISTICS

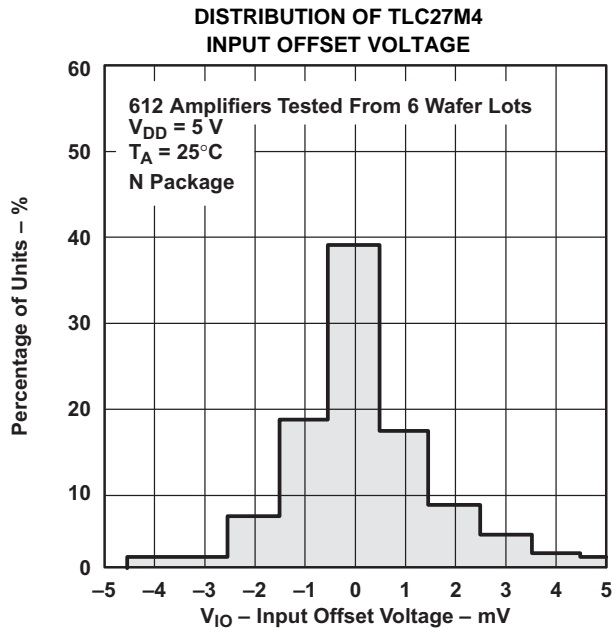


Figure 6.

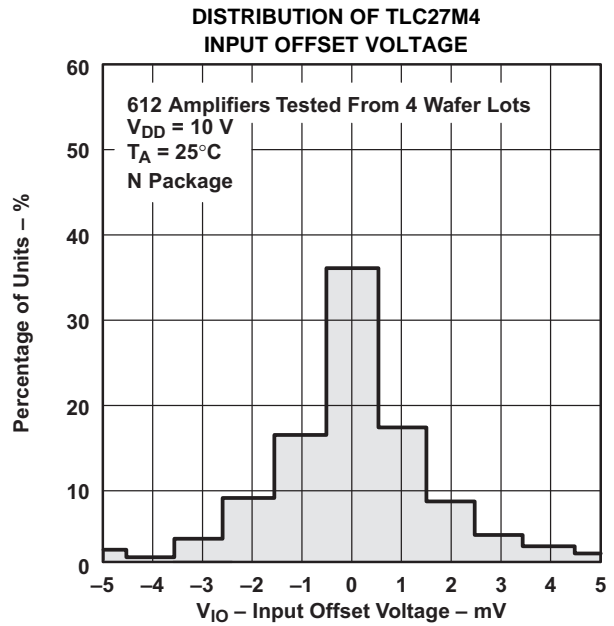


Figure 7.

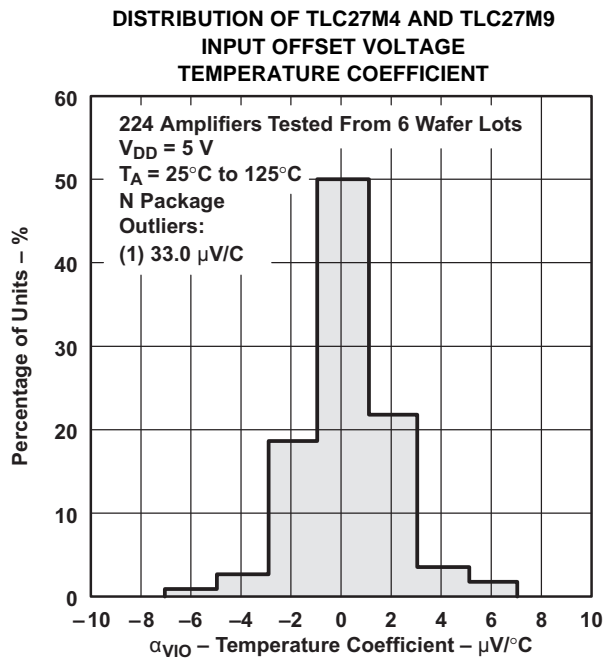


Figure 8.

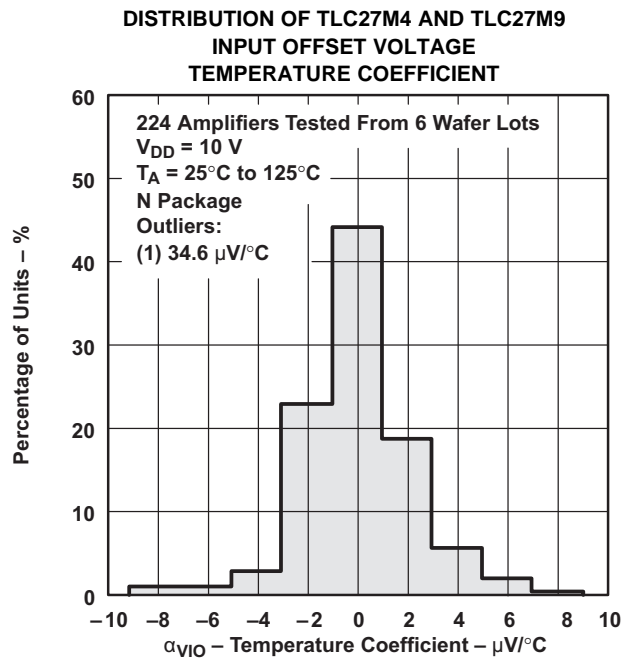


Figure 9.

TYPICAL CHARACTERISTICS (1)

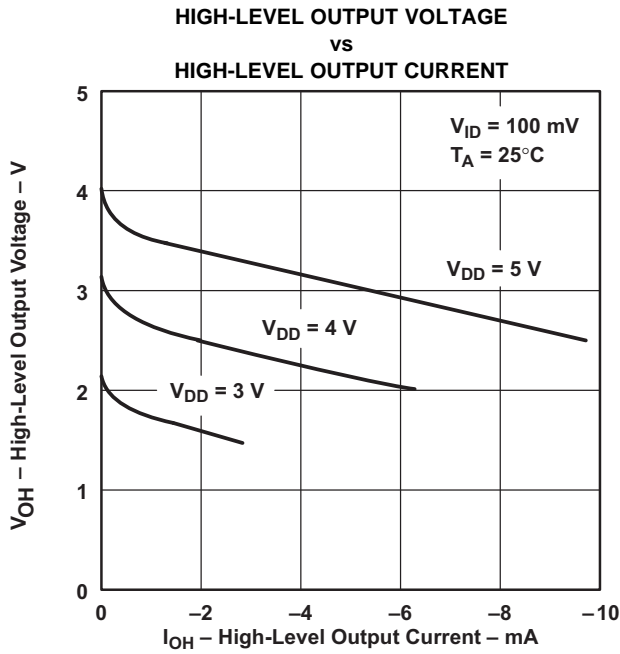


Figure 10.

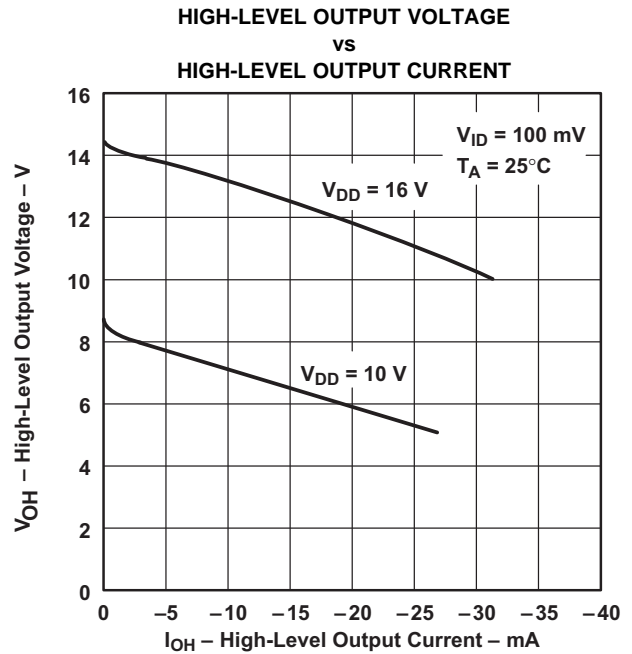


Figure 11.

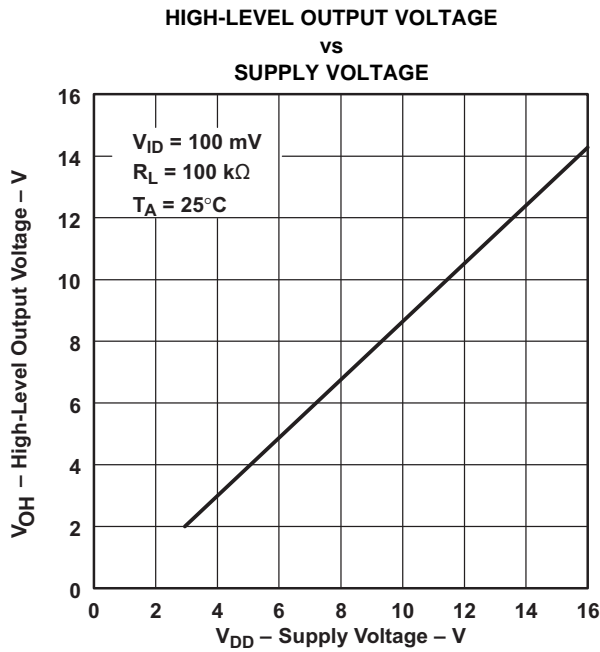


Figure 12.

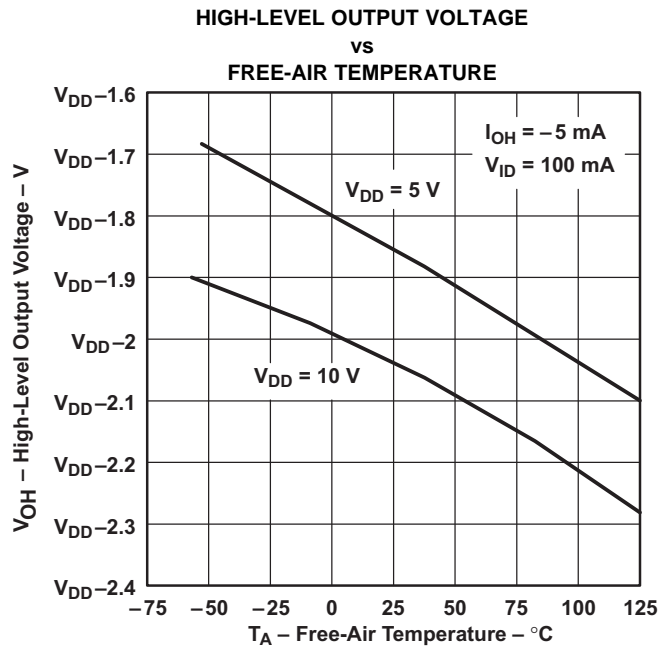


Figure 13.

(1) Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS ⁽¹⁾

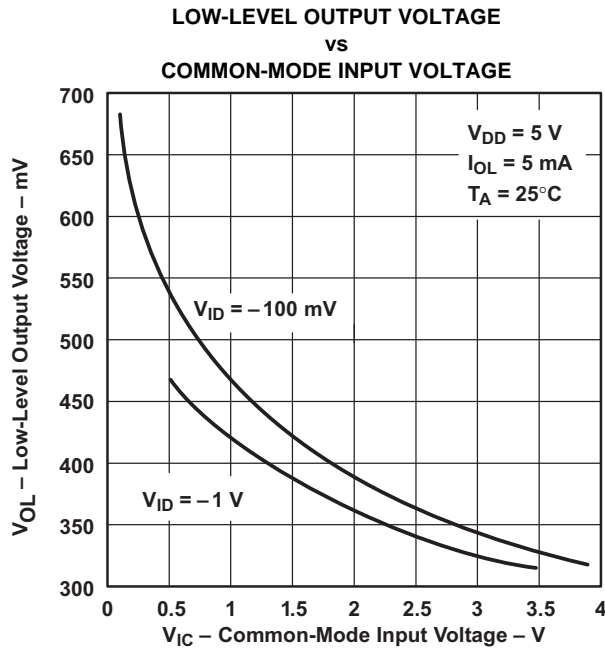


Figure 14.

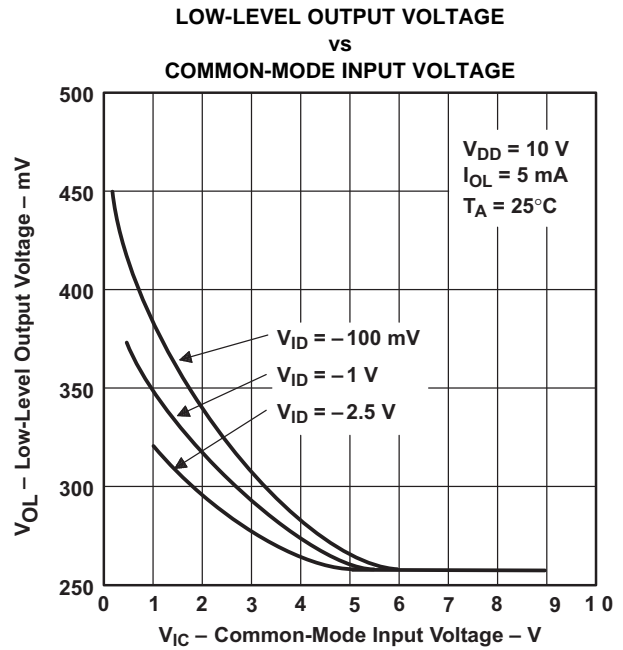


Figure 15.

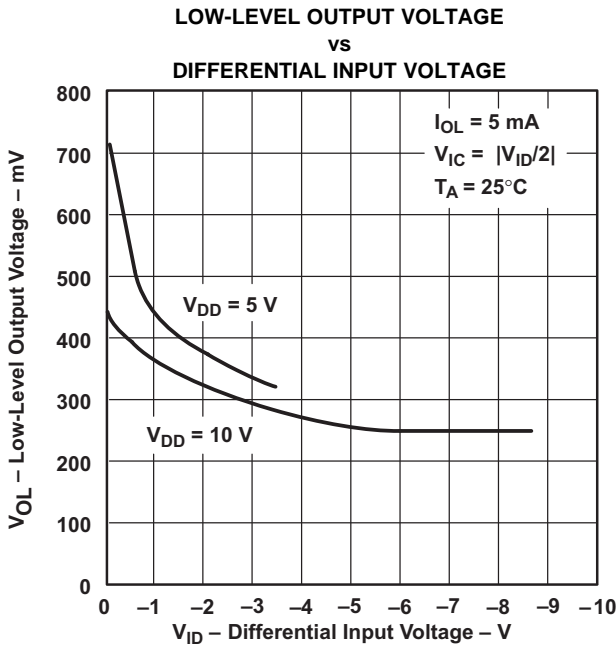


Figure 16.

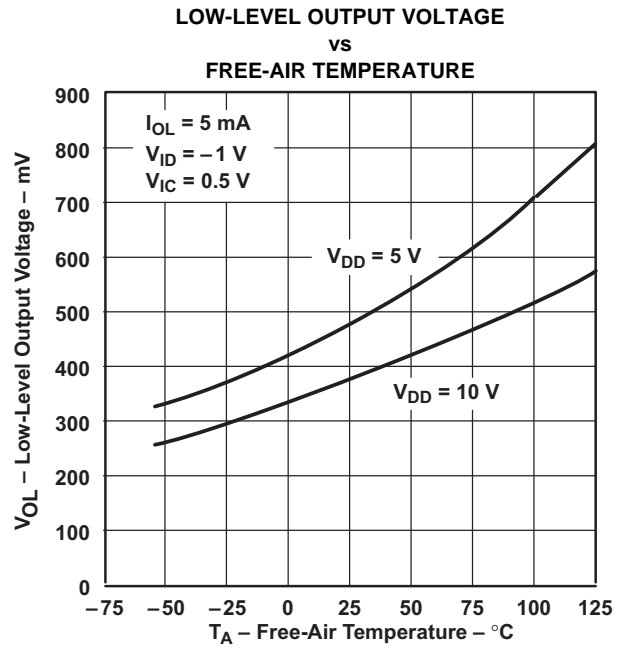


Figure 17.

(1) Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (1)

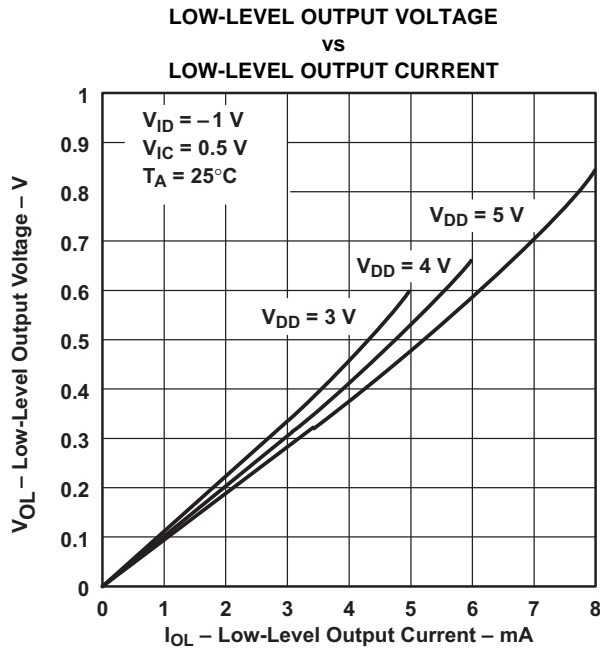


Figure 18.

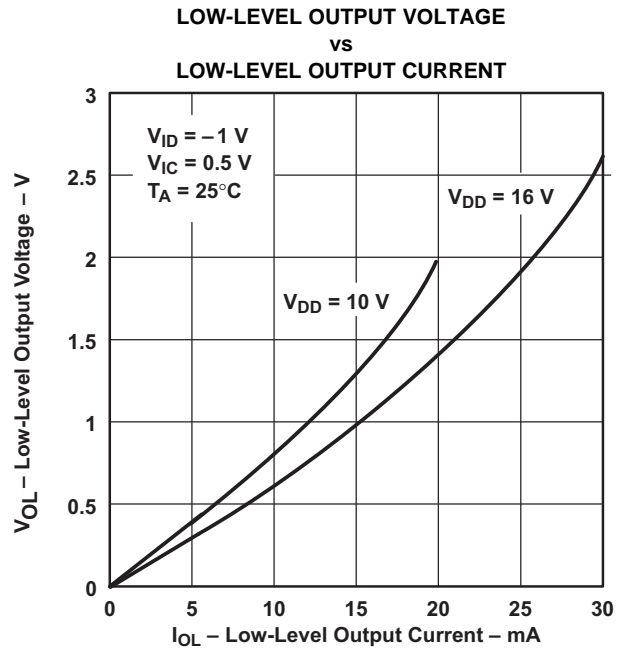


Figure 19.

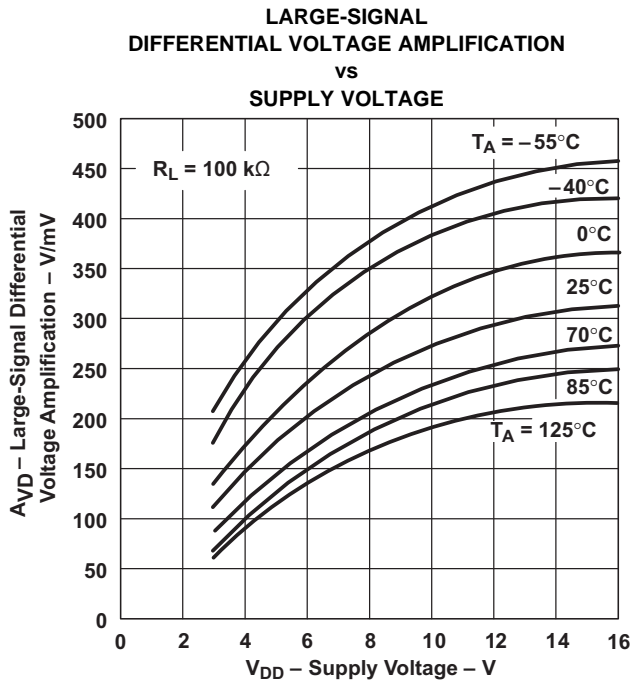


Figure 20.

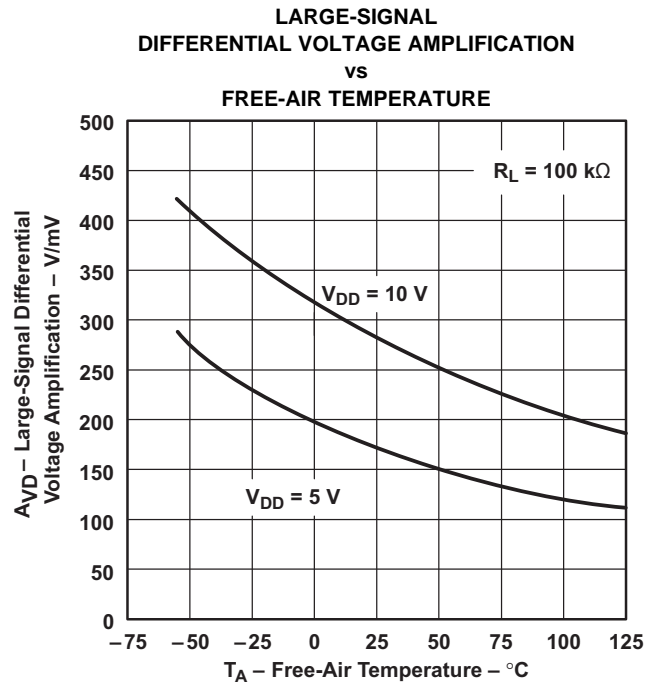
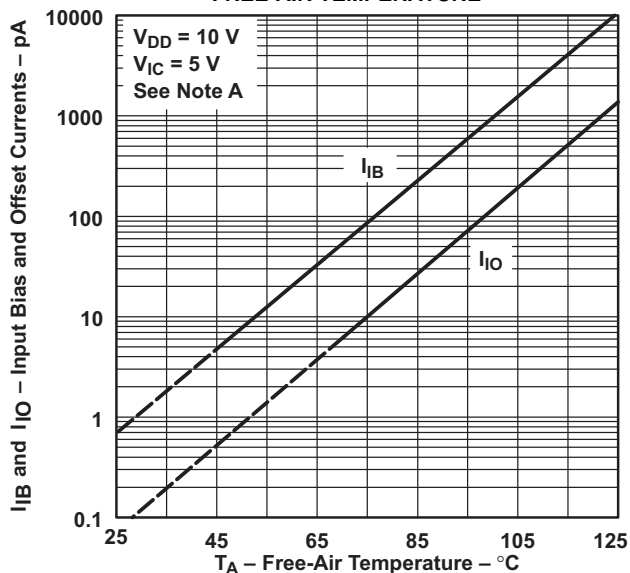


Figure 21.

(1) Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS ⁽¹⁾

INPUT BIAS CURRENT AND INPUT OFFSET CURRENT vs FREE-AIR TEMPERATURE



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 22.

COMMON-MODE INPUT VOLTAGE POSITIVE LIMIT vs SUPPLY VOLTAGE

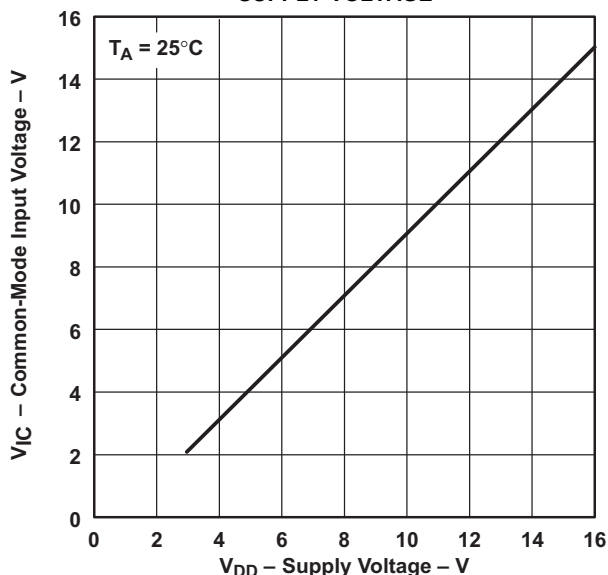


Figure 23.

SUPPLY CURRENT vs SUPPLY VOLTAGE

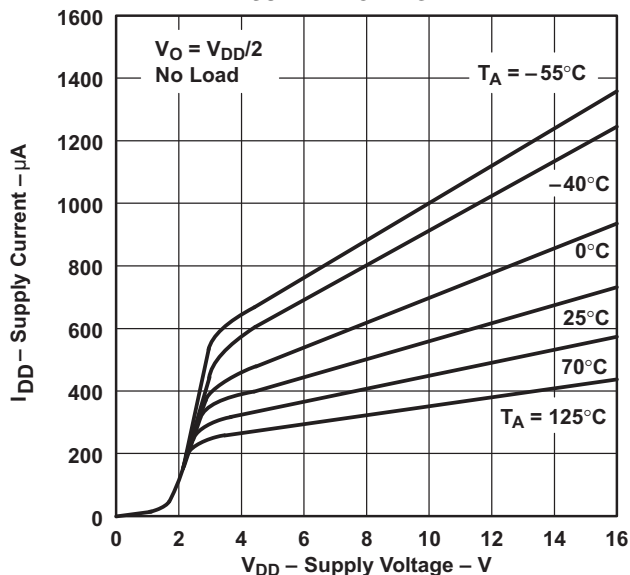


Figure 24.

SUPPLY CURRENT vs FREE-AIR TEMPERATURE

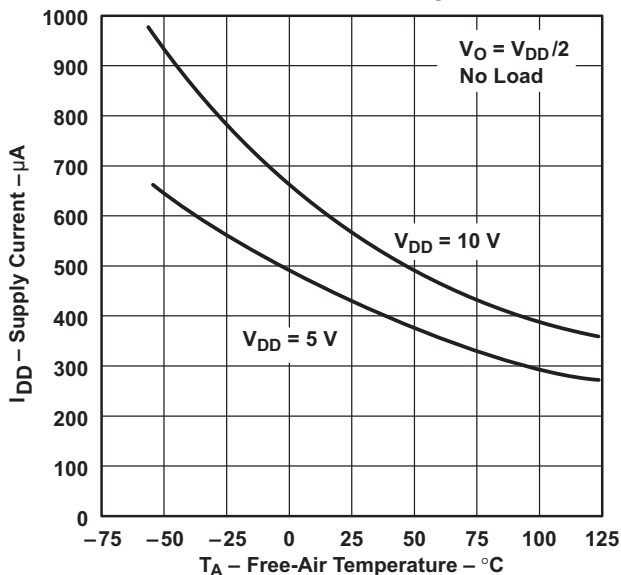


Figure 25.

(1) Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (1)

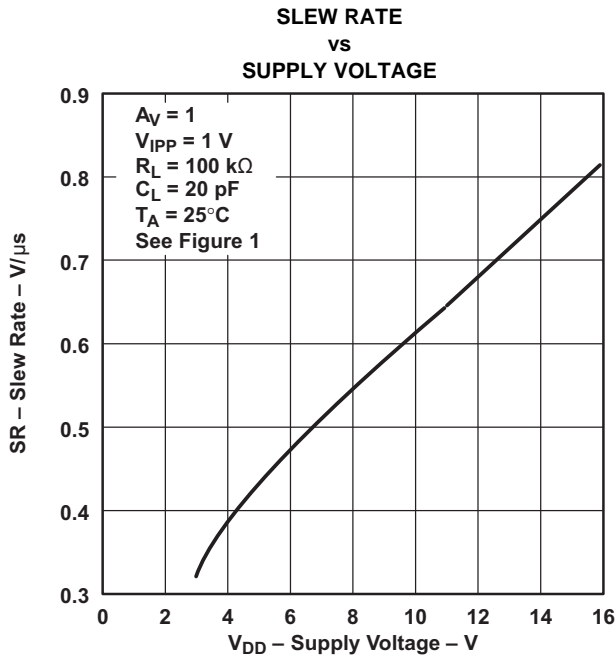


Figure 26.

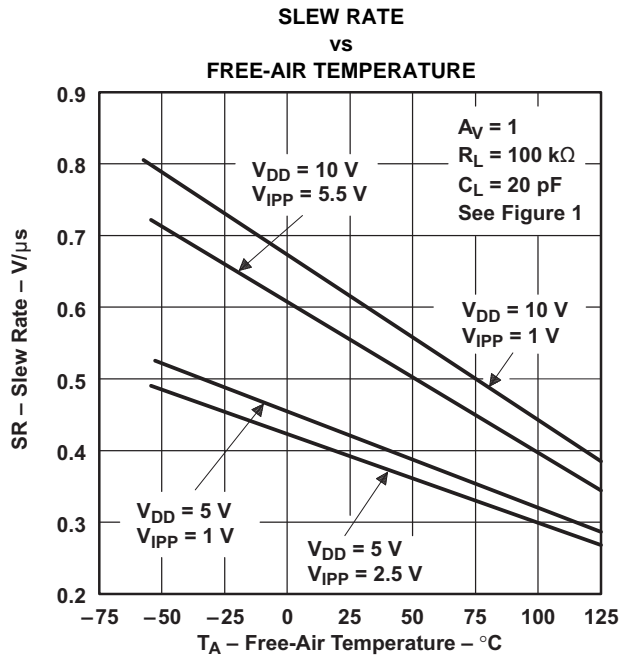


Figure 27.

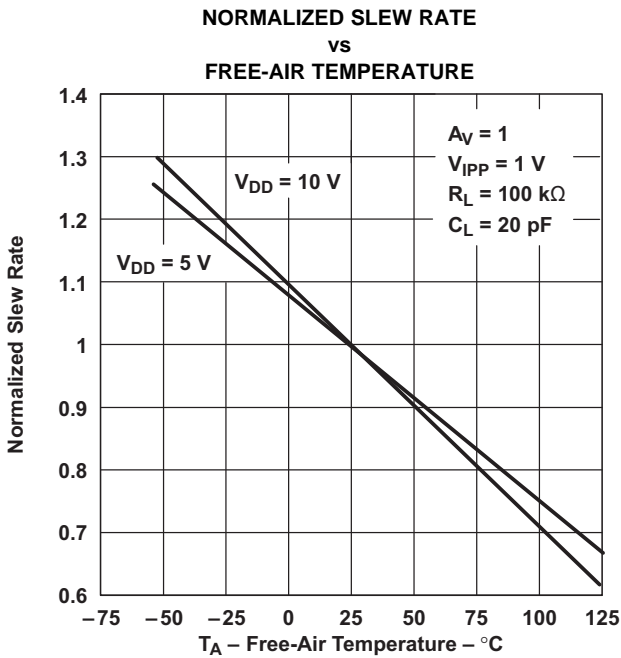


Figure 28.

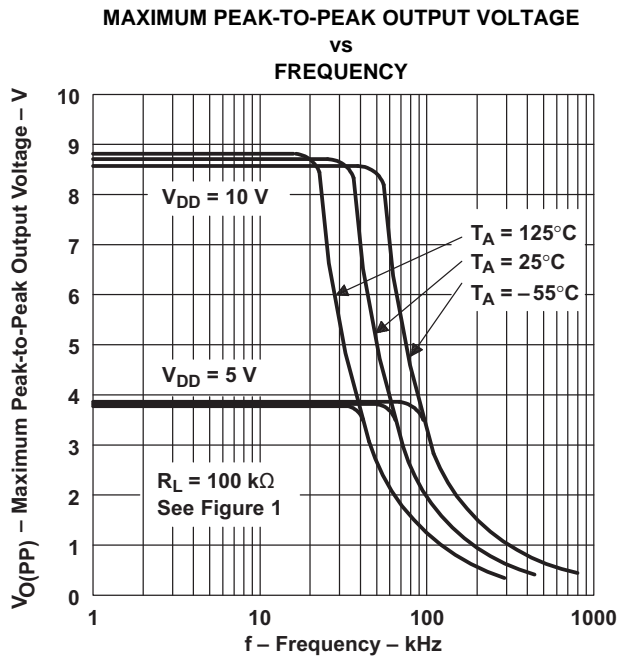


Figure 29.

(1) Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS ⁽¹⁾

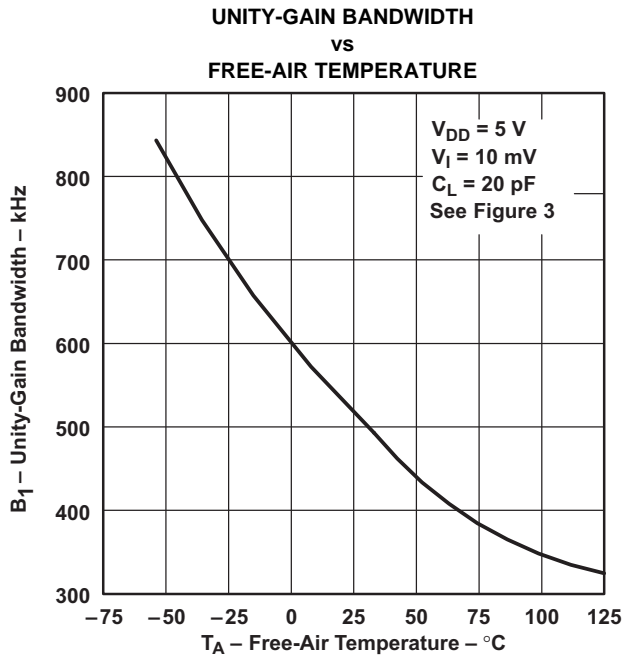


Figure 30.

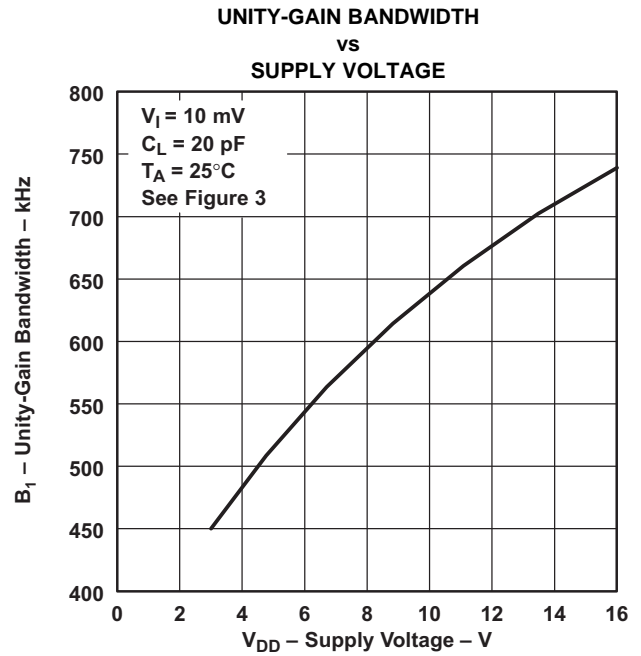


Figure 31.

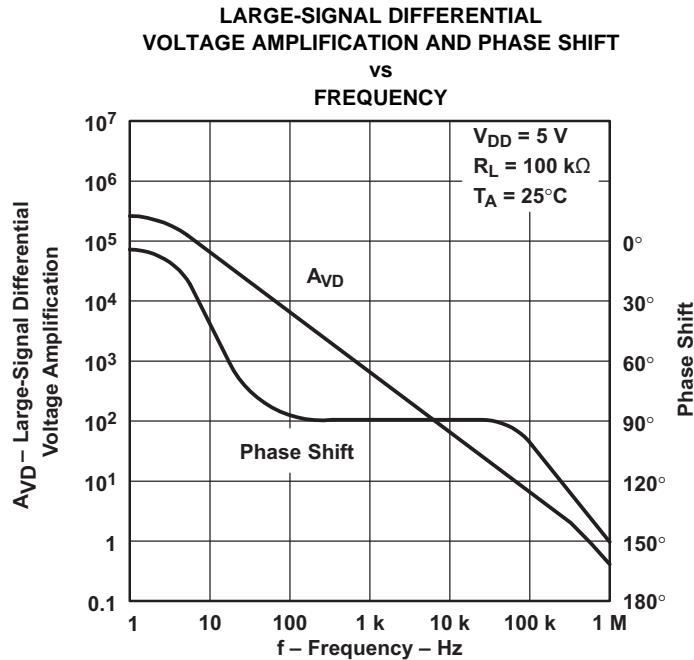


Figure 32.

(1) Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (1)

LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT

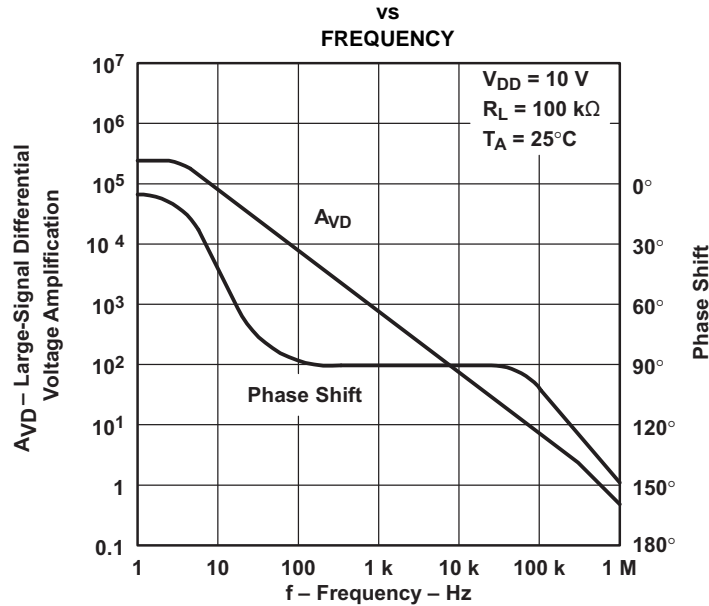


Figure 33.

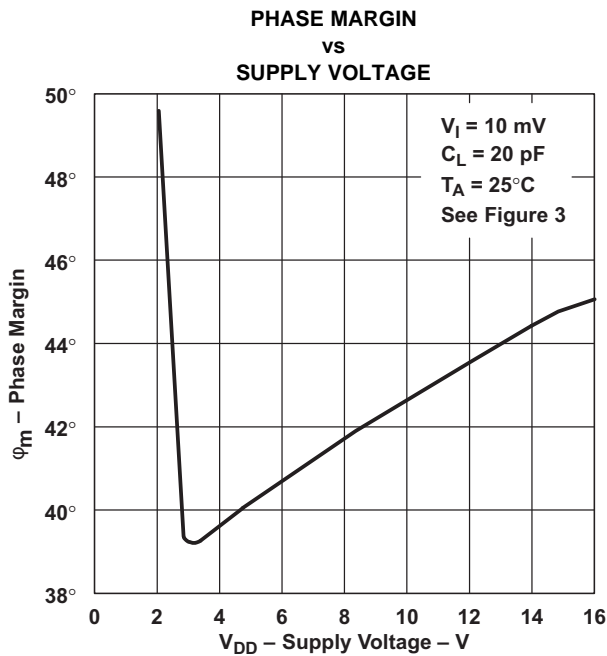


Figure 34.

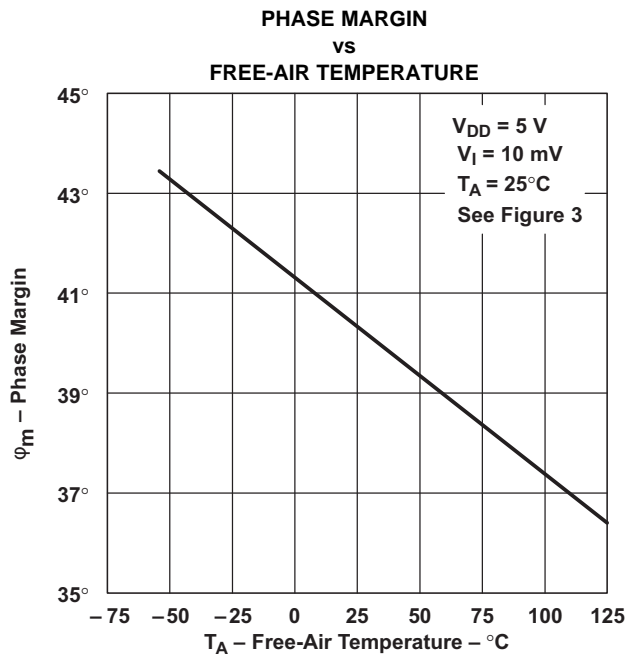


Figure 35.

(1) Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS

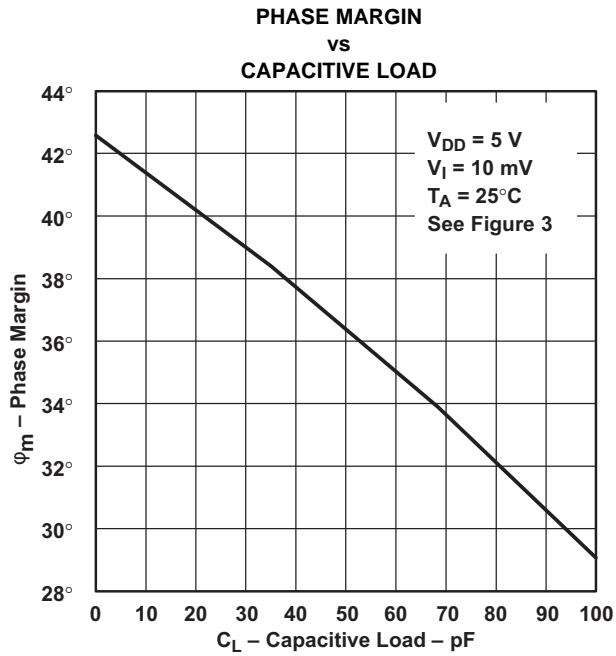


Figure 36.

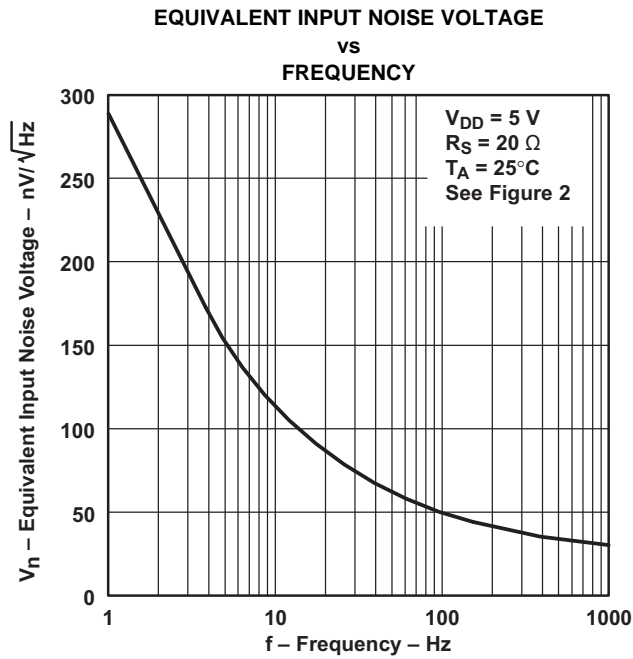


Figure 37.

APPLICATION INFORMATION

Single-Supply Operation

While the TLC27M4 and TLC27M9 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common-mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC27M4 and TLC27M9 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC27M4 and TLC27M9 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

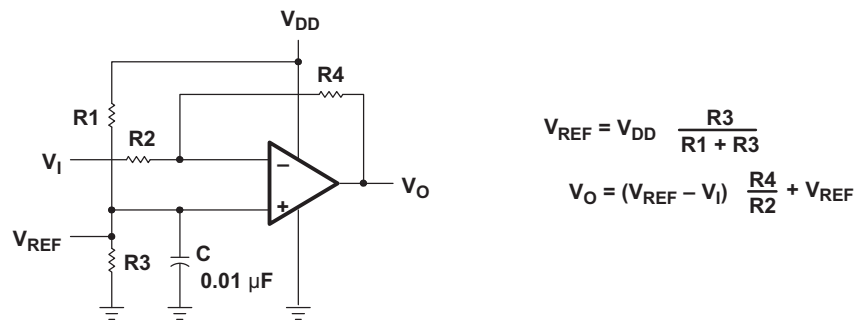


Figure 38. Inverting Amplifier With Voltage Reference

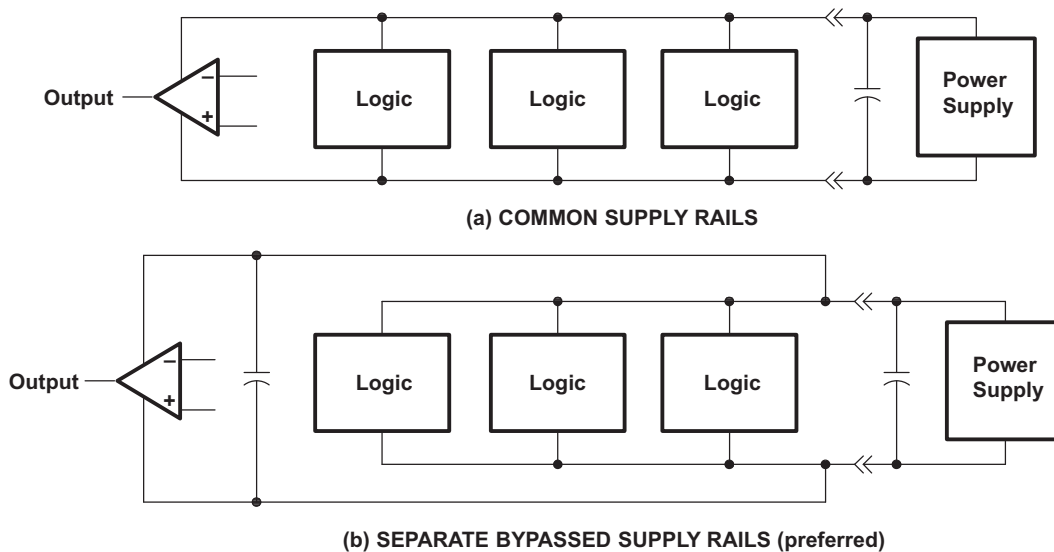


Figure 39. Common Versus Separate Supply Rails

Input Characteristics

The TLC27M4 and TLC27M9 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at $V_{DD} - 1$ V at $T_A = 25^\circ\text{C}$ and at $V_{DD} - 1.5$ V at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC27M4 and TLC27M9 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically $0.1 \mu\text{V}/\text{month}$, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC27M4 and TLC27M9 are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the *Parameter Measurement Information* section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

Unused amplifiers should be connected as unity-gain followers to avoid possible oscillation.

Noise Performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC27M4 and TLC27M9 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than $50 \text{ k}\Omega$, since bipolar devices exhibit greater noise currents.

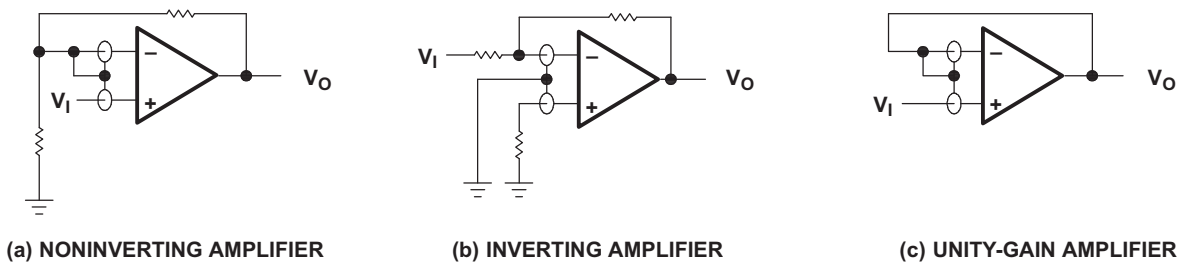


Figure 40. Guard-Ring Schemes

Output Characteristics

The output stage of the TLC27M4 and TLC27M9 is designed to sink and source relatively high amounts of current (see *typical characteristics*). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC27M4 and TLC27M9 were measured using a 20-pF load. The devices drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see [Figure 41](#)). In many cases, adding a small amount of resistance in series with the load capacitance alleviates the problem.

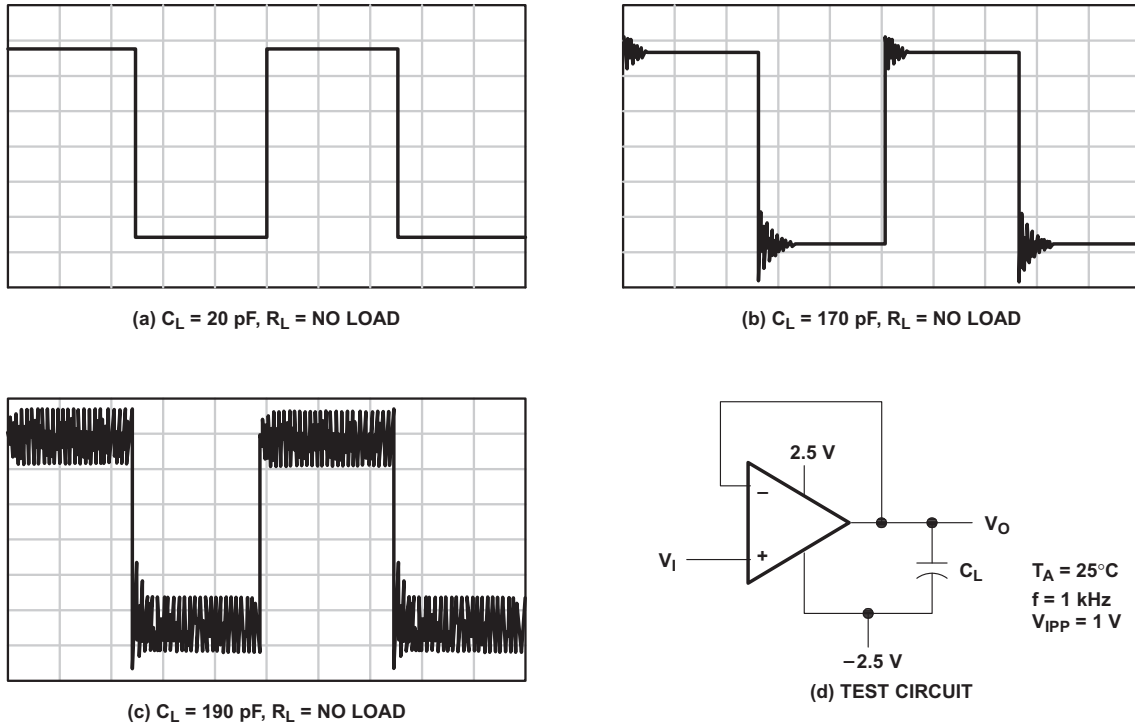
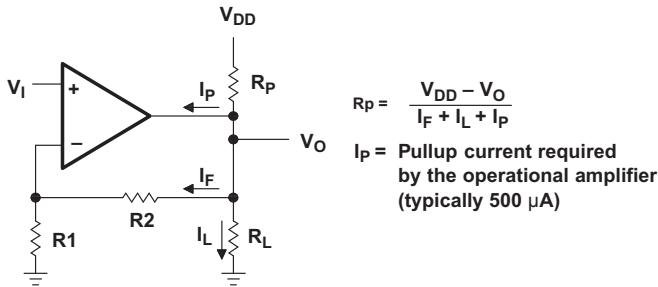
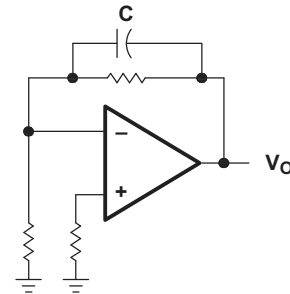


Figure 41. Effect of Capacitive Loads and Test Circuit

Although the TLC27M4 and TLC27M9 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor (RP) connected from the output to the positive supply rail (see [Figure 42](#)). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately 60 Ω and 180 Ω, depending on how hard the operational amplifier input is driven. With very low values of RP, a voltage offset from 0 V at the output occurs. Second, pullup resistor RP acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.


Figure 42. Resistive Pullup to Increase V_{OH}

Figure 43. Compensation for Input Capacitance

Feedback

Operational amplifier circuits nearly always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see [Figure 43](#)). The value of this capacitor is optimized empirically.

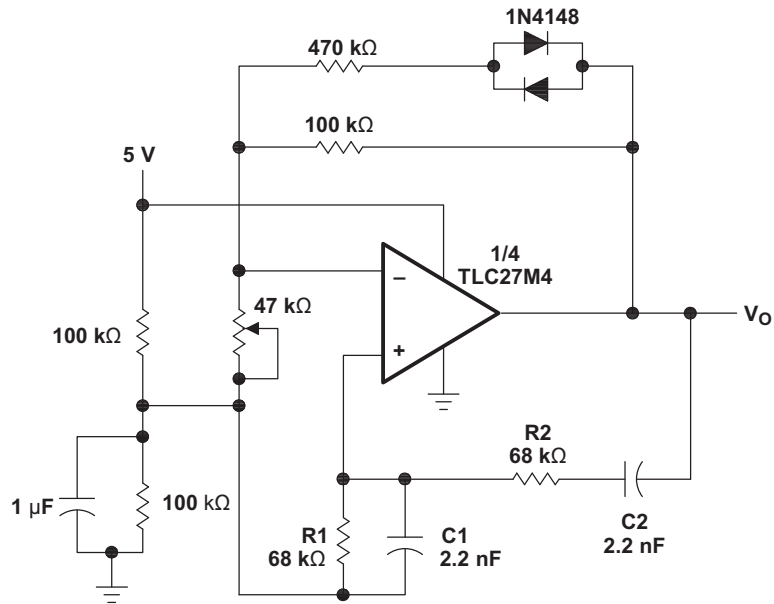
Electrostatic Discharge Protection

The TLC27M4 and TLC27M9 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices, as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature-dependent and have the characteristics of a reverse-biased diode.

Latch-Up

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC27M4 and TLC27M9 inputs and outputs were designed to withstand —100-mA surge currents without sustaining latch-up; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μ F typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the positive supply rail and ground; it can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.



NOTE: $V_{OPP} \approx 2\text{ V}$

$$f_o = \frac{1}{2\pi \sqrt{R_1 R_2 C_1 C_2}}$$

Figure 44. Wien Oscillator

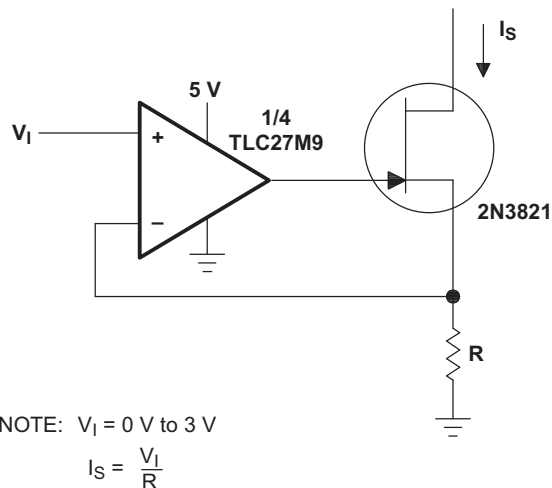


Figure 45. Precision Low-Current Sink

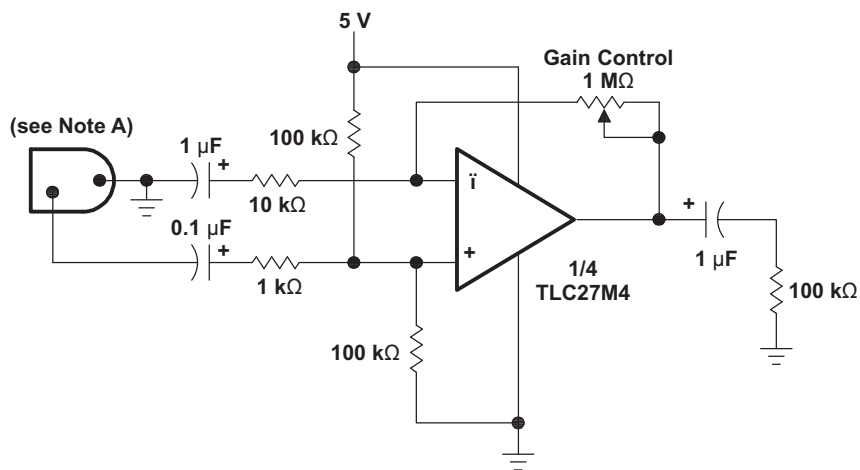


Figure 46. Microphone Preamplifier

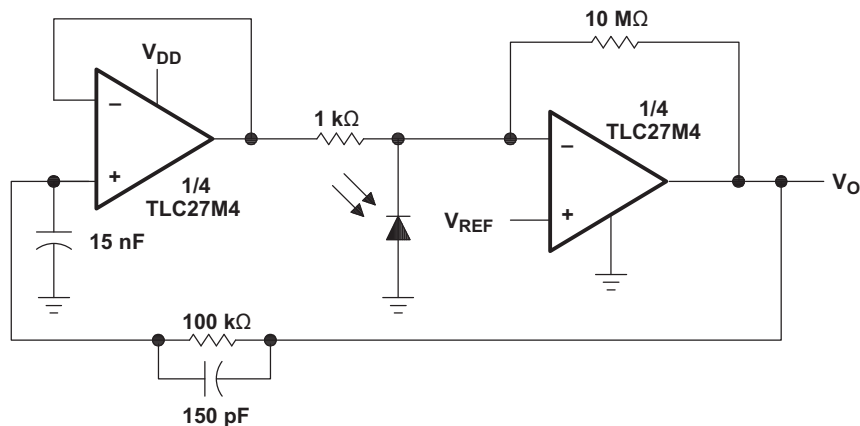
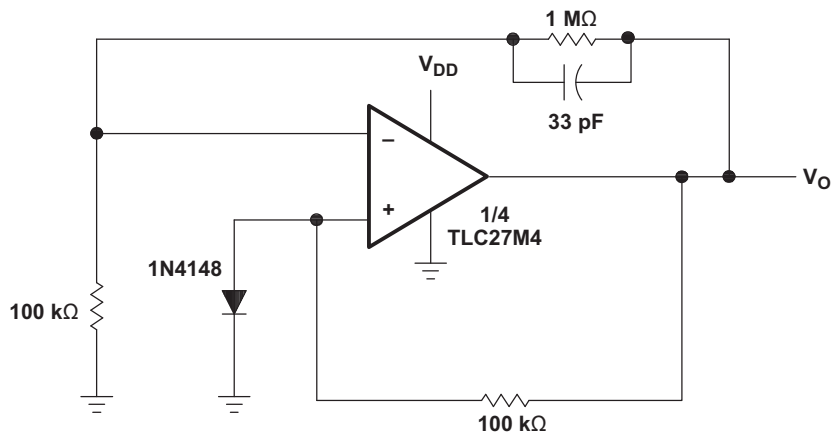


Figure 47. Photo-Diode Amplifier With Ambient Light Rejection



NOTE: $V_{DD} = 8\text{ V to }16\text{ V}$
 $V_O = 5\text{ V, }10\text{ mA}$

Figure 48. Low-Power Voltage Regulator

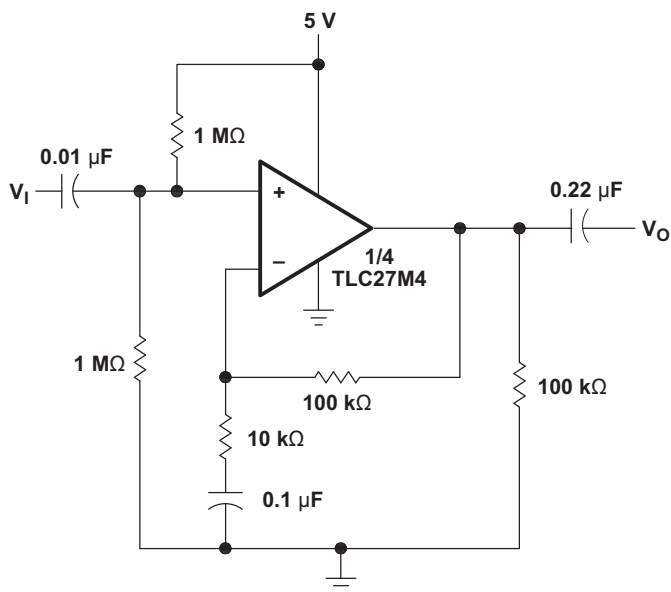


Figure 49. Single-Rail AC Amplifier

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TLC27M4ACD	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	0 to 70	27M4AC	
TLC27M4ACDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M4AC	Samples
TLC27M4ACN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC27M4ACN	Samples
TLC27M4ACNE4	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC27M4ACN	Samples
TLC27M4AID	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	-40 to 85	27M4AI	
TLC27M4AIDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M4AI	Samples
TLC27M4AIN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TLC27M4AIN	Samples
TLC27M4BCD	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	0 to 70	27M4BC	
TLC27M4BCDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	27M4BC	Samples
TLC27M4BCN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC27M4BCN	Samples
TLC27M4BID	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	-40 to 85	27M4BI	
TLC27M4BIDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	27M4BI	Samples
TLC27M4BIN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TLC27M4BIN	Samples
TLC27M4CD	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	0 to 70	TLC27M4C	
TLC27M4CDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC27M4C	Samples
TLC27M4CDRG4	ACTIVE	SOIC	D	14	2500	TBD	Call TI	Call TI	0 to 70		Samples
TLC27M4CN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC27M4CN	Samples
TLC27M4CNS	ACTIVE	SOP	NS	14	50	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC27M4	Samples
TLC27M4CNSR	ACTIVE	SOP	NS	14	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC27M4	Samples
TLC27M4CPWR	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	P27M4	Samples
TLC27M4ID	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	-40 to 85	TLC27M4I	
TLC27M4IDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TLC27M4I	Samples

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TLC27M4IDRG4	ACTIVE	SOIC	D	14	2500	TBD	Call TI	Call TI	-40 to 85		Samples
TLC27M4IN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TLC27M4IN	Samples
TLC27M4IPW	OBSOLETE	TSSOP	PW	14		TBD	Call TI	Call TI	-40 to 85	P27M4I	
TLC27M4IPWR	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	P27M4I	Samples
TLC27M9CD	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	0 to 70	TLC27M9C	
TLC27M9CDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC27M9C	Samples
TLC27M9CDRG4	ACTIVE	SOIC	D	14	2500	TBD	Call TI	Call TI	0 to 70		Samples
TLC27M9CN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC27M9CN	Samples
TLC27M9ID	OBSOLETE	SOIC	D	14		TBD	Call TI	Call TI	-40 to 85	TLC27M9I	
TLC27M9IDR	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TLC27M9I	Samples
TLC27M9IN	ACTIVE	PDIP	N	14	25	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TLC27M9IN	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC27M4ACDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC27M4ACDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC27M4AIDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC27M4BCDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC27M4BIDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC27M4CDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC27M4CDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC27M4CNSR	SOP	NS	14	2000	330.0	16.4	8.2	10.5	2.5	12.0	16.0	Q1
TLC27M4CPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TLC27M4CPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TLC27M4IDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC27M4IDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC27M4IPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TLC27M4IPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TLC27M9CDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC27M9IDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC27M4ACDR	SOIC	D	14	2500	353.0	353.0	32.0
TLC27M4ACDR	SOIC	D	14	2500	350.0	350.0	43.0
TLC27M4AIDR	SOIC	D	14	2500	353.0	353.0	32.0
TLC27M4BCDR	SOIC	D	14	2500	350.0	350.0	43.0
TLC27M4BIDR	SOIC	D	14	2500	350.0	350.0	43.0
TLC27M4CDR	SOIC	D	14	2500	350.0	350.0	43.0
TLC27M4CDR	SOIC	D	14	2500	353.0	353.0	32.0
TLC27M4CNSR	SOP	NS	14	2000	356.0	356.0	35.0
TLC27M4CPWR	TSSOP	PW	14	2000	356.0	356.0	35.0
TLC27M4CPWR	TSSOP	PW	14	2000	356.0	356.0	35.0
TLC27M4IDR	SOIC	D	14	2500	350.0	350.0	43.0
TLC27M4IDR	SOIC	D	14	2500	353.0	353.0	32.0
TLC27M4IPWR	TSSOP	PW	14	2000	356.0	356.0	35.0
TLC27M4IPWR	TSSOP	PW	14	2000	356.0	356.0	35.0
TLC27M9CDR	SOIC	D	14	2500	350.0	350.0	43.0
TLC27M9IDR	SOIC	D	14	2500	350.0	350.0	43.0

TUBE


*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
TLC27M4ACN	N	PDIP	14	25	506	13.97	11230	4.32
TLC27M4ACNE4	N	PDIP	14	25	506	13.97	11230	4.32
TLC27M4AIN	N	PDIP	14	25	506	13.97	11230	4.32
TLC27M4BCN	N	PDIP	14	25	506	13.97	11230	4.32
TLC27M4BIN	N	PDIP	14	25	506	13.97	11230	4.32
TLC27M4CN	N	PDIP	14	25	506	13.97	11230	4.32
TLC27M4CNS	NS	SOP	14	50	530	10.5	4000	4.1
TLC27M4IN	N	PDIP	14	25	506	13.97	11230	4.32
TLC27M9CN	N	PDIP	14	25	506	13.97	11230	4.32
TLC27M9IN	N	PDIP	14	25	506	13.97	11230	4.32



D0014A

PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4220718/A 09/2016

NOTES:

1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.43 mm, per side.
5. Reference JEDEC registration MS-012, variation AB.

EXAMPLE BOARD LAYOUT

D0014A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
SCALE:8X



SOLDER MASK DETAILS

4220718/A 09/2016

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0014A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE:8X

4220718/A 09/2016

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

MECHANICAL DATA

NS (R-PDSO-G**)

PLASTIC SMALL-OUTLINE PACKAGE

14-PINS SHOWN



- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.

N (R-PDIP-T**)

PLASTIC DUAL-IN-LINE PACKAGE

16 PINS SHOWN



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - Falls within JEDEC MS-001, except 18 and 20 pin minimum body length (Dim A).
 - The 20 pin end lead shoulder width is a vendor option, either half or full width.



4220202/B 12/2023

NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-153.

EXAMPLE BOARD LAYOUT

PW0014A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 10X



4220202/B 12/2023

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

PW0014A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE: 10X

4220202/B 12/2023

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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