

# TRF2001 860MHz to 930MHz ISM Band Multiprotocol and Wi-SUN RF FEM

## 1 Features

- 860MHz to 930MHz RF front-end module
- Transmit (TX):
  - Saturated output power ( $P_{SAT}$ ): 28dBm (3.3V)
  - PA gain: 23.1dB
  - PAE: 39% at 27dBm
  - HD2 / HD3: -56dBc / -67dBc
- Receive (RX) LNA:
  - Gain: 15.9dB
  - Noise figure (NF): 1.3dB
  - IP1dB: -6.6dBm
- ANT to RX\_FLT insertion loss: 1.7dB
- Integrated 50Ω RF match
- Integrated linear-in-dB power detector
- Supply voltage: 3.1V to 4.25V
- Total supply current at 3.3V:
  - 386mA (TX  $P_O = 27$ dBm)
  - 60.7mA (TX, no RF)
  - 10.3mA (RX only)
- Low sleep mode current: 0.05μA
- Operating ambient temperature range:
  - 40°C to 85°C

## 2 Applications

- 860MHz to 930MHz wireless systems
- IEEE 802.15.4 systems
- Smart grid and smart meters:
  - [Electricity meters](#)
  - [Water meters](#)
  - [Gas meters](#)
  - [Heat meters](#)
- Smart data [concentrators & collectors](#)
- [Energy infrastructure wireless communications](#)
- [Wireless building automation systems](#)
- [Wireless field transmitters & sensors](#)
- [Wireless EV charging stations](#)

## 3 Description

The TRF2001 is a high-performance RF front-end module (FEM) for low-power wireless applications in sub-1GHz industrial, scientific, and medical (ISM) bands operating over an 860MHz to 930MHz frequency range. The TRF2001 requires minimal external BOM and includes functions such as a range-extender power amplifier (PA) and a low noise amplifier (LNA), transmit-receive and antenna switches, antenna port low pass filter, and a linear-in-dB RF power detector, for a cost-effective design in a space-saving 4.5mm × 4.5mm QFN-28 package.

The TRF2001 increases the link budget and allows range extension in wireless systems by providing high output power with the integrated PA, beyond the capabilities of a WMCU or transceiver, and improves the receiver sensitivity with a low noise figure LNA. The integrated power detector can be used for system calibration or to monitor the power being delivered to the antenna. The device features fully matched, 50Ω RF interfaces for ease of use and design flexibility.

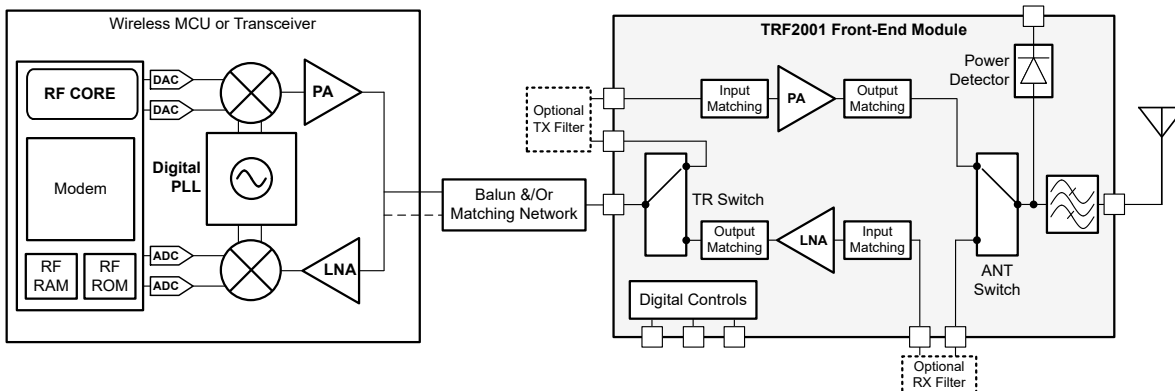
The TRF2001 is operational over a wide, 3.1V to 4.25V supply range, with digital controls that are compatible with 1.6V to 3.3V CMOS logic levels.

### Device Information

PART NUMBER	PACKAGE <sup>(1)</sup>	PACKAGE SIZE <sup>(2)</sup>
TRF2001	VBA (WQFN-FCRLF, 28) with NiPdAu finish	4.5mm × 4.5mm × 0.7mm

(1) For more information, see [Section 10](#).

(2) The package size (length × width × height) is a nominal value and includes pins.



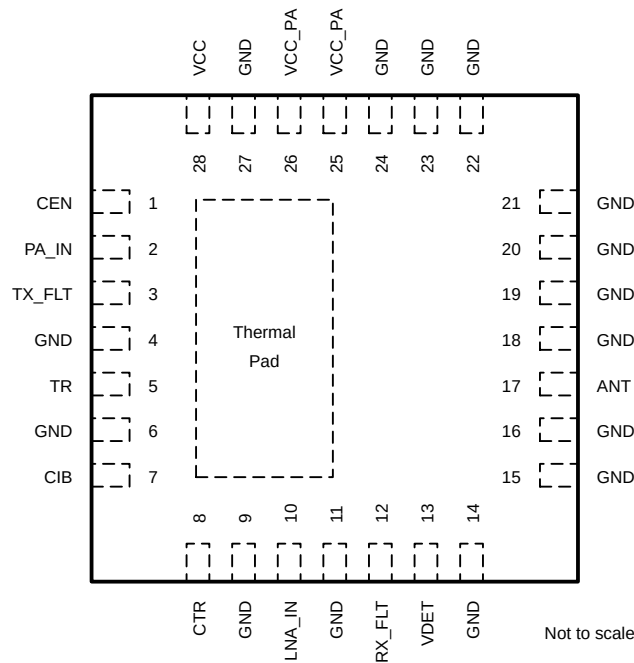
**TRF2001 Typical Configuration with Wireless MCU (WMCU) or Transceiver**



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## 4 Pin Configuration and Functions



**Figure 4-1. VBA Package, 28-Pin WQFN-FCRLF (Top View)**

**Table 4-1. Pin Functions**

PIN		TYPE <sup>(1)</sup>	DESCRIPTION
NAME	NO.		
ANT	17	I/O	Antenna port.
CEN	1	D	Chip enable digital control logic.
CIB	7	D	Internal bias digital control logic.
CTR	8	D	Transmit and receive path select digital control logic.
GND	4, 6, 9, 11, 14, 15, 16, 18, 19, 20, 21, 22, 23, 24, 27	–	RF ground.
LNA_IN	10	I	LNA input. Can be shorted to RX_FLT if receive filter is not required.
PA_IN	2	I	PA input.
RX_FLT	12	O	Receive signal from ANT pin. Typically RX filter connected between RX_FLT and LNA_IN.
TR	5	I/O	Transmit/receive.
TX_FLT	3	O	Transmit signal from TR pin. Typically TX filter connected between TX_FLT and PA_IN.
VCC	28	P	LNA and digital control logic supply voltage.
VCC_PA	25, 26	P	PA supply voltage.
VDET	13	O	Power detector voltage output.
Thermal Pad	Pad	–	Thermal pad and serves as ground reference. Connect to heat-dissipating ground plane on the board.

(1) I = analog input, O = analog output, D = digital control logic, P = power,

## 5 Specifications

### 5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

			MIN	MAX	UNIT
V <sub>CC_PA</sub>	PA supply voltage	No RF		4.5	V
V <sub>CC</sub>	VCC pin supply voltage	No RF		4.5	V
	Input RF level	ANT, RX mode		10	dBm
		TR and PA_IN, TX mode		16	
		LNA_IN		10	
	Digital control logic pins		-0.5	4.25	V
VSWR	ANT voltage standing wave ratio			10:1	-
T <sub>J</sub>	Maximum junction temperature		-40	125	°C
T <sub>stg</sub>	Storage temperature		-55	150	°C

- (1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

### 5.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	±2000	V
		Charged device model (CDM), per JEDEC specification JS-002, all pins <sup>(2)</sup>	±1000	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.  
 (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V <sub>CC_PA</sub>	PA supply voltage	3.1	3.3	4.25	V
V <sub>CC</sub>	VCC pin supply voltage	3.1	3.3	4.25	V
T <sub>A</sub>	Ambient operating air temperature	-40	25	85	°C

### 5.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		TRF2001	UNIT
		VBA (WQFN-FCRLF)	
		28 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	34.0	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	15.5	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	9.6	°C/W
Ψ <sub>JT</sub>	Junction-to-top characterization parameter	0.1	°C/W
Ψ <sub>JB</sub>	Junction-to-board characterization parameter	9.5	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	5.5	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 5.5 Electrical Characteristics

at  $T_A = 25^\circ\text{C}$ ,  $V_{CC\_PA} = 3.3\text{V}$ ,  $V_{CC} = 3.3\text{V}$ ,  $f = 915\text{MHz}$ , shorted RX\_FLT and LNA\_IN pins, shorted TX\_FLT and PA\_IN pins, RF transmit specifications from TR to ANT pin, RF receive specifications from ANT to TR pin,  $50\Omega$  source and load at input and output RF pins, respectively; measured on EVM and de-embedded up to the device pins (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
<b>RF TRANSMIT</b>							
	Frequency range			860		930	MHz
G <sub>P</sub>	Transmit small signal gain	P <sub>IN</sub> = -25dBm	f = 867MHz	23.7			dB
			f = 902MHz	23.4			
			f = 915MHz to 928MHz	23.1			
P <sub>SAT</sub>	Saturated output power	f = 867MHz		27.8			dBm
		f = 902MHz to 928MHz		28.0			
		V <sub>CC_PA</sub> = 3.6V, V <sub>CC</sub> = 3.6V	f = 867MHz to 928MHz	28.6			
OP1dB	Output 1-dB compression point	f = 867MHz to 928MHz		26.5			dBm
		V <sub>CC_PA</sub> = 3.6V, V <sub>CC</sub> = 3.6V	f = 867MHz	26.4			
			f = 902MHz	26.8			
			f = 915MHz	27.6			
			f = 928MHz	26.9			
PAE	Power added efficiency	P <sub>O</sub> = 27dBm	f = 867MHz	41.0%			
			f = 902MHz	39.5%			
			f = 915MHz, 928MHz	39.0%			
HDx	Harmonic distortion <sup>(1)</sup>	f = 867MHz, P <sub>O</sub> = 27dBm	2 <sup>nd</sup> harmonic	-57.0			dBc
			3 <sup>rd</sup> harmonic	-64.5			
		f = 928MHz, P <sub>O</sub> = 27dBm	2 <sup>nd</sup> harmonic	-56.0			
			3 <sup>rd</sup> harmonic	-68.0			
	Input return loss at TR	P <sub>IN</sub> = -27dBm, f = 867MHz to 928MHz		11			dB
	Maximum input power at TR	f = 867MHz	VSWR = 1:1	16			dBm
			VSWR = 2:1	16			
			VSWR = 4:1	7			
		f = 928MHz	VSWR = 1:1	16			
			VSWR = 2:1	11			
			VSWR = 4:1	6.5			
	Ruggedness	CW, P <sub>O</sub> into 50Ω load, VSWR = 10:1, without permanent damage		27			dBm
<b>POWER DETECTOR</b>							
	Power detector power range	Power at ANT pin, f = 860MHz to 930MHz		5		P <sub>SAT</sub>	dBm
V <sub>DET</sub>	Power detector voltage range			0		1.8	V

## 5.5 Electrical Characteristics (continued)

at  $T_A = 25^\circ\text{C}$ ,  $V_{CC\_PA} = 3.3\text{V}$ ,  $V_{CC} = 3.3\text{V}$ ,  $f = 915\text{MHz}$ , shorted RX\_FLT and LNA\_IN pins, shorted TX\_FLT and PA\_IN pins, RF transmit specifications from TR to ANT pin, RF receive specifications from ANT to TR pin,  $50\Omega$  source and load at input and output RF pins, respectively; measured on EVM and de-embedded up to the device pins (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
<b>RF RECEIVE</b>							
	Frequency range			860		930	MHz
	Receive small signal gain	$P_{IN} = -27\text{dBm}$ , $f = 867\text{MHz}$ to $928\text{MHz}$			15.9		dB
NF	Noise figure	$f = 867\text{MHz}$			3.1		dB
		$f = 915\text{MHz}$			3.3		
		LNA_IN to TR, $f = 867\text{MHz}$ to $928\text{MHz}$			1.3		
	ANT to RX_FLT insertion loss	$P_{IN} = -27\text{dBm}$			1.7		dB
IIP3	Input 3 <sup>rd</sup> order compression point	$P_{IN} = -27\text{dBm}$	$f = 867\text{MHz}$		-5.4		dBm
			$f = 902\text{MHz}$ to $928\text{MHz}$		-6.3		
IP1dB	Input 1dB compression point	$f = 867\text{MHz}$ , $902\text{MHz}$			-7.1		dBm
		$f = 915\text{MHz}$ , $928\text{MHz}$			-6.6		
	Input return loss at ANT	$P_{IN} = -27\text{dBm}$	$f = 867\text{MHz}$		18.5		dB
			$f = 902\text{MHz}$		15.5		
			$f = 915\text{MHz}$		14		
			$f = 928\text{MHz}$		12.5		
	Output return loss at TR	$P_{IN} = -27\text{dBm}$	$f = 867\text{MHz}$ to $915\text{MHz}$		15.5		dB
			$f = 928\text{MHz}$		14.5		

## 5.5 Electrical Characteristics (continued)

at  $T_A = 25^\circ\text{C}$ ,  $V_{CC\_PA} = 3.3\text{V}$ ,  $V_{CC} = 3.3\text{V}$ ,  $f = 915\text{MHz}$ , shorted  $\text{RX\_FLT}$  and  $\text{LNA\_IN}$  pins, shorted  $\text{TX\_FLT}$  and  $\text{PA\_IN}$  pins, RF transmit specifications from TR to ANT pin, RF receive specifications from ANT to TR pin,  $50\Omega$  source and load at input and output RF pins, respectively; measured on EVM and de-embedded up to the device pins (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
<b>DC SPECIFICATIONS</b>							
$I_{CC\_PA}$	VCC_PA supply current	TX mode	$P_O = 27\text{dBm}$		351		mA
			$P_O = 24\text{dBm}$		260		
			No RF		60.7		
		RX mode, $P_{IN} = -27\text{dBm}$		0.3			
$I_{CC}$	VCC supply current	TX mode	$P_O = 27\text{dBm}$		35		mA
			$P_O = 24\text{dBm}$		22		
			No RF		7		
		RX mode, $P_{IN} = -27\text{dBm}$		10			
	Powerdown supply current <sup>(2)</sup>	No RF			0.05	1	$\mu\text{A}$
<b>DIGITAL CONTROL LOGIC SPECIFICATIONS</b>							
$V_{IH}$	High voltage threshold	With respect to GND	High (logic 1)	1.6	3.3	3.45	V
$V_{IL}$	Low voltage threshold		Low (logic 0)		0	0.5	V
$I_{IH}$	Pin-high input current	Pin voltage = 3.3V				1	$\mu\text{A}$
$I_{IL}$	Pin-low input current	Pin voltage = 0V				1	$\mu\text{A}$

(1) No external filter on ANT pin. Refer to [Figure 5-4](#) for higher order harmonic performance.

(2) Sum of currents into the VCC\_PA and VCC pins.

## 5.6 Timing Requirements

		MIN	NOM	MAX	UNIT
<b>RF TRANSMIT</b>					
$t_{ON}$	Turn-on time: time for RF output power at ANT to reach 90% of final value from 50% of CTR edge		1.4		$\mu\text{s}$
$t_{OFF}$	Turn-off time: time for RF output power at ANT to reach 10% of final value from 50% of CTR edge		0.1		$\mu\text{s}$
<b>RF RECEIVE</b>					
$t_{ON}$	Turn-on time: time for RF output power at TR to reach 90% of final value from 50% of CTR edge		0.8		$\mu\text{s}$
$t_{OFF}$	Turn-off time: time for RF output power at TR to reach 10% of final value from 50% of CTR edge		0.1		$\mu\text{s}$

## 5.7 Digital Mode Control Logic

Device Pin Configuration			Description
CEN	CIB	CTR	
1	0	0	Not supported
1	1	0	RX mode: RX path enabled (RF receive), TX PA powered down
1	0	1	Not supported
1	1	1	TX mode: TX path enabled (RF transmit)
0	X <sup>(1)</sup>	X <sup>(1)</sup>	Power down mode: Device in powered down state
0	0	0	Guaranteed powered down supply current state

(1) Pin logic is ignored

### 5.8 Typical Characteristics - Transmit

at  $T_A = 25^\circ\text{C}$ ,  $V_{CC\_PA} = V_{CC} = 3.3\text{V}$ , input = TR, output = ANT, RX\_FLT shorted to LNA\_IN, TX\_FLT shorted to PA\_IN, 50Ω source and load at input and output RF pins, respectively, and de-embedded up to the device pins, frequency and ambient temperatures shown (unless otherwise noted)

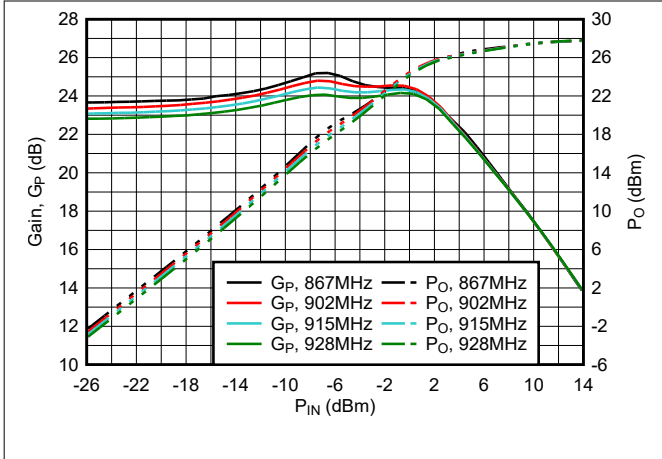


Figure 5-1. Gain and  $P_O$  vs  $P_{IN}$  Across Frequency

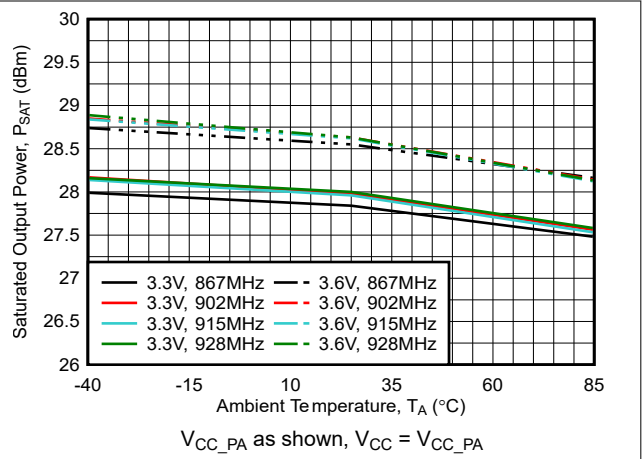


Figure 5-2.  $P_{SAT}$  vs Temperature Across Frequency

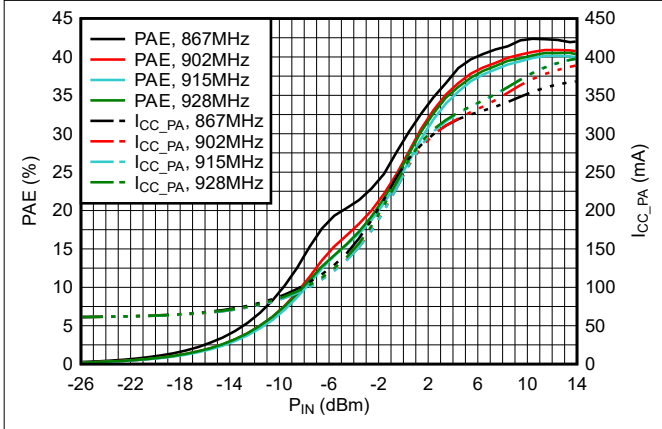


Figure 5-3. PAE and Supply Current vs  $P_{IN}$  Across Frequency

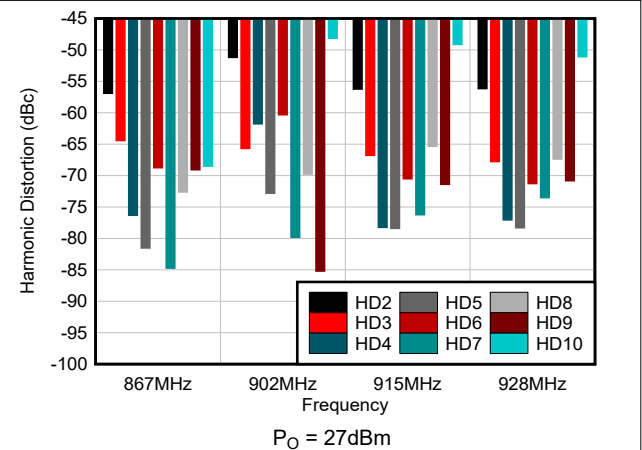


Figure 5-4. Harmonic Distortion Across Frequency

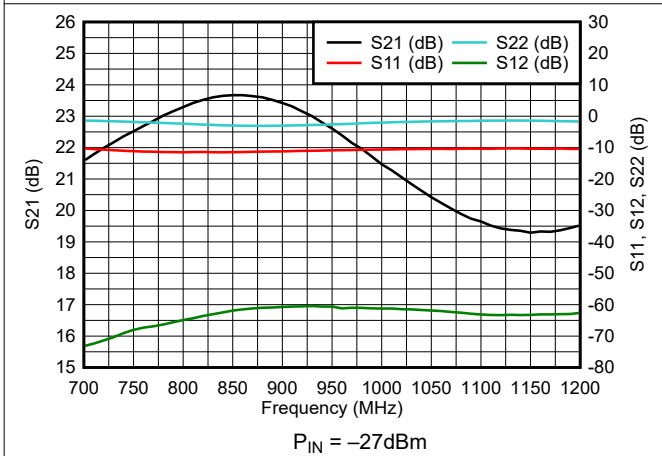


Figure 5-5. Small-Signal S-Parameters

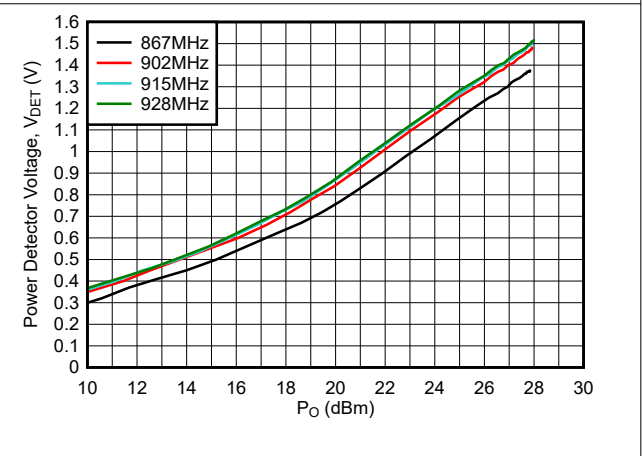


Figure 5-6.  $V_{DET}$  vs  $P_O$  Across Frequency

### 5.8 Typical Characteristics - Transmit (continued)

at  $T_A = 25^\circ\text{C}$ ,  $V_{CC\_PA} = V_{CC} = 3.3\text{V}$ , input = TR, output = ANT, RX\_FLT shorted to LNA\_IN, TX\_FLT shorted to PA\_IN, 50Ω source and load at input and output RF pins, respectively, and de-embedded up to the device pins, frequency and ambient temperatures shown (unless otherwise noted)

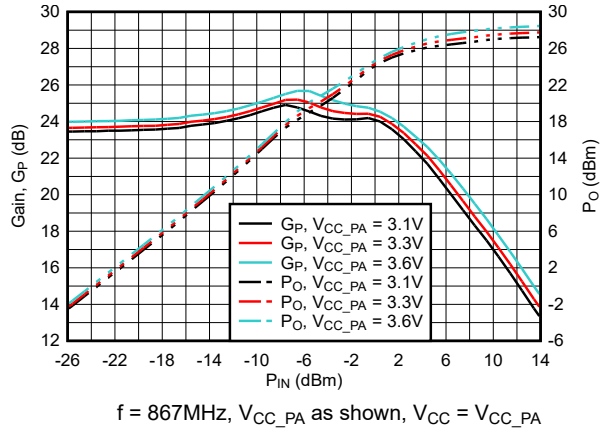


Figure 5-7. Gain and  $P_O$  vs  $P_{IN}$  Across Supply Voltage

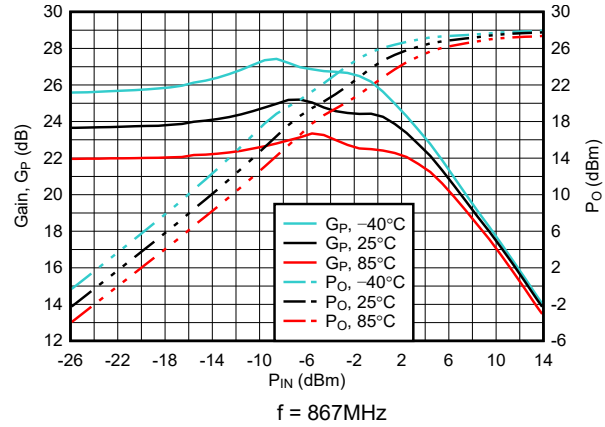


Figure 5-8. Gain and  $P_O$  vs  $P_{IN}$  Across Temperature

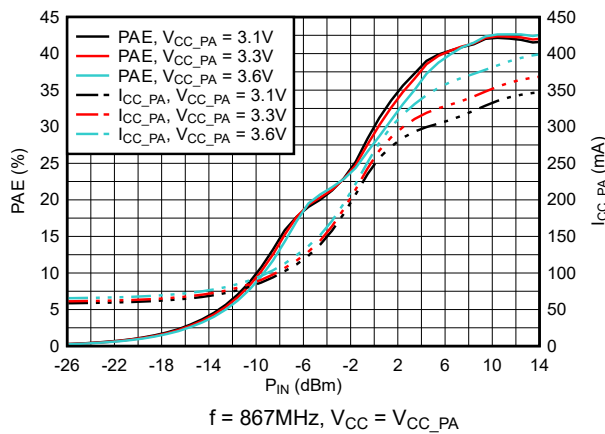


Figure 5-9. PAE and Supply Current vs  $P_{IN}$  Across Supply Voltage

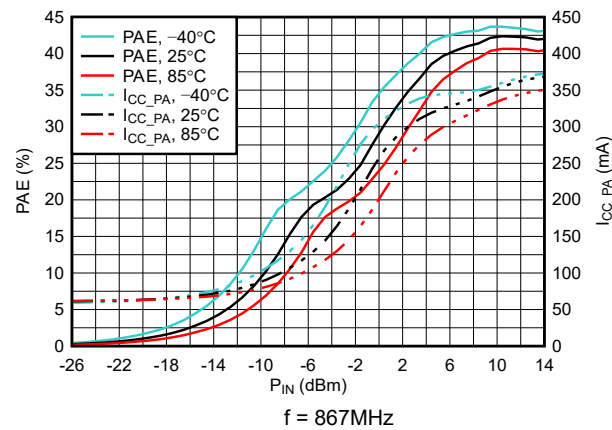


Figure 5-10. PAE and Supply Current vs  $P_{IN}$  Across Temperature

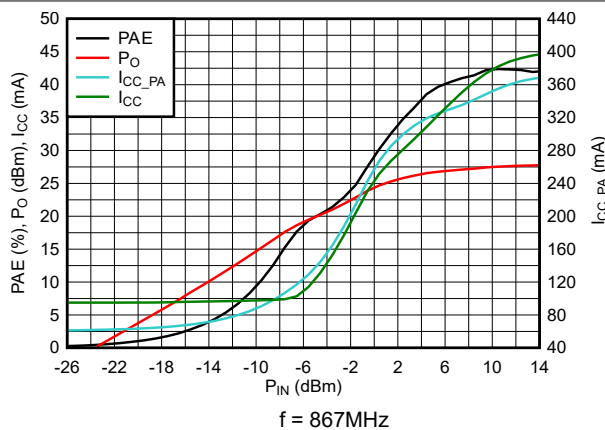


Figure 5-11. PAE,  $P_O$ , and Supply Current vs  $P_{IN}$

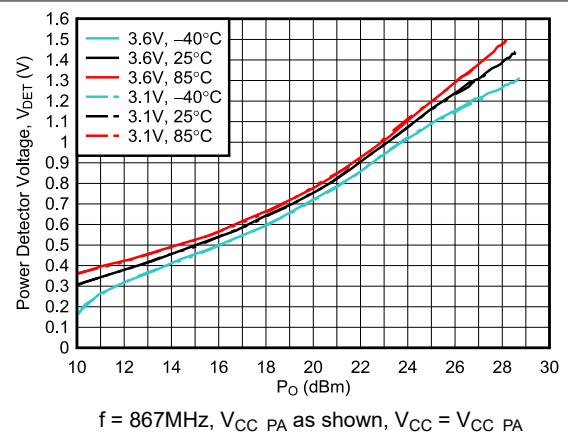


Figure 5-12.  $V_{DET}$  vs  $P_O$  Across Supply Voltage and Temperature

## 5.8 Typical Characteristics - Transmit (continued)

at  $T_A = 25^\circ\text{C}$ ,  $V_{CC\_PA} = V_{CC} = 3.3\text{V}$ , input = TR, output = ANT, RX\_FLT shorted to LNA\_IN, TX\_FLT shorted to PA\_IN, 50Ω source and load at input and output RF pins, respectively, and de-embedded up to the device pins, frequency and ambient temperatures shown (unless otherwise noted)

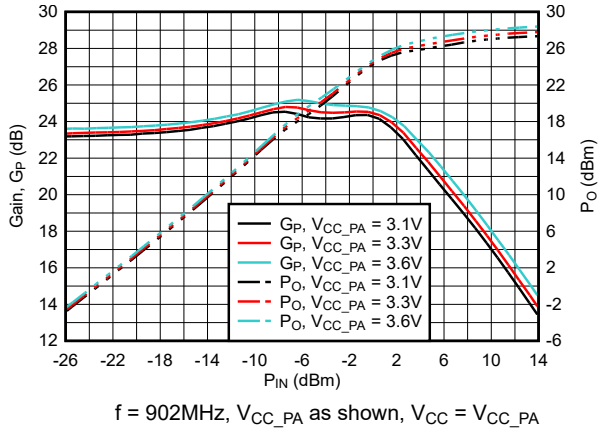


Figure 5-13. Gain and  $P_O$  vs  $P_{IN}$  Across Supply Voltage

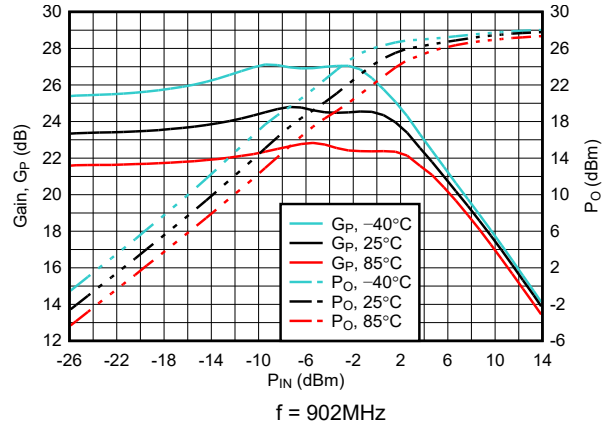


Figure 5-14. Gain and  $P_O$  vs  $P_{IN}$  Across Temperature

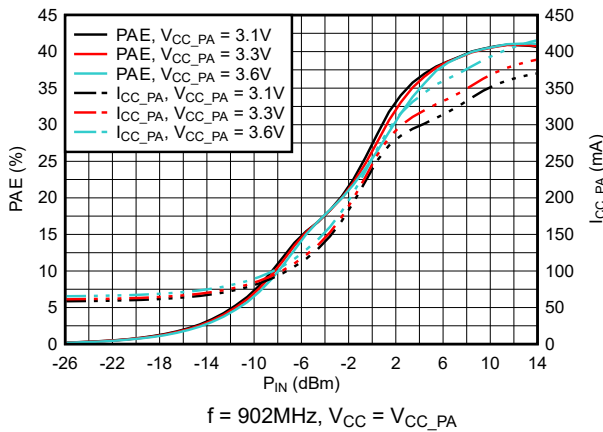


Figure 5-15. PAE and Supply Current vs  $P_{IN}$  Across Supply Voltage

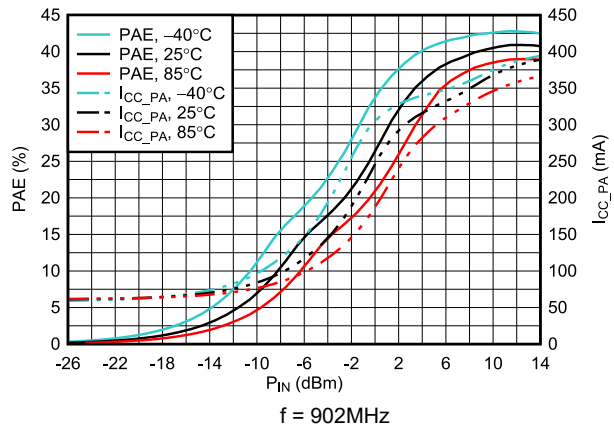


Figure 5-16. PAE and Supply Current vs  $P_{IN}$  Across Temperature

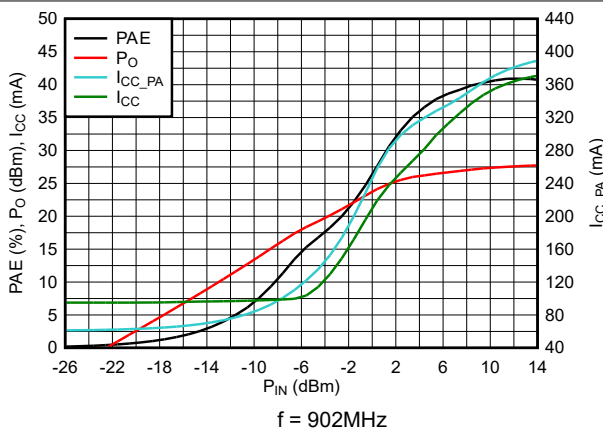


Figure 5-17. PAE,  $P_O$ , and Supply Current vs  $P_{IN}$

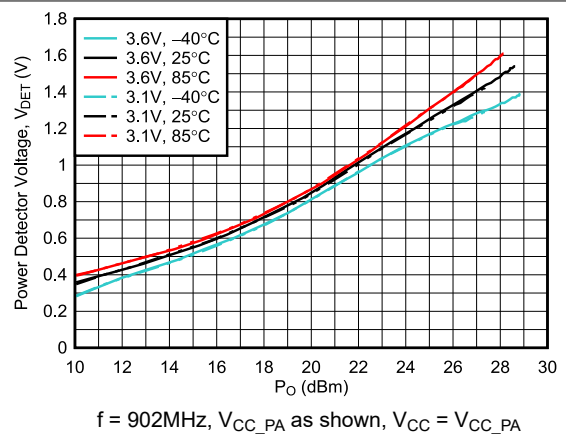


Figure 5-18.  $V_{DET}$  vs  $P_O$  Across Supply Voltage and Temperature

### 5.8 Typical Characteristics - Transmit (continued)

at  $T_A = 25^\circ\text{C}$ ,  $V_{CC\_PA} = V_{CC} = 3.3\text{V}$ , input = TR, output = ANT, RX\_FLT shorted to LNA\_IN, TX\_FLT shorted to PA\_IN, 50Ω source and load at input and output RF pins, respectively, and de-embedded up to the device pins, frequency and ambient temperatures shown (unless otherwise noted)

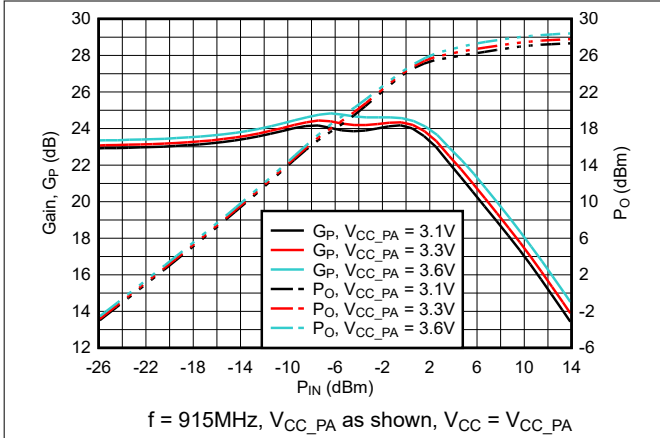


Figure 5-19. Gain and  $P_O$  vs  $P_{IN}$  Across Supply Voltage

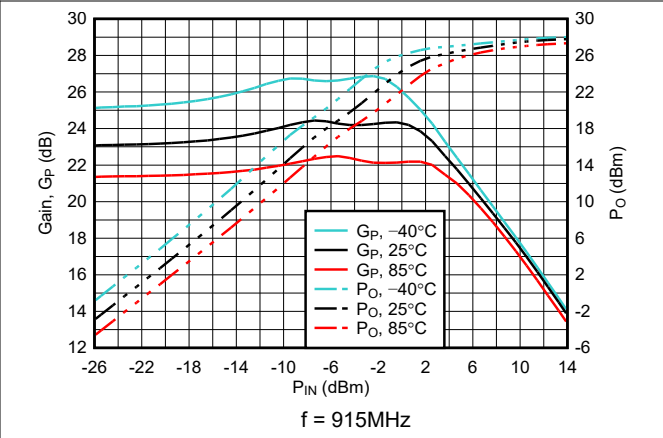


Figure 5-20. Gain and  $P_O$  vs  $P_{IN}$  Across Temperature

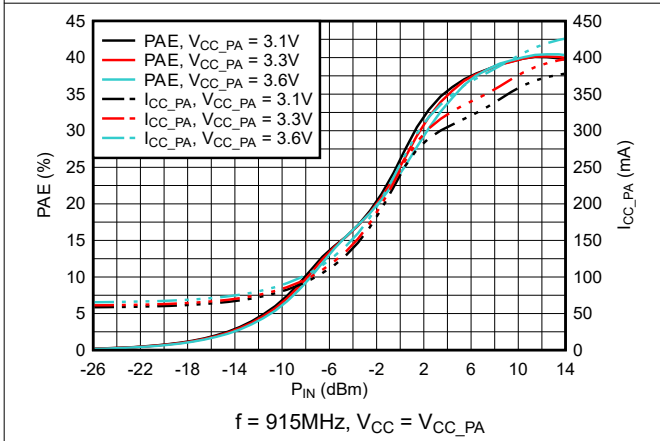


Figure 5-21. PAE and Supply Current vs  $P_{IN}$  Across Supply Voltage

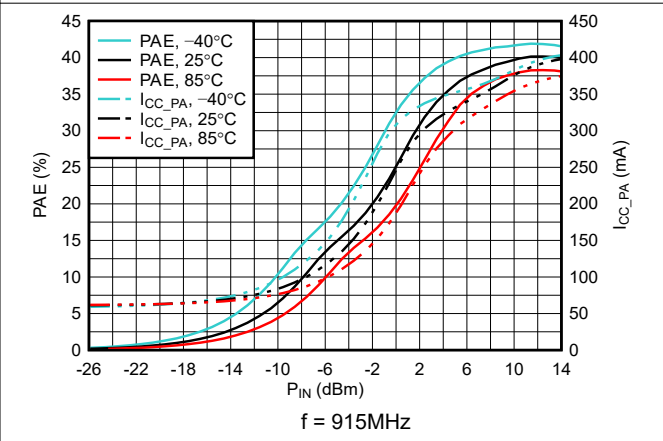


Figure 5-22. PAE and Supply Current vs  $P_{IN}$  Across Temperature

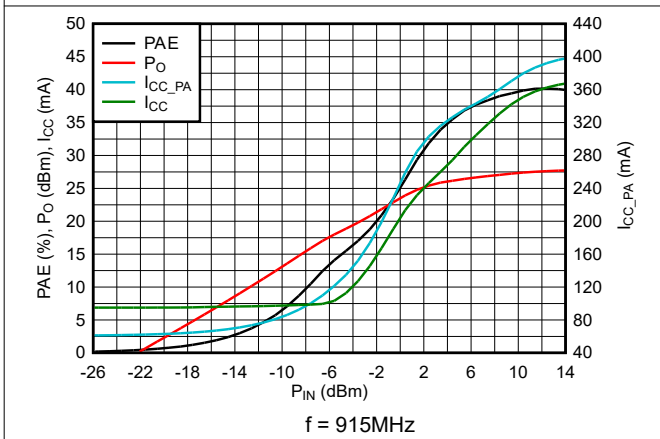


Figure 5-23. PAE,  $P_O$ , and Supply Current vs  $P_{IN}$

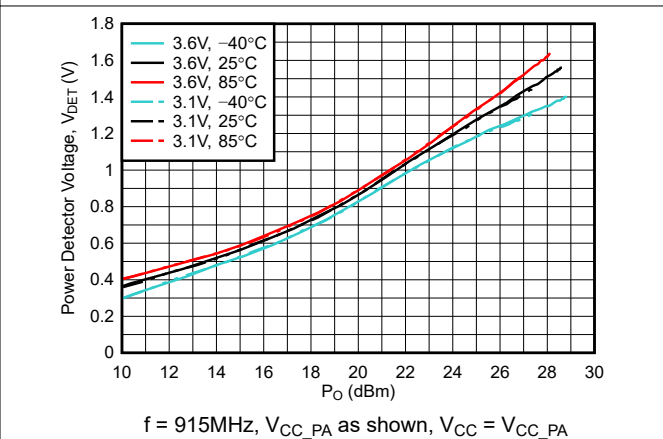


Figure 5-24.  $V_{DET}$  vs  $P_O$  Across Supply Voltage and Temperature

### 5.8 Typical Characteristics - Transmit (continued)

at  $T_A = 25^\circ\text{C}$ ,  $V_{CC\_PA} = V_{CC} = 3.3\text{V}$ , input = TR, output = ANT, RX\_FLT shorted to LNA\_IN, TX\_FLT shorted to PA\_IN, 50Ω source and load at input and output RF pins, respectively, and de-embedded up to the device pins, frequency and ambient temperatures shown (unless otherwise noted)

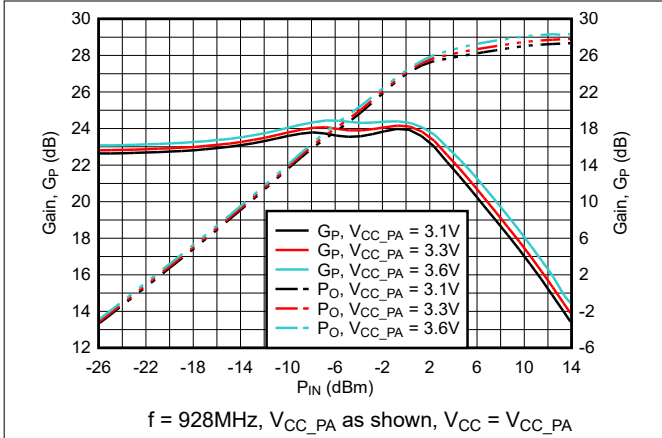


Figure 5-25. Gain and  $P_O$  vs  $P_{IN}$  Across Supply Voltage

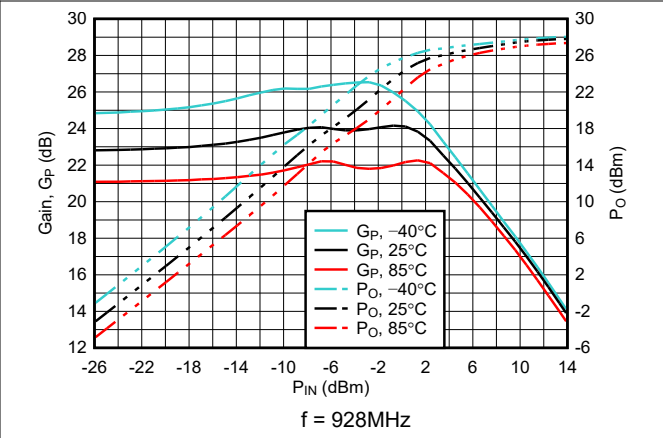


Figure 5-26. Gain and  $P_O$  vs  $P_{IN}$  Across Temperature

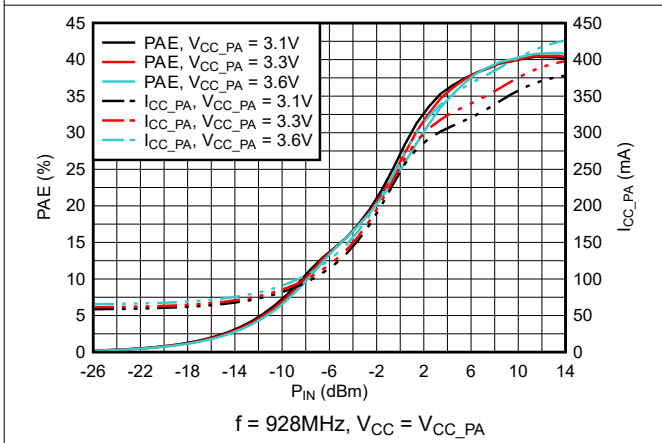


Figure 5-27. PAE and Supply Current vs  $P_{IN}$  Across Supply Voltage

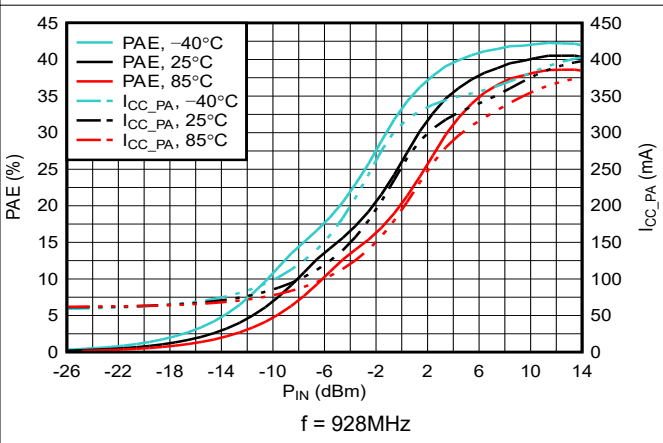


Figure 5-28. PAE and Supply Current vs  $P_{IN}$  Across Temperature

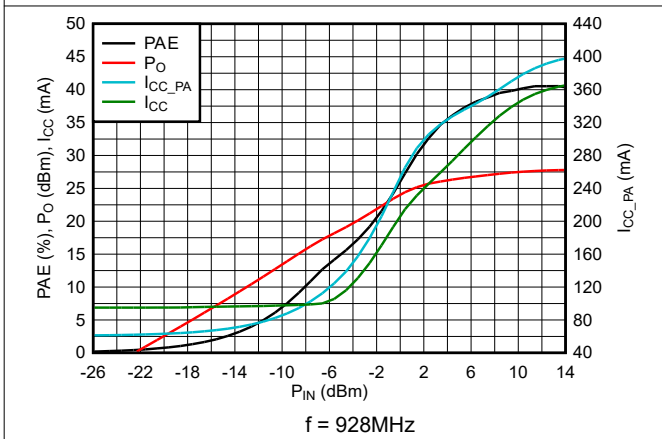


Figure 5-29. PAE,  $P_O$ , and Supply Current vs  $P_{IN}$

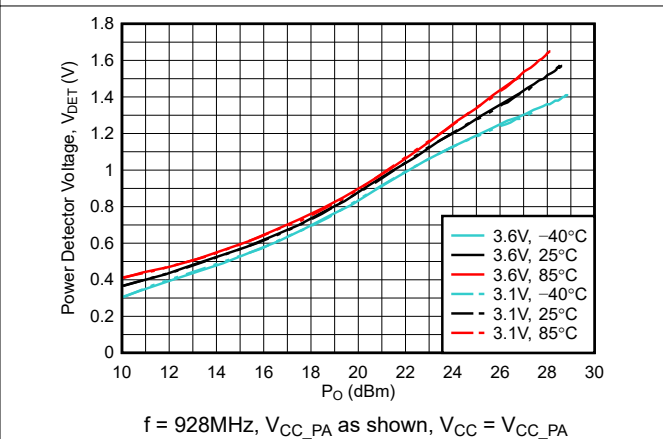


Figure 5-30.  $V_{DET}$  vs  $P_O$  Across Supply Voltage and Temperature

### 5.9 Typical Characteristics - Receive

at  $T_A = 25^\circ\text{C}$ ,  $V_{CC\_PA} = V_{CC} = 3.3\text{V}$ , input = ANT, output = TR, RX\_FLT shorted to LNA\_IN, TX\_FLT shorted to PA\_IN,  $50\Omega$  source and load at input and output RF pins, respectively, and de-embedded up to the device pins, frequency and ambient temperatures shown (unless otherwise noted)

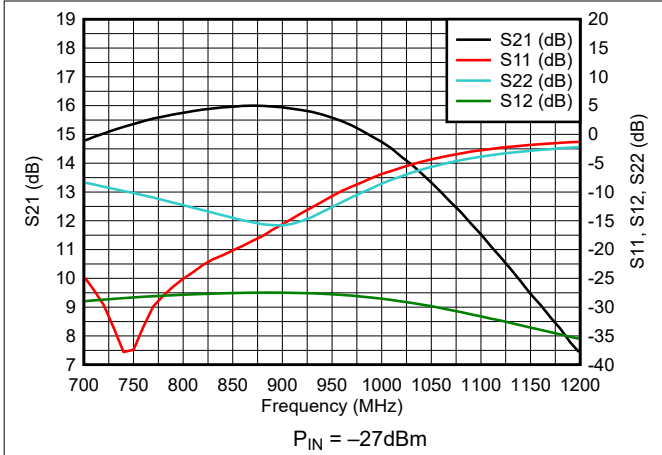


Figure 5-31. Small-Signal S-Parameters

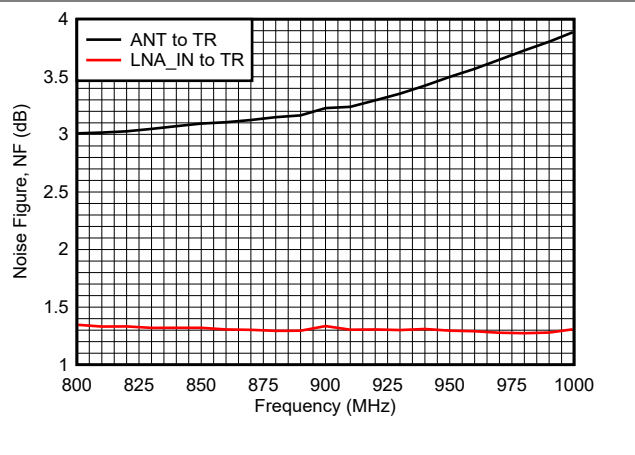


Figure 5-32. Noise Figure

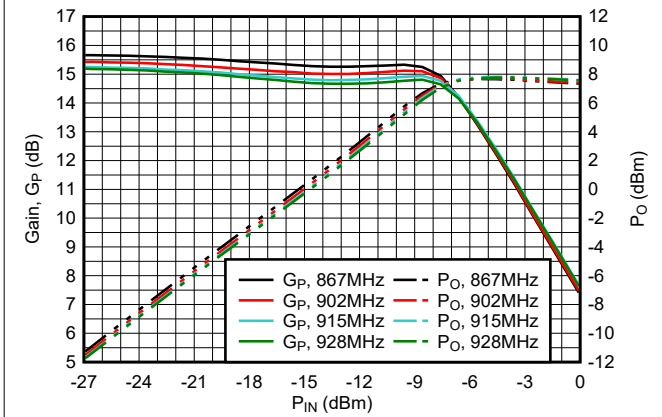


Figure 5-33. Gain and  $P_O$  vs  $P_{IN}$  Across Frequency

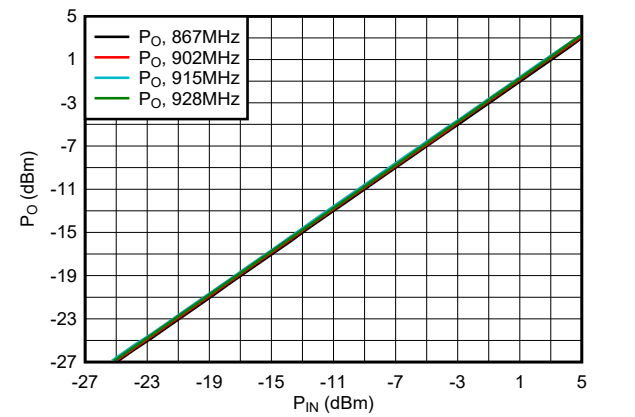


Figure 5-34.  $P_O$  vs  $P_{IN}$  Across Frequency

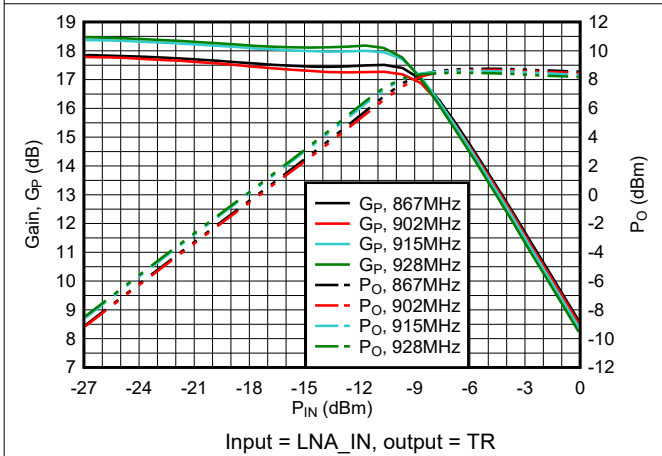


Figure 5-35. Gain and  $P_O$  vs  $P_{IN}$  Across Frequency

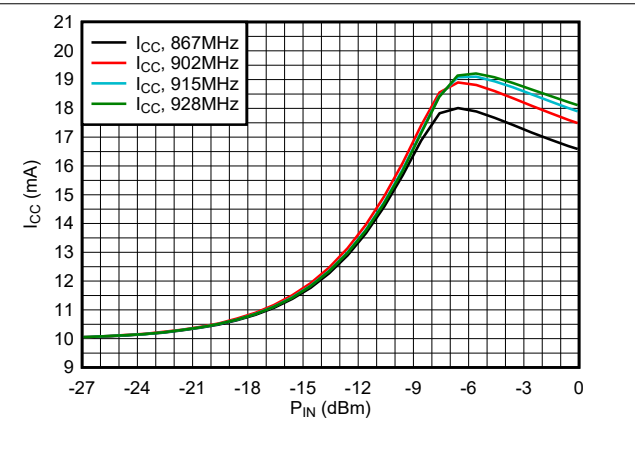


Figure 5-36. Supply Current vs  $P_{IN}$

### 5.9 Typical Characteristics - Receive (continued)

at  $T_A = 25^\circ\text{C}$ ,  $V_{CC\_PA} = V_{CC} = 3.3\text{V}$ , input = ANT, output = TR, RX\_FLT shorted to LNA\_IN, TX\_FLT shorted to PA\_IN,  $50\Omega$  source and load at input and output RF pins, respectively, and de-embedded up to the device pins, frequency and ambient temperatures shown (unless otherwise noted)

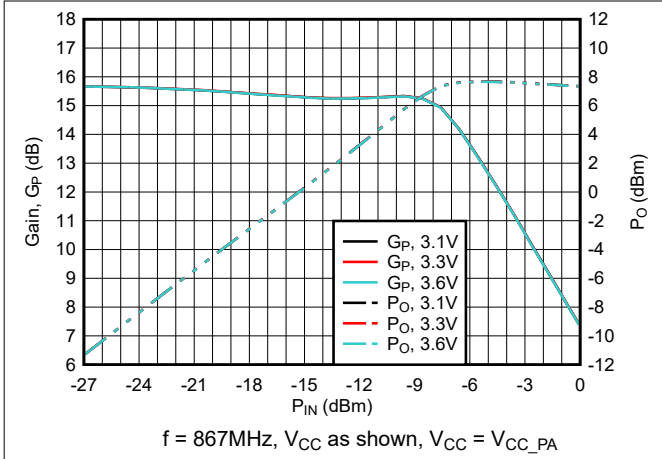


Figure 5-37. Gain and  $P_O$  vs  $P_{IN}$  Across Supply Voltage

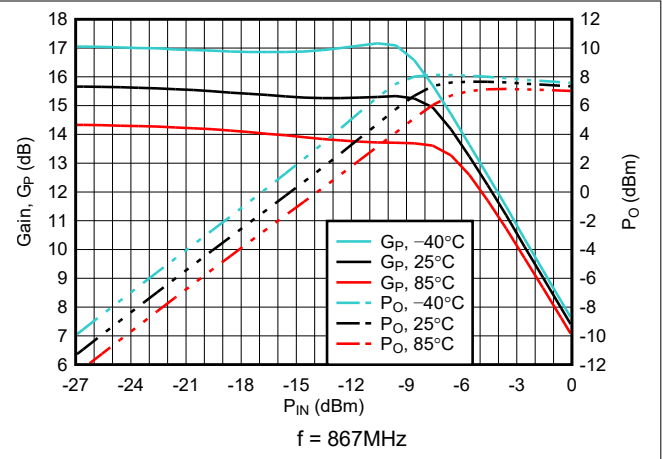


Figure 5-38. Gain and  $P_O$  vs  $P_{IN}$  Across Temperature

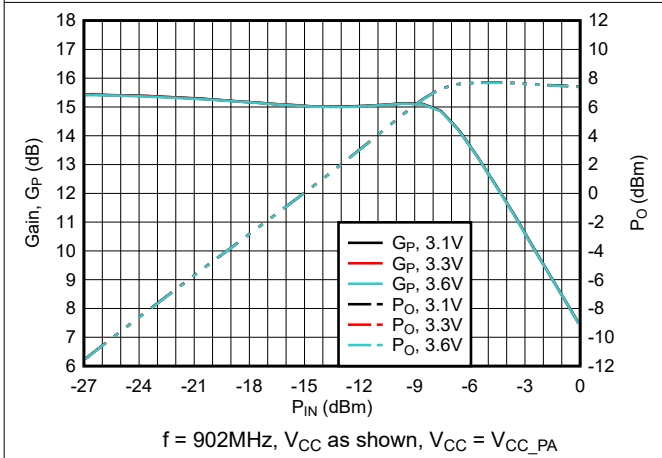


Figure 5-39. Gain and  $P_O$  vs  $P_{IN}$  Across Supply Voltage

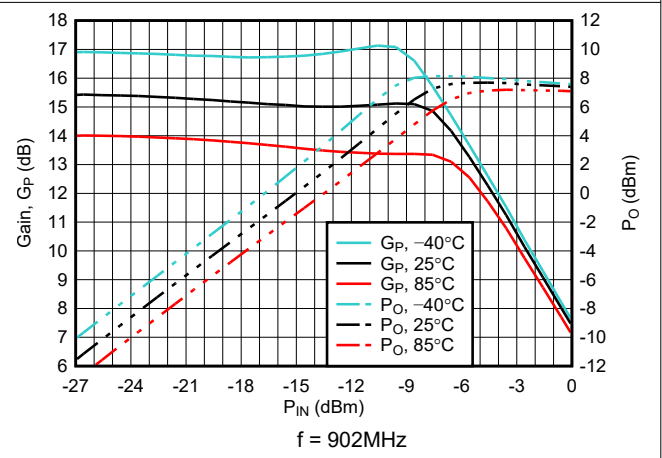


Figure 5-40. Gain and  $P_O$  vs  $P_{IN}$  Across Temperature

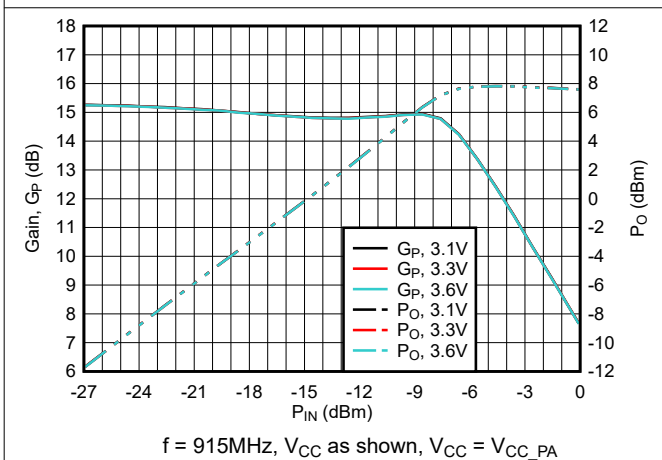


Figure 5-41. Gain and  $P_O$  vs  $P_{IN}$  Across Supply Voltage

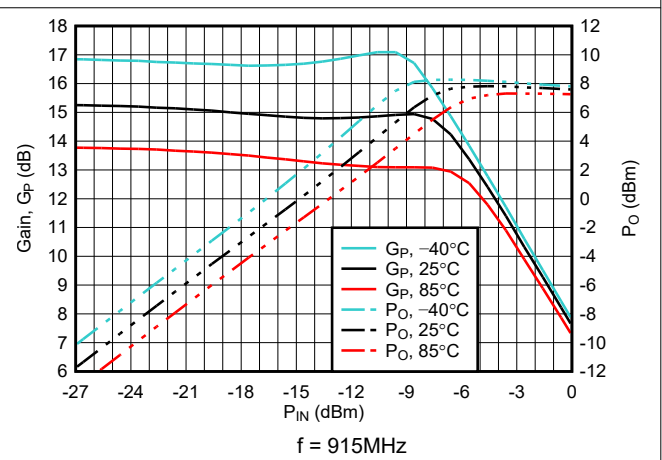


Figure 5-42. Gain and  $P_O$  vs  $P_{IN}$  Across Temperature

### 5.9 Typical Characteristics - Receive (continued)

at  $T_A = 25^\circ\text{C}$ ,  $V_{CC\_PA} = V_{CC} = 3.3\text{V}$ , input = ANT, output = TR, RX\_FLT shorted to LNA\_IN, TX\_FLT shorted to PA\_IN, 50Ω source and load at input and output RF pins, respectively, and de-embedded up to the device pins, frequency and ambient temperatures shown (unless otherwise noted)

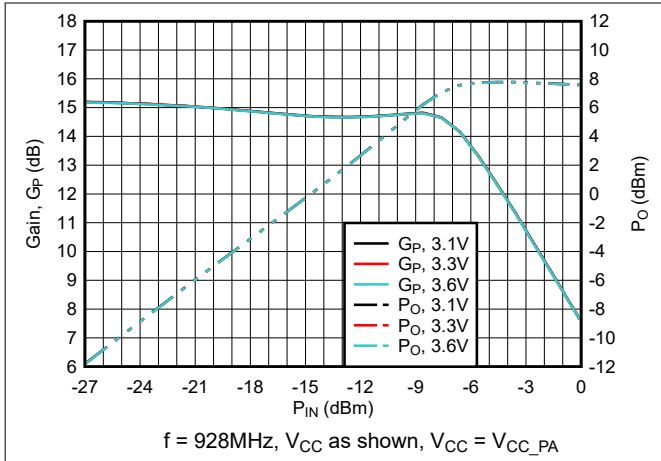


Figure 5-43. Gain and  $P_O$  vs  $P_{IN}$  Across Supply Voltage

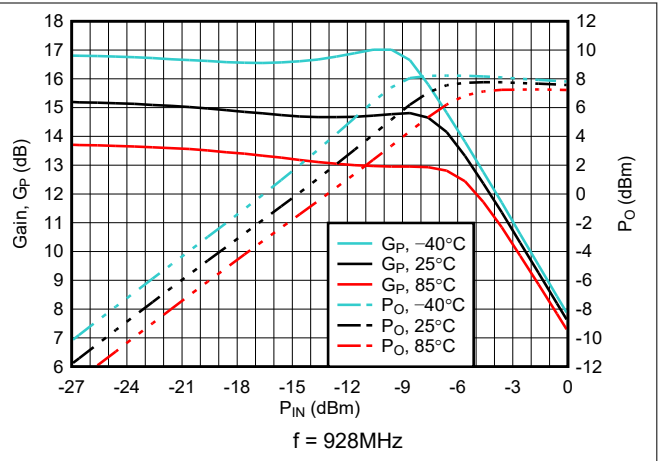


Figure 5-44. Gain and  $P_O$  vs  $P_{IN}$  Across Temperature

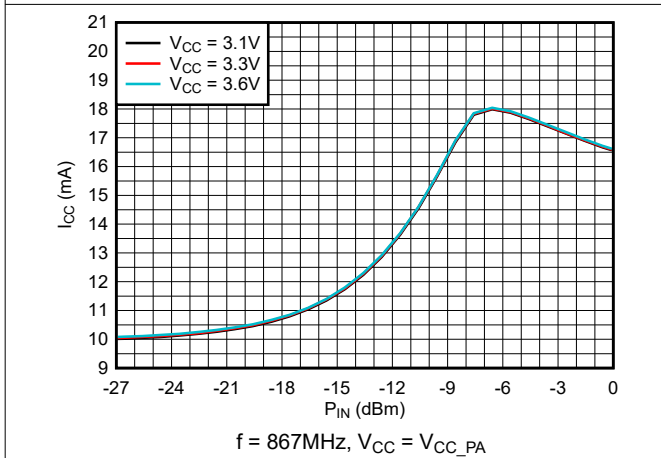


Figure 5-45. Supply Current vs  $P_{IN}$  Across Supply Voltage

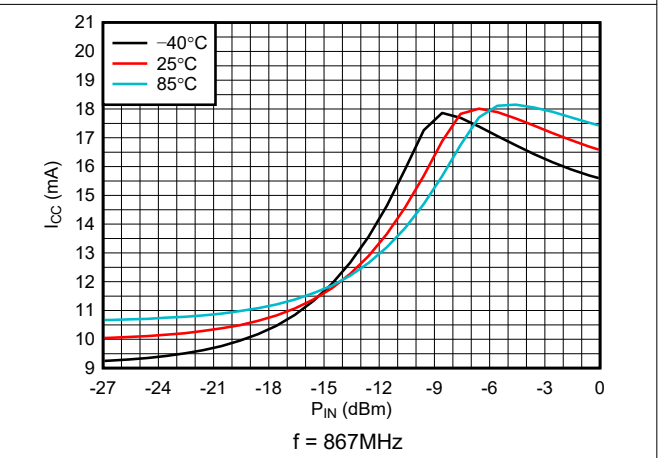


Figure 5-46. Supply Current vs  $P_{IN}$  Across Temperature

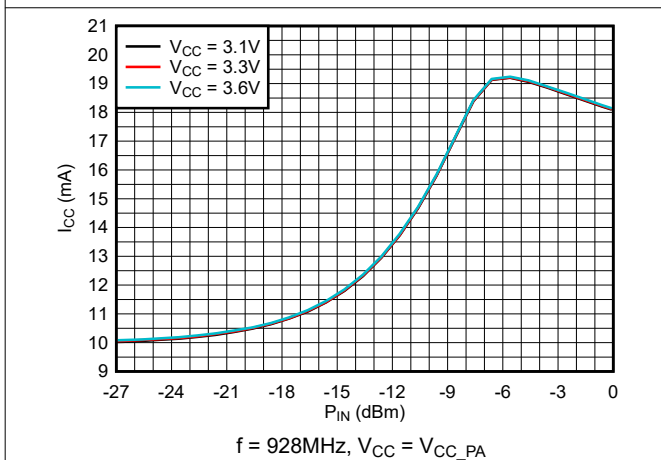


Figure 5-47. Supply Current vs  $P_{IN}$  Across Supply Voltage

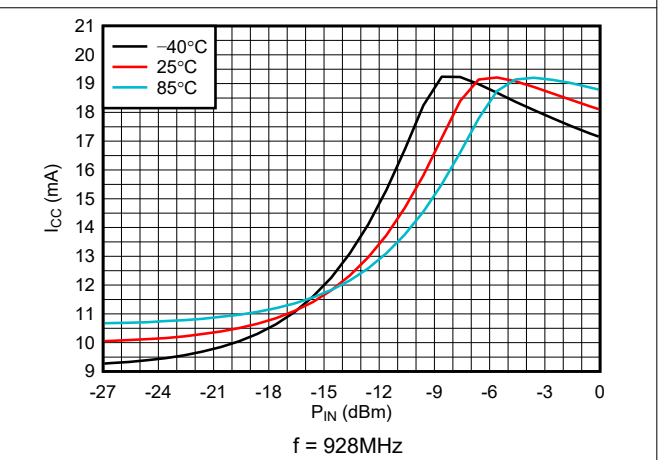


Figure 5-48. Supply Current vs  $P_{IN}$  Across Temperature

## 6 Detailed Description

### 6.1 Overview

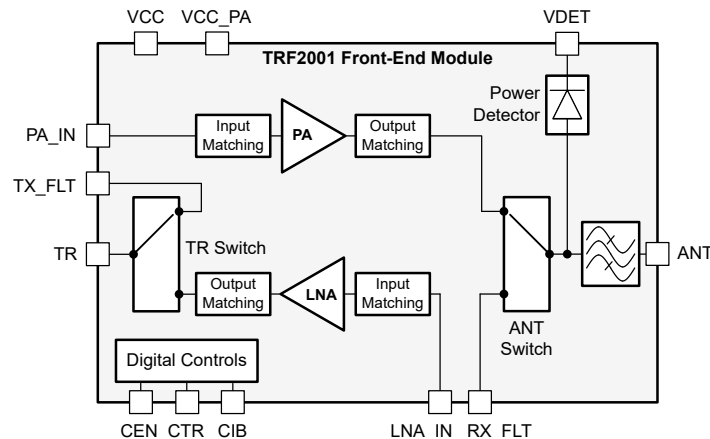
The TRF2001 is a high-performance RF front-end module (FEM) designed to be paired with wireless microcontroller (MCU) and system-on-chip (SoC) ICs for applications primarily supporting various sub-1GHz ISM bands. The device improves the link budget by increasing the system's TX power well beyond the capabilities of the wireless MCUs and SoCs with the integrated PA and improving the RX sensitivity with the integrated LNA.

A key limitation of most wireless SoCs designed for the sub-1GHz ISM band is that the transmit (TX) output power is typically limited to about 16dBm, with very few supporting up to 22dBm. When these SoCs are used at TX powers greater than 10dBm, the spurious-free dynamic range (SFDR) often deteriorates. The increased spurious levels require additional filtering and metal shielding in many cases, to comply with regulatory requirements such as those set by the FCC and ETSI standards. The TRF2001 allows the SoCs to operate at lower output power levels by providing RF gain and high TX output power capability that exceed 27dBm, thus reducing the SoC's spurious levels, and often eliminating the need for additional filtering and shielding.

With integrated, fully-matched 50Ω RF interfaces, the TRF2001 simplifies interface design to the antenna and the wireless SoCs. The digital control logic pins (CEN, CTR, and CIB) of the device are used to configure the device in TX, RX, or device powered down mode, and are compatible with CMOS levels of 1.6V to 3.3V.

The TRF2001 operates over a wide single supply voltage range of 3.1V to 4.25V and achieves over 27.8dBm  $P_{SAT}$  on 3.3V supply. The device is available in a space-saving 4.5mm × 4.5mm, 28-pin, WQFN-FCRLP package.

### 6.2 Functional Block Diagram



### 6.3 Feature Description

Besides the PA, LNA, the TR switch, and the ANT switch, the TRF2001 integrates additional functions such as a harmonic rejection filter and a power detector in a small 4.5mm × 4.5mm package.

The integrated harmonic rejection filter either eliminates the need for an external filter to the antenna or notably relaxes the rejection requirements of the external filter if the system requires one. The TRF2001 inherently achieves second and third order harmonic rejection below  $-56\text{dBc}$  and  $-64.5\text{dBc}$ , respectively, at  $P_O = 27\text{dBm}$  without any external filtering. The integrated power detector provides analog voltage output corresponding to the output power as shown in [Figure 5-6](#) and has a very stable response across temperature and supply voltage.

### 6.4 Device Functional Modes

The TRF2001 has three functional modes: Transmit (TX), Receive (RX), and power down mode. The operating mode of the device is set using the digital control pins, CEN, CTR, and CIB as shown in [Digital Mode Control Logic](#). In the TX mode, the PA path is enabled by connecting the TR switch to the TX\_FLT pin and the ANT switch to the PA output path. In the RX mode, the TR switch connects to the LNA path and the ANT switch connects to the RX\_FLT pin. For the TX path, connect an optional TX filter between TX\_FLT and PA\_IN to reject out-of-band spurs from the SoC or short the TX\_FLT to PA\_IN. Connect an optional RX filter between RX\_FLT and LNA\_IN pins to reject out-of-band signals coming in at the ANT pin or short RX\_FLT to LNA\_IN to engage the LNA in the receive path.

## 7 Application and Implementation

### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

### 7.1 Application Information

The TRF2001 is a FEM typically used with wireless SoCs to extend the communication range beyond what the SoCs are natively capable of, and to provide an improved link budget. The primary application of the TRF2001 is in the ISM band frequency range of 860MHz to 930MHz.

### 7.2 Typical Application

#### 7.2.1 TRF2001 as Range Extender

Figure 7-1 shows a typical application of the TRF2001 when used as a range extender, paired with CC1314R10 or similar wireless MCUs.

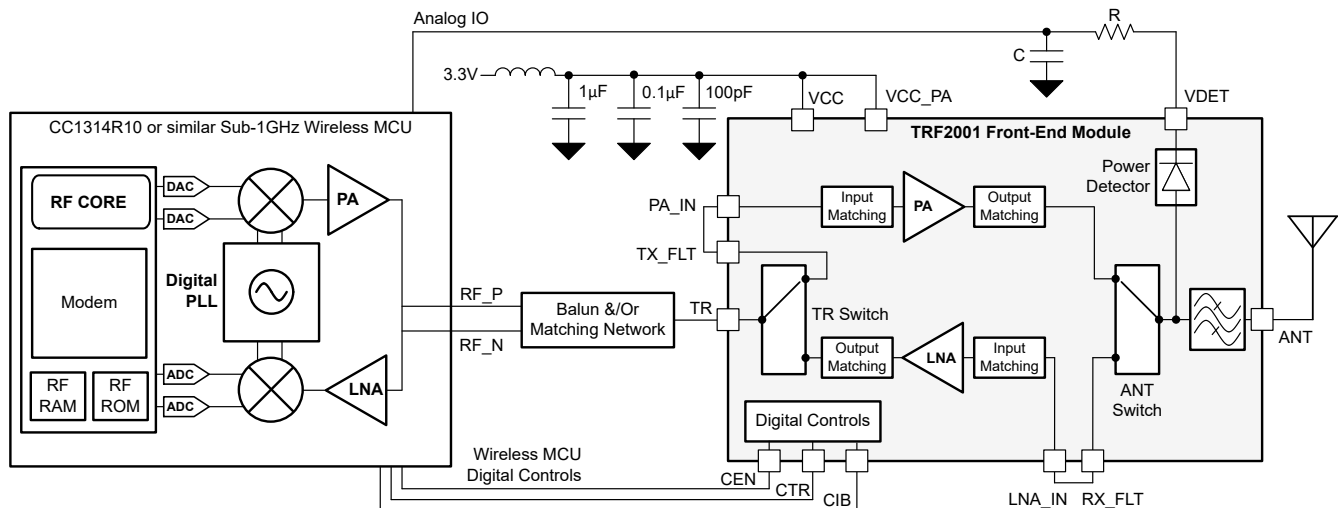


Figure 7-1. TRF2001 as Range Extender Paired with CC1314R10 Wireless MCU

### 7.2.1.1 Design Requirements

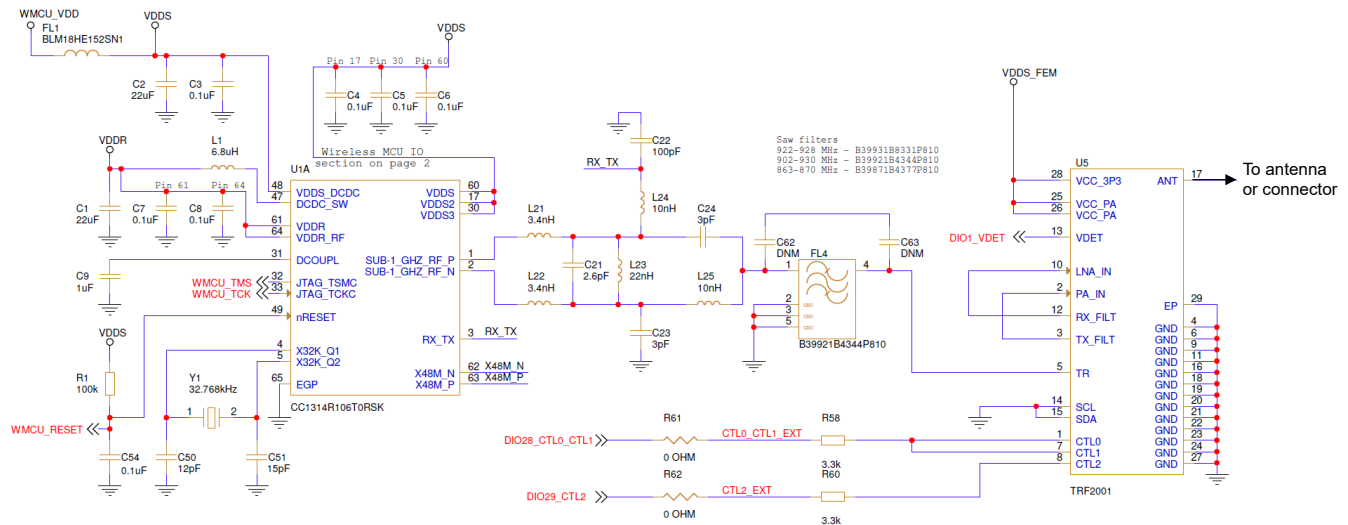
Use TRF2001 to extend the communication range and the TX output power in a sub-1GHz system using CC1314R10 wireless MCU. The design parameters in [Table 7-1](#) are for a transmitter and receiver setup, operating at 915MHz frequency with an approximately 1:1 VSWR antenna system.

**Table 7-1. Design Parameters**

SETUP	PROTOCOL	CONFIGURATION	DESIGN PARAMETER	VALUE
Setup A: Over-the-air (OTA) packet error rate (PER) testing	SimpleLink Long Range	<ul style="list-style-type: none"> <li>34kHz bandwidth</li> <li>1:4 direct sequence spread spectrum (DSSS)</li> <li>2.5kbps effective data rate</li> </ul>	TX power	≥ 27dBm
			PER%	≤ 10%
Setup B: Receiver sensitivity testing	Wi-SUN 2-GFSK	<ul style="list-style-type: none"> <li>50kbps</li> <li>±25kHz deviation</li> <li>100kHz RX bandwidth</li> </ul>	RX sensitivity improvement at antenna for PER < 10% vs standalone CC1314R10	≥ 5dB
				≥ 7dB

### 7.2.1.2 Detailed Design Procedure

The TRF2001 has integrated 50Ω matching elements and in a 50Ω matched system, does not require any external matching components at the TR or ANT pins. Use the [CC1314R10 LaunchPad™](#) design as reference to design the balun and the matching network interface between the CC1314RF10 and TRF2001 as shown in [Figure 7-2](#). Add an optional SAW filter, FL4 in [Figure 7-2](#), to filter any spurs from CC1314RF10 from reaching the TRF2001.

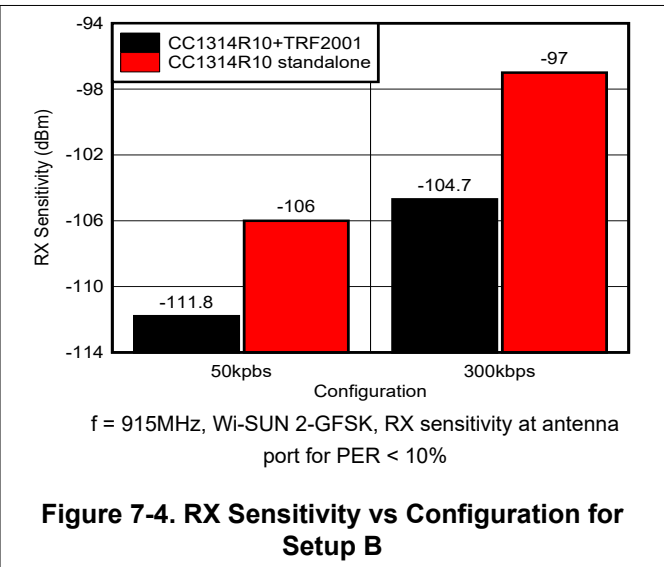
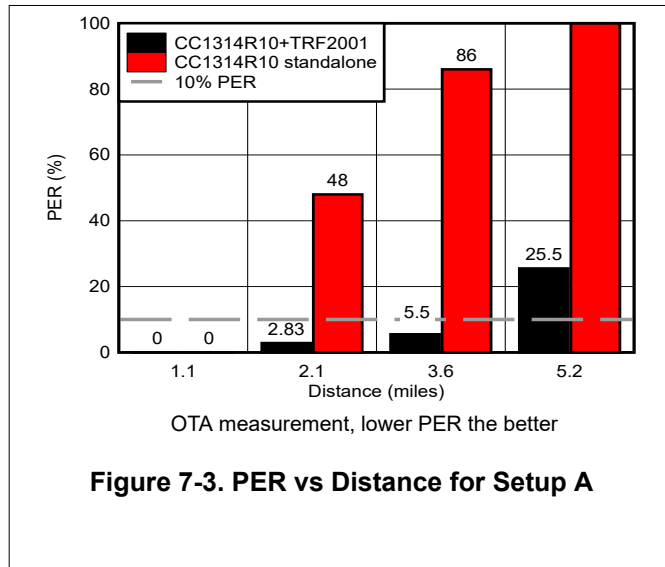


**Figure 7-2. TRF2001 Interface Design to CC1314R10 Wireless MCU**

### 7.2.1.3 Application Curves

Figure 7-3 shows the OTA PER results for setup A, measured over a relatively straight distance between the transmitter and the receiver, with TRF2001 as the range extender function (CC1314R10+TRF2001) and without the TRF2001 (CC1314R10 standalone). PER of approximately 5.5%, well below the 10% target, is achieved over a distance of 3.6 miles with TRF2001, while without the TRF2001 PER worsens by almost 5× the target.

Figure 7-4 shows the RX sensitivity comparison for setup B, with and without the TRF2001 as the range extender. The conducted measurements show, TRF2001 improves RX sensitivity by approximately 6dB to 8dB versus the standalone CC1314R10 setup.



## 7.3 Power Supply Recommendations

### 7.4 Layout

#### 7.4.1 Layout Guidelines

Figure 7-5 shows example layout for TRF2001. Only the top signal layer (layer 1) and second ground layer (layer 2) are shown. Use a multilayer board to maintain signal integrity and power integrity.

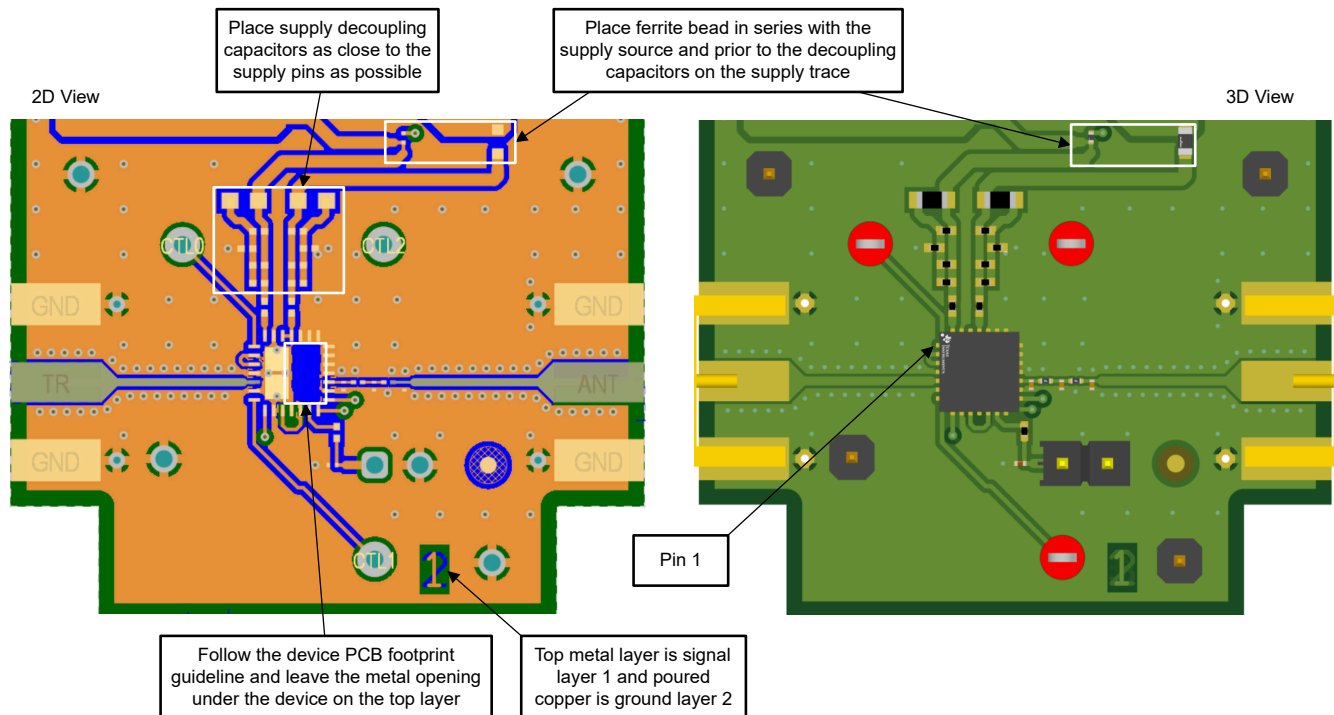
- Route the RF signals as grounded coplanar waveguide (GCPW) traces.
- Maintain that the ground planes on the top and any internal layers are well stitched with vias, and the second layer of the PCB has a continuous ground layer without any cutouts in the vicinity of the device.
- Avoid routing clocks and digital control lines near RF signal lines.
- Do not route RF or DC signal lines over noisy power planes.
- Place supply decoupling caps close to the device.
- Use small-footprint, passive components wherever possible.

See the [TRF2001 Evaluation Module user's guide](#) for more details on board layout and design. The TRF2001 can be evaluated using EVM boards that can be ordered from the [TRF2001 Evaluation Module](#) web page.

##### 7.4.1.1 Thermal Considerations

The TRF2001 is packaged in a WQFN-FCRLF package that has excellent thermal properties. Connect the thermal pads underneath the device to the thermally dissipative ground plane on the board. For good thermal design, use thermal vias to connect the thermal pad plane on the top layer of the PCB to the ground planes in the inner layers.

#### 7.4.2 Layout Example



**Figure 7-5. Layout Example**

## 8 Device and Documentation Support

### 8.1 Third-Party Products Disclaimer

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### 8.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on [ti.com](http://ti.com). Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 8.3 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

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All trademarks are the property of their respective owners.

### 8.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 8.6 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

## 9 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

DATE	REVISION	NOTES
April 2026	*	Initial Release

## 10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

## PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package   Pins	Package qty   Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
<a href="#">TRF2001VBAR</a>	Active	Production	WQFN-FCRLF (VBA)   28	3000   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	T201A21

(1) **Status:** For more details on status, see our [product life cycle](#).

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

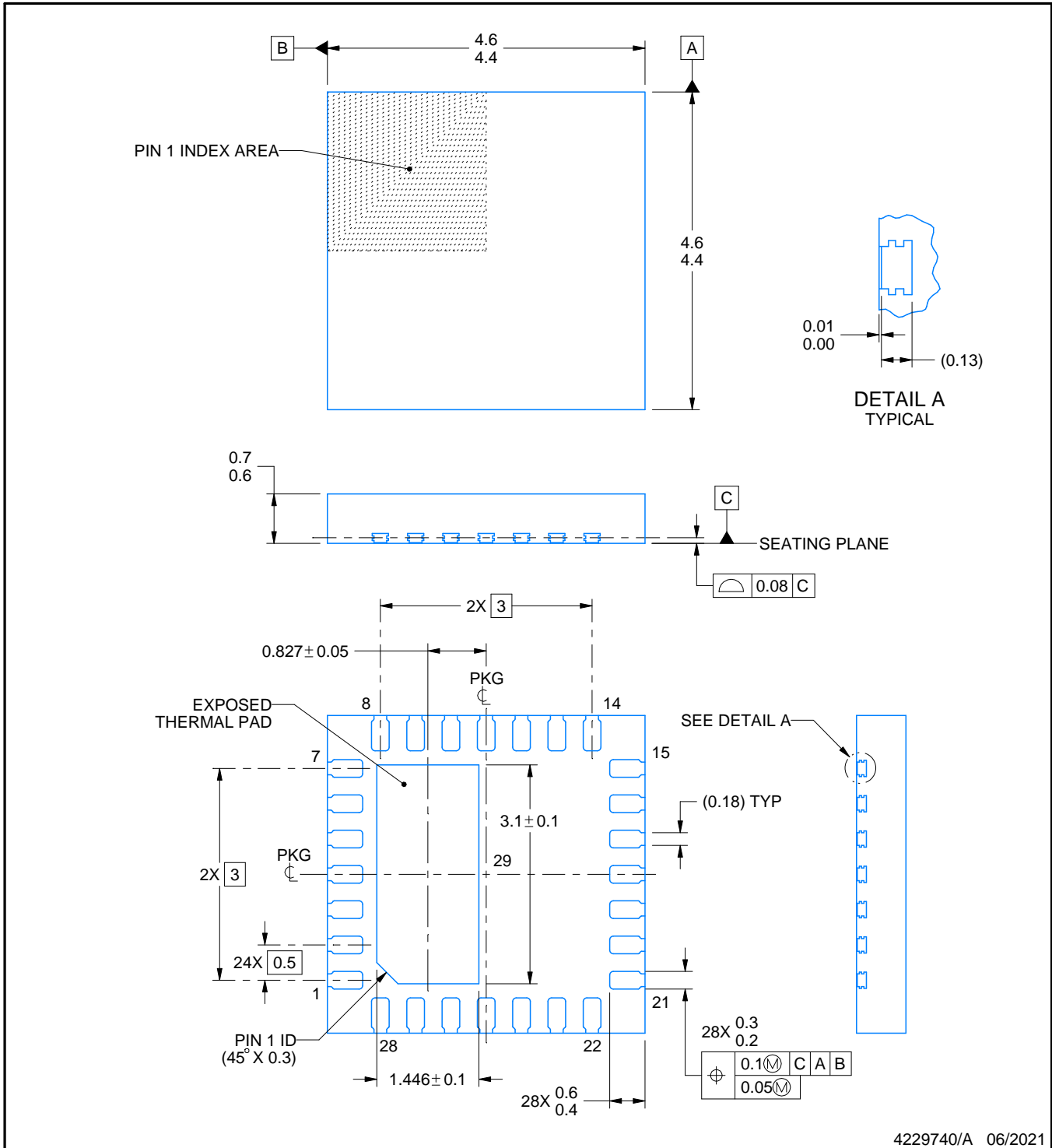
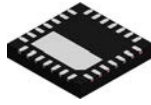
(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.



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NOTES:

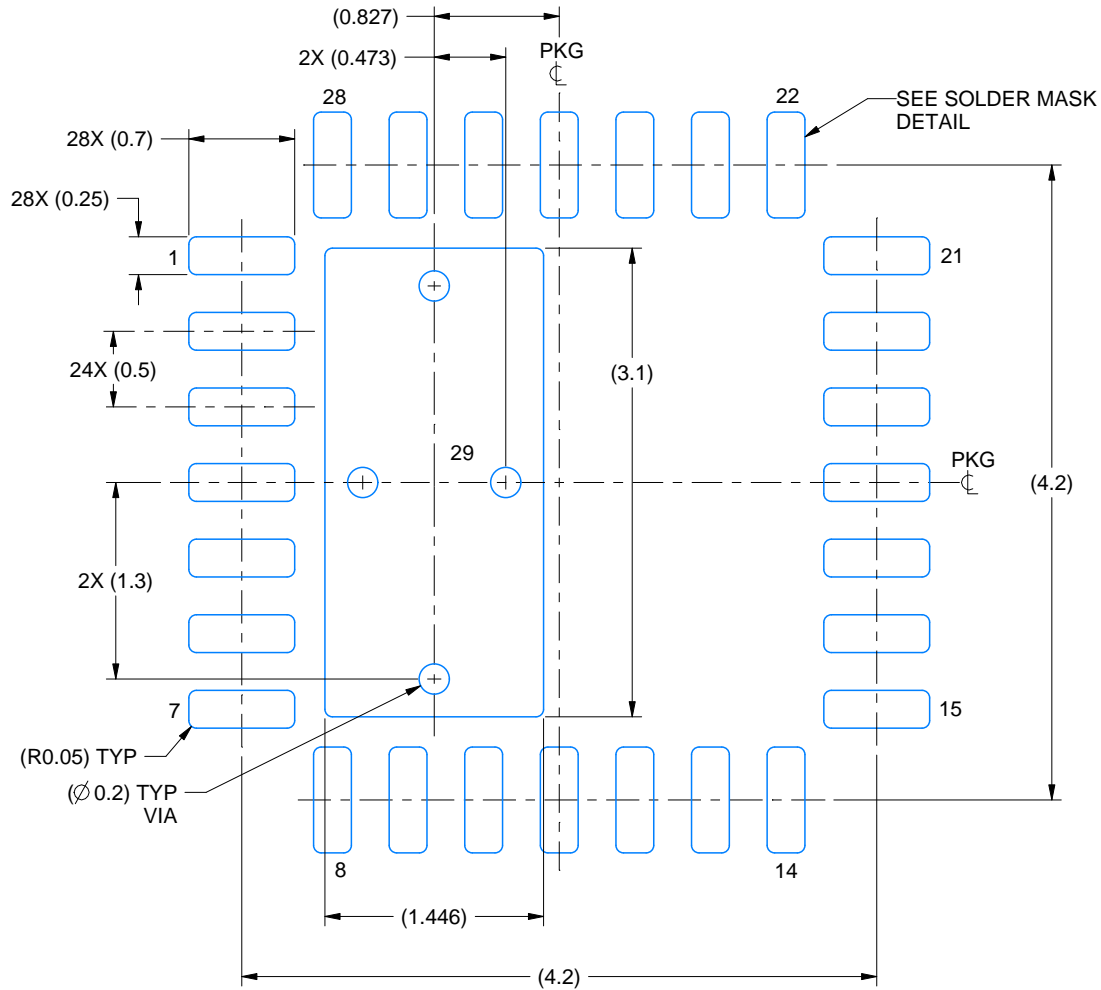
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

# EXAMPLE BOARD LAYOUT

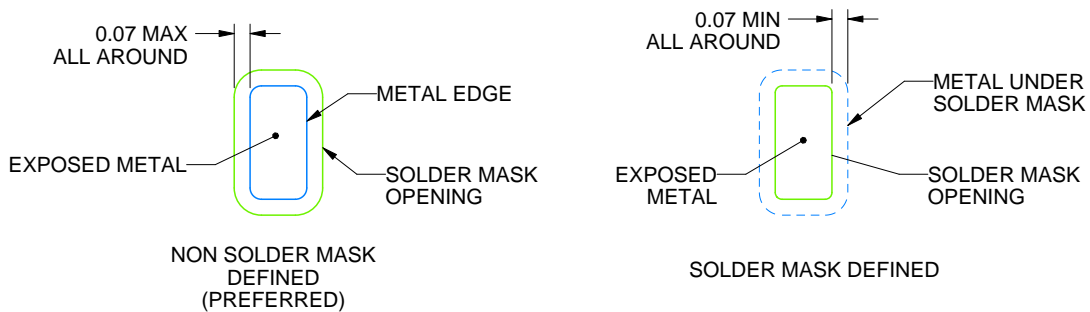
VBA0028A

WQFN-FCRLF - 0.7 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE: 20X



SOLDER MASK DETAILS

4229740/A 06/2021

NOTES: (continued)

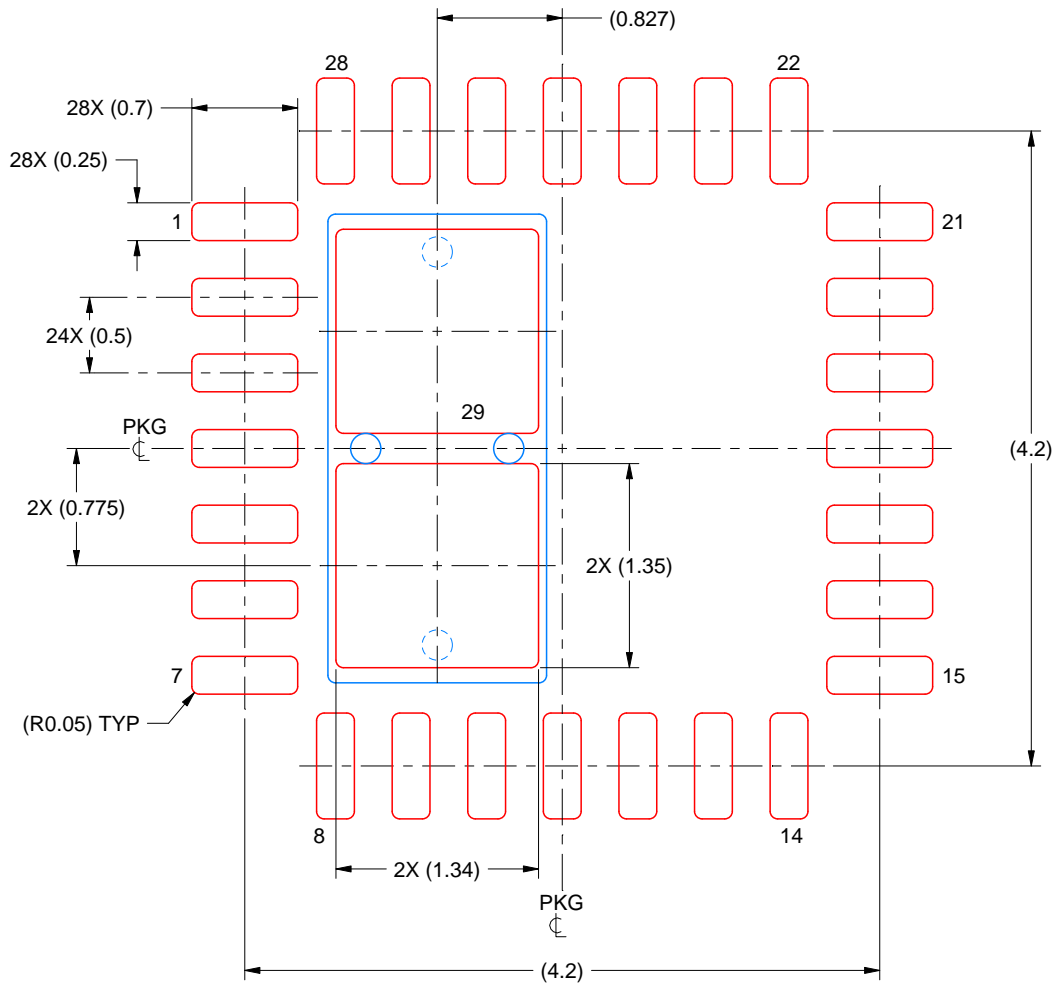
- This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 ([www.ti.com/lit/sluea271](http://www.ti.com/lit/sluea271)).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

# EXAMPLE STENCIL DESIGN

VBA0028A

WQFN-FCRLF - 0.7 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE  
BASED ON 0.125 MM THICK STENCIL  
SCALE: 20X

EXPOSED PAD 29  
35% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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